

EE 115A

Final Exam

Fall 2006

Your Name:

Solutions

Name of Person to Your Left:

Name of Person to Your Right:

Time Limit: 3 Hours

Where applicable, place answers inside designated boxes.

Use all approximations specified in each problem.

1. 5

2. 5

3. 10

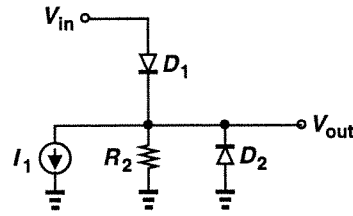
4. 8

5. 12

6. 10

Total: 50

1. Plot V_{out} and I_{D2} as a function of V_{in} as V_{in} goes from $-\infty$ to $+\infty$. Determine the coordinates of each break point in the plots. Assume a constant-voltage diode model and $I_1 R_2 > V_{D,on}$.



$V_{in} \rightarrow -\infty$

D_1 off

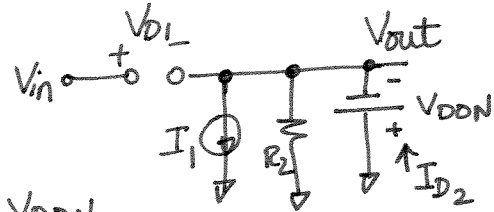
$$V_{out} = -V_{D,on}$$

D_2 ON

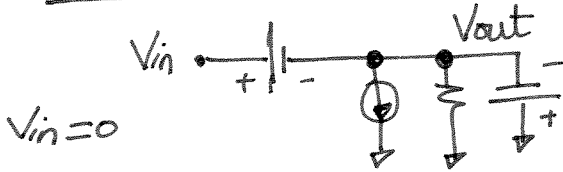
$$V_{D1} < V_{D,on} \rightarrow V_{in} - V_{out} < V_{D,on}$$

$$\rightarrow V_{in} < 0$$

$$I_1 = I_{D2} + \frac{V_{D,on}}{R_2} \rightarrow I_{D2} = \frac{1}{R_2} (I_1 R_2 - V_{D,on}) > 0$$

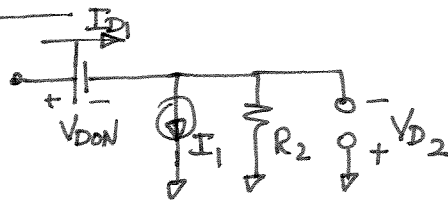


at $V_{in} = 0$ D_1 turns on:



D_1 : ON

D_2 : OFF



$$V_{out} = V_{in} - V_{D,on}$$

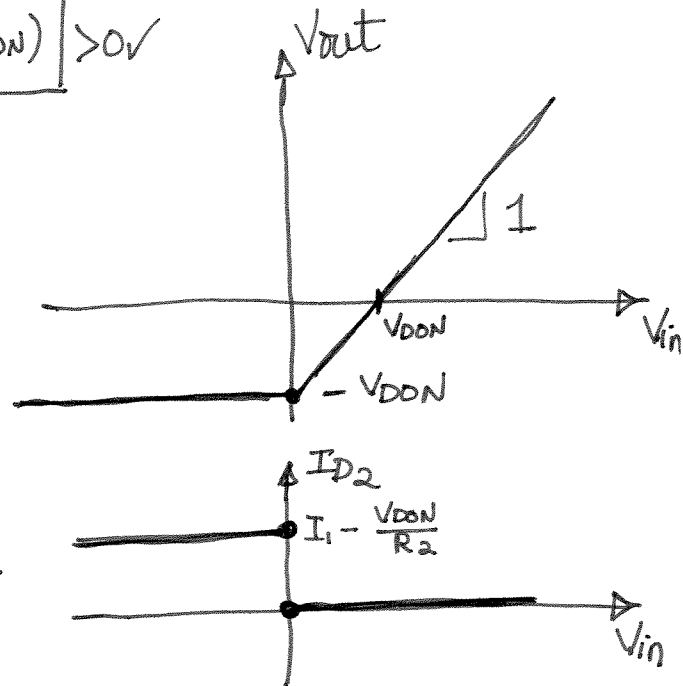
$$I_{D2} = 0$$

$$V_{D2} = -V_{out} < V_{D,on} \rightarrow -V_{in} + V_{D,on} < V_{D,on}$$

$$\rightarrow V_{in} > 0$$

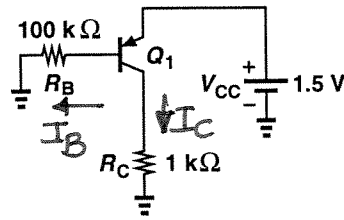
$$I_{D1} = I_1 + \frac{V_{out}}{R_2} > 0$$

$$I_{D1} = I_1 + \frac{V_{in}}{R_2} - \frac{V_{D,on}}{R_2} > 0 \rightarrow V_{in} > \underbrace{-(I_1 R_2 - V_{D,on})}_{\text{negative}}$$



2. Determine the value of β that places Q_1 at the edge of saturation. Assume $V_A = \infty$ and $I_S = 8 \times 10^{-16}$

A.



edge of saturation $\rightarrow V_C = V_B$

$$V_B = 100K \cdot I_B$$

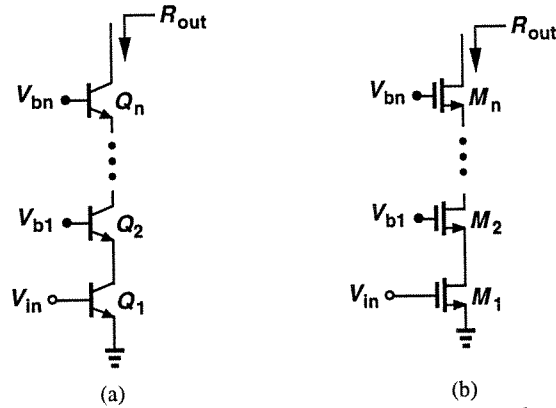
$$V_C = 1K \cdot I_C$$

$$I_C = \beta \cdot I_B$$

$$\Rightarrow 100K \frac{I_C}{\beta} = 1K I_C \Rightarrow \boxed{\beta = 100}$$

$$\beta = \boxed{100}$$

3. Compute the output impedance of each circuit as $n \rightarrow \infty$. Assume $V_A < \infty$, $\lambda > 0$, and $\beta < \infty$.



a) R_n : R_{out} when we have 'n' transistors.

$$R_1 = r_o \quad R_2 = r_o (1 + g_m (r_o \parallel r_{\pi})) + r_o \parallel r_{\pi}, \dots, R_n = r_o (1 + g_m (R_{n-1} \parallel r_{\pi})) + R_{n-1} \parallel r_{\pi}$$

as $n \rightarrow \infty$; $R_n = R_{n-1}$ so

$$R_{\infty} = r_o (1 + g_m (R_{\infty} \parallel r_{\pi})) + R_{\infty} \parallel r_{\pi}$$

$$\Rightarrow R_{\infty} = r_o \cdot \frac{1 + \beta + \sqrt{(1 + \beta)^2 + 4 \frac{r_{\pi}}{r_o}}}{2} \simeq r_o (1 + \beta)$$

b) $R_1 = r_o$; $R_2 = r_o (1 + g_m r_o) + r_o$; \dots ; $R_n = r_o (1 + g_m R_{n-1}) + R_{n-1}$

$$R_n = r_o + (1 + g_m r_o) R_{n-1}$$

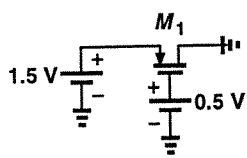
Since $1 + g_m r_o > 1$; R_n keeps increasing as $n \rightarrow \infty$

so $R_{\infty} = \infty$

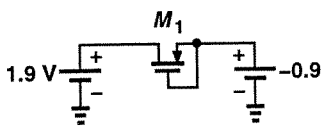
(a): $R_{out} = \simeq r_o (1 + \beta)$

(b): $R_{out} = \infty$

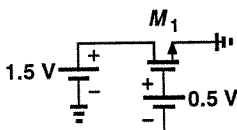
4. Determine the region of operation of each transistor (off, triode region, or saturation region). Assume a threshold voltage of 0.4 V for NMOS devices and -0.4 V for PMOS devices.



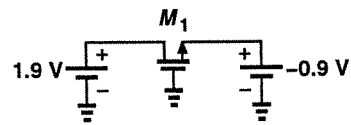
(a)



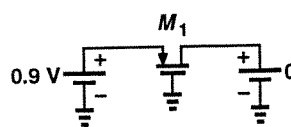
(b)



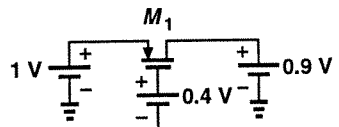
(c)



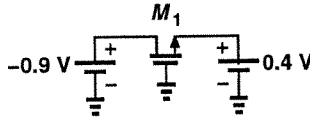
(d)



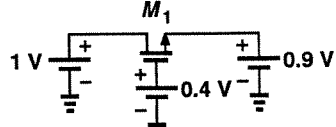
(e)



(f)



(g)



(h)

a) PMOS

$$V_{GD} = 0.5 > V_{THP}$$

$$V_{GS} = -1 < V_{THP}$$

SAT

b) PMOS

$$V_{GD} = 0 > V_{THP}$$

$$V_{GS} = -1 < V_{THP}$$

SAT

c) NMOS

$$V_{GD} = -1 < V_{THN}$$

$$V_{GS} = 0.5 > V_{THN}$$

SAT

d) NMOS

$$V_{GD} = -1.9 < V_{THN}$$

$$V_{GS} = 0.9 > V_{THN}$$

SAT

e) PMOS

$$V_{GS} = -0.9 < V_{THP}$$

$$V_{GD} = -0.4 = V_{THP}$$

Edge of Sat

f) PMOS

$$V_{GS} = -0.5 < V_{THP}$$

$$V_{GD} = -0.5 < V_{THP}$$

TRIODE

g) NMOS

$$V_{GD} = -0.4 < V_{THN}$$

$$V_{GS} = 0.9 > V_{THN}$$

SAT

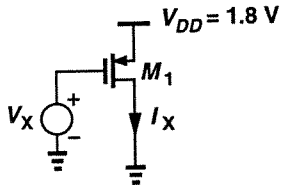
h) NMOS

$$V_{GS} = -0.5 < V_{THN}$$

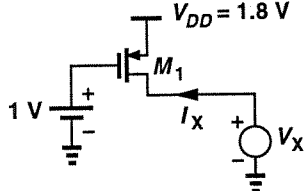
OFF

(a) SAT (b) SAT (c) SAT (d) SAT (e) EDGE (f) TRIODE (g) SAT (h) OFF

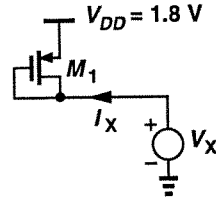
5. Plot I_X as a function of V_X as V_X varies from 0 to 1.8 V. Assume $\lambda = 0$ and $V_{THP} = -0.4$ V. Determine at what value of V_X the device changes its region of operation.



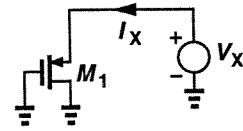
(a)



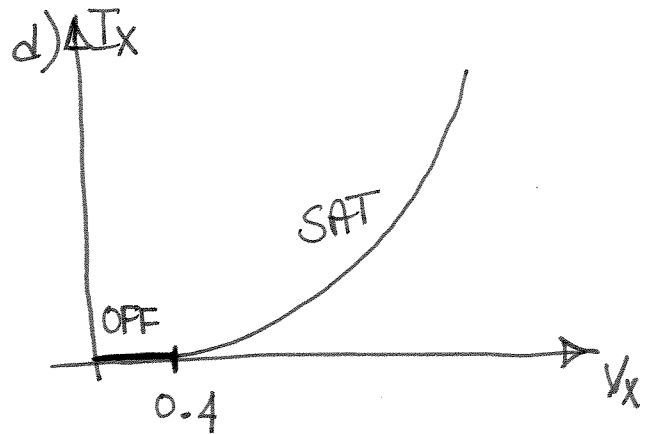
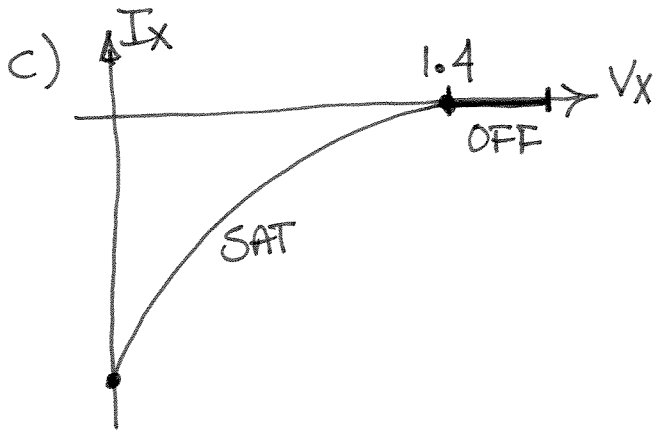
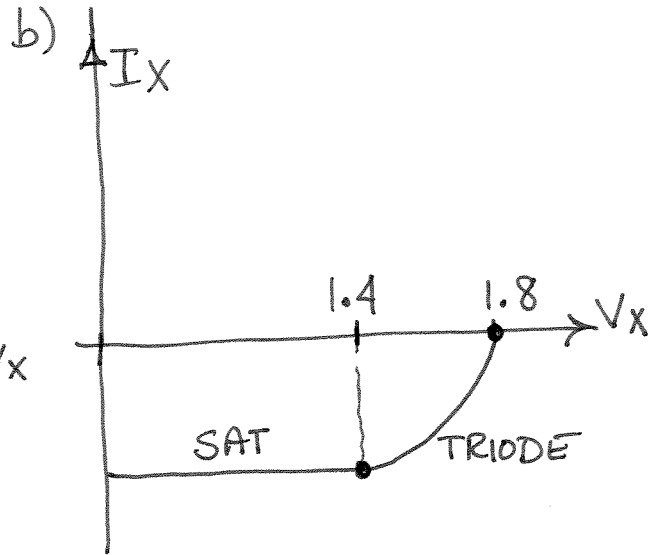
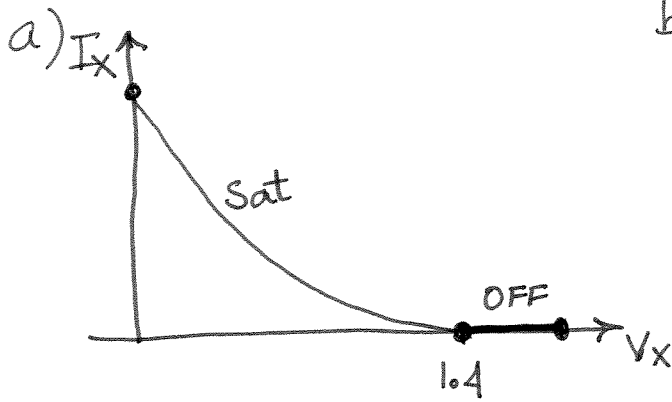
(b)



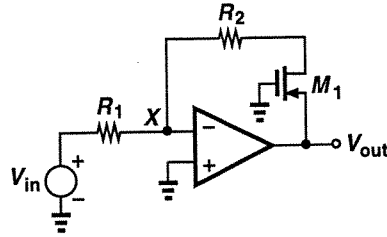
(c)



(d)



6. In the circuit shown below, assume the op amp is ideal and has an infinite gain. Also, $\lambda = 0$.
- (a) Assuming $R_2 = 0$, determine V_{out} as a function of V_{in} . For what range of V_{in} does the circuit produce a meaningful output, i.e., M_1 remains on?
- (b) Now assume $R_2 > 0$ and calculate the value of V_{in} that places M_1 at the edge of saturation.

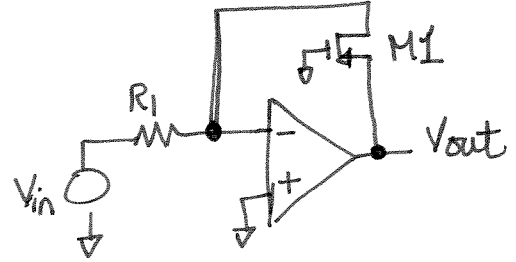


a)

$$\frac{1}{2} \mu_p C_{ox} \frac{W}{L} (V_{SG} - |V_{THP}|)^2 = -\frac{V_{in}}{R_1}$$

$$V_{SG} = V_{out}$$

$$\Rightarrow V_{out} = \sqrt{\frac{-2V_{in}}{R_1 \mu_p C_{ox} \frac{W}{L}}} + |V_{THP}|$$



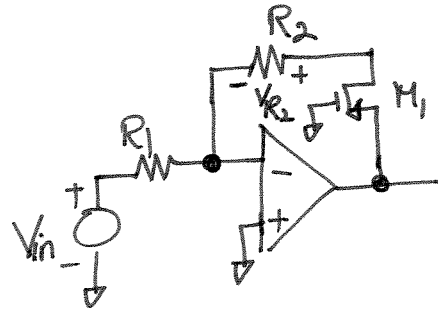
$$V_{in} \leq 0$$

b)

$$V_{R_2} = -\frac{V_{in}}{R_1} \cdot R_2$$

$$\text{EDGE OF SAT: } V_{R_2} = |V_{THP}|$$

$$\Rightarrow V_{in} = -\frac{R_1}{R_2} |V_{THP}|$$



$(a) V_{out} = \sqrt{\frac{-2V_{in}}{R_1 \mu_p C_{ox} \frac{W}{L}}} + V_{THP} \quad V_{in} \leq 0$	$(b) V_{in} = -\frac{R_1}{R_2} V_{THP} $
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