

UCLA
DEPARTMENT OF ELECTRICAL ENGINEERING
Spring 2008

Your Name (LAST, Middle, First): _____

EE102: SYSTEMS & SIGNALS

FINAL EXAMINATION

June 13, 2008

Table 1: Points per problem (Total 100)

Problem	part (i)	part (ii)	part (iii)	part (iv)
1	10	10	10	
2	10			
3	5			
4	5			
5	5	5	5	5
6	5	5	5	5
7	5	5		

(Question 1)(30pts)

Consider the cascaded combination S_{12} of **LTI,C** systems S_1 and S_2 :

$$x(t) \rightarrow [S_1] \rightarrow [S_2] \rightarrow z(t)$$

where

$$x(t) = U(t)$$

and

$$z(t) = [\cos(t) + 2\sin(t) - e^{-t/2}]U(t)$$

(i)(10 pts) Compute the system function $H_{12}(s)$ of the cascaded system. Now, suppose that the systems S_1 is described by the IPOP relation:

$$y(t) = x(t) - \frac{1}{2} \int_{-\infty}^t e^{-\frac{1}{2}(t-\sigma)} x(\sigma) d\sigma, t > -\infty$$

Find the IRF $h_2(t)$ of S_2 .

(ii)(10 pts) Derive the FRF $H_2(i\omega)$ of system S_2 —from its System Function $H_2(s)$ — if possible. If it is NOT possible, what would you do? Now given:

$$x(t) = 1 + \sin(2t) \rightarrow [S_2] \rightarrow y(t)$$

Find output $y(t)$ for $t \in (-\infty, \infty)$.

(iii)(10 pts) Consider

$$v(t) \rightarrow [S_1] \rightarrow w(t)$$

where S_1 is described in part (i) and

$$v(t) = e^{2t}, t < 0,$$

and

$$v(t) = e^{-2t}, t \geq 0.$$

Your problem is to find the output $w(t)$ —by any method which you are most comfortable with.

(Question 2)(10 points) The system function $H(s)$ of LTI,C is

$$H(s) = \frac{s}{s+1}$$

and its input is

$$x(t) = e^{-t}U(t-1) + \cos(t), t \in (-\infty, \infty)$$

Find the corresponding output $y(t)$.

(Question 3)(5 points) Assume the S is a **LTI** system with IRF $h(t)$. Is it possible to have the following situation?

$$3 + 5\cos(20t) \rightarrow [S] \rightarrow \sin(20t + \theta)$$

Explain why/why not.

(Question 4)(5 points) If signals $x(t)$, $y(t)$, $z(t)$ are such that

$$\int_{-\infty}^{\infty} x(t)dt = A$$

$$\int_{-\infty}^{\infty} y(t)dt = B$$

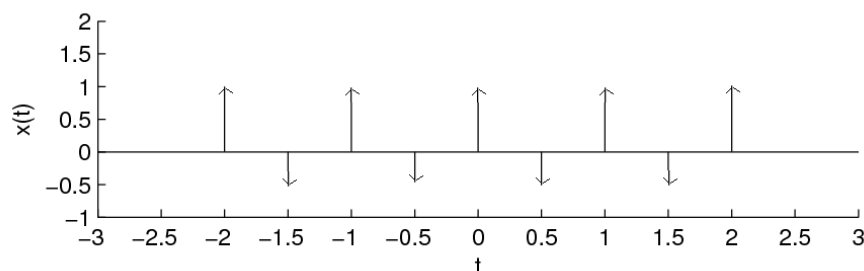
and $z(t) = x(t) * y(t)$ ($*$ denotes convolution). Evaluate the following integral

$$\int_{-\infty}^{\infty} z(t)dt = ?$$

for given A and B .

(Question 5)(20 points) Let $x(t)$ be the periodic signal defined as:

$$x(t) = \sum_{n=-\infty}^{\infty} (\delta(t - nT) - 0.5\delta(t - nT - 0.5))$$



(i)(5 pts) Write down Fourier series representation of $x(t)$.

(ii)(5 pts) Find Fourier transform of $x(t)$.

(iii)(5 pts) The $x(t)$ is applied to the **LTI-C** system described by

$$\frac{dy(t)}{dt} + 2y(t) = 2x(t), y(0) = 0 = x(0).$$

Find the output $y(t)$ in closed form. *Hint: Use the fact that output is also periodic signal. Thus, you should compute $y(t)$ only in one period.*

(iv)(5 pts) Write down Fourier series representation of $y(t)$.

Find MSE when $y(t)$ is approximated by:

$$y_2(t) = \sum_{n=-2}^2 Y_n e^{in\omega_0 t}$$

(Question 6)(20 points) Consider a **LTI** system S with IPOP relation:

$$y(t) = \int_t^{t+1} x(\tau) d\tau - \int_{t-1}^t x(\tau) d\tau.$$

(i)(5 pts) Find IRF $h(t)$ of system S . Also, sketch $h(t)$.

(ii)(5 pts) Find frequency response function $H(i\omega)$ of system S .

(iii)(5 pts) If input is

$$x_1(t) = \cos(\pi t - 1)$$

find the corresponding output $y_1(t)$.

(iv)(5 pts) Given the input

$$x_2(t) = \sum_{n=-\infty}^{\infty} e^{-|t-n|} \cos(t-n),$$

Show that $x_2(t)$ is periodic and find its period.

Find the output $y_2(t)$ due to input $x_2(t)$.

(Question 7)(10 points) The signal $x(t) = \cos(200\pi t)\cos(1000\pi t)$ is sampled by multiplying it with

$$p(t) = \sum_{n=-\infty}^{\infty} \delta(t - nT)$$

Sampling period is denoted by T .

(i)(5 pts)

If $1/T = 2000$ Hz, sketch the magnitude spectrum of $x_{\text{sampled}}(t) = x(t)p(t)$.

Is it possible to perfectly reconstruct $x(t)$ from $x_{\text{sampled}}(t)$?

If $1/T = 900$ Hz, sketch the magnitude spectrum of $x_{\text{sampled}}(t) = x(t)p(t)$.

Is it possible to perfectly reconstruct $x(t)$ from $x_{\text{sampled}}(t)$?

What is the maximum sampling period T that provides perfect reconstruction of $x(t)$ from $x_{\text{sampled}}(t)$?

(ii)(5 pts) For sampling period that provides perfect reconstruction, suggest the Frequency Response Function (FRF) of the reconstruction filter $H(i\omega)$.

Practice Final solution, WINTER 2011, EE 102

1. (i)

$$X(s) = 1/s, Z(s) = \frac{s+2}{s^2+1} - \frac{1}{s+1/2} = \frac{5s}{(s^2+1)(2s+1)}.$$

$$H_{12}(s) = \frac{Z}{X} = \frac{5s^2}{(s^2+1)(2s+1)}$$

$$h_1(t) = \delta(t) - \frac{1}{2}e^{-t/2}U(t).$$

$$H_1(s) = 1 - \frac{1}{2s+1} = \frac{2s}{2s+1}$$

$$H_2(s) = \frac{H_{12}(s)}{H_1(s)} = \frac{5s/2}{s^2+1}.$$

$$h_2(t) = \frac{5}{2} \cos t U(t).$$

(ii) No, as the roots are on the $i\omega$ axis.

$$H_2(i\omega) = \frac{5}{2} \left[\frac{\pi}{2} (\delta(\omega-1) - \delta(\omega+1)) + \frac{i\omega}{1-\omega^2} \right]$$

$$X(i\omega) = 2\pi\delta(\omega) + i\pi[\delta(\omega+2) - \delta(\omega-2)]$$

$$Y(i\omega) = X(i\omega)H_2(i\omega) = \frac{i\omega 5/2}{1-\omega^2} [2\pi\delta(\omega) + i\pi[\delta(\omega+2) - \delta(\omega-2)]]$$

$$= \frac{-5/2\pi}{-3} [-2\delta(\omega+2) - 2\delta(\omega-2)] = \frac{-5}{3} \cos 2t.$$

(iii)

$$\begin{aligned} v(t) = e^{-2|t|}, S_1 : w(t) &= v(t) - \frac{1}{2} \int_{-\infty}^t e^{-\frac{t-\sigma}{2}} v(\sigma) d\sigma = e^{-2|t|} - \frac{1}{2} \int_{-\infty}^t e^{-2|\tau|} e^{-\frac{(t-\tau)}{2}} d\tau \\ &= e^{-2|t|} - \frac{1}{2} e^{-t/2} \int_{-\infty}^t e^{-2|\tau|} e^{\frac{\tau}{2}} d\tau \end{aligned}$$

For $t > 0$:

$$w(t) = e^{-2|t|} - \frac{1}{2} e^{-t/2} \left[\int_{-\infty}^0 e^{2\tau} e^{\frac{\tau}{2}} d\tau + \int_0^t e^{-2\tau} e^{\frac{\tau}{2}} d\tau \right] = e^{-2|t|} + \left(\frac{-8}{15} e^{-t/2} + \frac{1}{3} e^{-2t} \right)$$

For $t \leq 0$:

$$w(t) = e^{-2|t|} - \frac{1}{2}e^{-t/2} \int_{-\infty}^t e^{2\tau} e^{\frac{\tau}{2}} d\tau = e^{-2|t|} - \frac{1}{5}e^{2t}$$

2.

$$H(s) = \frac{s}{s+1} = 1 - \frac{1}{s+1}, h(t) = \delta(t) - e^{-t}U(t).$$

$$h(t) = \delta(t) - e^{-t}U(t).$$

Let $x_1(t) = e^{-t}U(t-1)$ and $x_2(t) = \cos t$. Let $y_1(t)$ be the output caused by $x_1(t)$ and $y_2(t)$ be the output caused by $x_2(t)$. $y_1(t)$ can be calculated using Laplace Transform as $h(t)$ and $x_1(t)$ equal 0 for $t < 0$. $y_2(t)$ has to be calculated using Fourier Transform or in time domain as $x_2(t) \neq 0$ for $-\infty < t < 0$.

$$x_1(t) = e^{-1}e^{-(t-1)}U(t-1) \rightarrow X_1(s) = \frac{e^{-1}e^{-s}}{s+1} \rightarrow Y_1(s) = H(s)X_1(s) = \frac{e^{-1}e^{-s}s}{(s+1)^2}.$$

$$\text{Let } V(s) = \frac{Y_1(s)}{e^{-1}e^{-s}} = \frac{s}{(s+1)^2} \text{ and } G(s) = \frac{V(s)}{s} = \frac{1}{(s+1)^2}.$$

$$G(s) = \frac{1}{(s+1)^2} = \frac{-d}{ds} \frac{1}{s+1}$$

$$\therefore g(t) = te^{-t}U(t)$$

$$L\left\{\frac{d}{dt}g(t)\right\} = sG(s) - g(0) = sG(s) = V(s)$$

$$\therefore v(t) = \frac{d}{dt}g(t) = (-te^{-t} + e^{-t})U(t) + te^{-t}\delta(t) = (-te^{-t} + e^{-t})U(t) = e^{-t}(1-t)U(t).$$

$$\therefore y_1(t) = e^{-1}e^{-(t-1)}(1 - (t-1))U(t-1) = e^{-t}(2-t)U(t-1).$$

$$X_2(i\omega) = \pi[\delta(\omega-1) + \delta(\omega+1)], H(i\omega) = \frac{i\omega}{1+i\omega}.$$

$$Y_2(i\omega) = \pi\left[\frac{i}{1+i}\delta(\omega-1) - \frac{i}{1-i}\delta(\omega+1)\right] = \pi\left[\frac{1+i}{2}\delta(\omega-1) + \frac{1-i}{2}\delta(\omega+1)\right]$$

$$\therefore y_2(t) = \frac{\cos t - \sin t}{2}.$$

$$\therefore y(t) = e^{-t}(2-t)U(t-1) + \frac{\cos t - \sin t}{2}.$$

3. Yes.

$$3 \rightarrow H(0) = 0 \rightarrow 0.$$

$5 \cos(20t) \rightarrow H(i\omega) \rightarrow \sin(20t+\theta)$ is valid, as an LTI system with a sinusoidal input produces a sinusoidal output with the same frequency.

4.

$$\int_{-\infty}^{\infty} x(t)dt = A = X(0), \int_{-\infty}^{\infty} y(t)dt = B = Y(0).$$

$$\int_{-\infty}^{\infty} x(t) * y(t)dt \rightarrow Z(0) = \int_{-\infty}^{\infty} z(t)dt$$

$$Z(0) = X(0)Y(0) = AB.$$

5i.

$$F_0 = \frac{1}{T} \int_T f(t)dt = 0.5.$$

$$F_n = \frac{1}{T} \int_T e^{-in\omega_0 t} [\delta(t) - 0.5\delta(t-0.5)]dt = 1 - 0.5e^{-in\omega_0 0.5} = 1 - 0.5(-1)^n.$$

$$x(t) = \frac{1}{2} + \sum_{n=-\infty}^{\infty} (1 - 0.5(-1)^n) e^{i2\pi n t}.$$

ii)

$$X(i\omega) = \pi\delta(\omega) + 2\pi \sum_{n=-\infty}^{\infty} (1 - 0.5(-1)^n) \delta(\omega - 2\pi n).$$

iii)

$$i\omega Y(i\omega) + 2Y(i\omega) = 2X(i\omega)$$

$$\begin{aligned} y(t) &= h(t) * x(t) = 2e^{-2t}U(t) * \sum_{-\infty}^{\infty} \delta(t-n) - \frac{1}{2}\delta(t-n-\frac{1}{2}) \\ &= \sum_{-\infty}^{\infty} 2e^{-2(t-n)}U(t-n) - e^{-2(t-n-\frac{1}{2})}U(t-n-\frac{1}{2}) \\ &= e^{-2t} \sum_{-\infty}^{\infty} 2e^{2n}U(t-n) - e^{2n}U(t-n-\frac{1}{2}) \end{aligned}$$

For $0 < t < \frac{1}{2}$:

$$y(t) = e^{-2t} \frac{2 - e^{-1}}{1 - e^{-2}}$$

For $\frac{1}{2} \leq t \leq 1$:

$$y(t) = e^{-2t} \frac{2 - e^1}{1 - e^{-2}}.$$

iv)

$$y(t) = \frac{1}{2} + \sum_{n=-\infty}^{\infty} (1 - 0.5(-1)^n) e^{i2\pi tn} \frac{2}{2 + i2\pi n}.$$

$$MSE = \frac{1}{T} \int_T |y(t)|^2 dt - \sum_{n=-2}^2 |Y_n|^2,$$

$$\frac{1}{T} \int_T |y(t)|^2 dt = \int_0^{0.5} e^{-4t} \left(\frac{2 - e^{-1}}{1 - e^{-2}} \right)^2 dt + \int_{0.5}^1 e^{-4t} \left(\frac{2 - e^1}{1 - e^{-2}} \right)^2 dt = 0.790376,$$

$$\sum_{n=-2}^2 |Y_n|^2 = 0.67635,$$

$$\therefore MSE = 0.114025.$$

6i.

$$y(t) = \int_{-\infty}^{\infty} [U(\tau-t)U(t+1-\tau) - U(\tau-t+1)U(t-\tau)]x(\tau)d\tau$$

$$h(t-\tau) = U(\tau-t)U(t+1-\tau) - U(\tau-t+1)U(t-\tau)]$$

$$h(t) = U(-t)U(t+1) - U(-t+1)U(t)$$

ii)

$$H(i\omega) = \int_{-1}^0 e^{-i\omega t} dt - \int_0^1 e^{-i\omega t} dt = \frac{1}{-i\omega}[1 - e^{i\omega}] - \frac{1}{-i\omega}[e^{-i\omega} - 1] = \frac{2}{i\omega}[\cos \omega - 1].$$

iii)

$$y(t) = \int_t^{t+1} \cos(\pi\tau - 1) d\tau - \int_{t-1}^t \cos(\pi\tau - 1) d\tau = \frac{-4}{\pi} \sin(\pi t - 1)$$

iv)

$$x_2(t) = \sum_{n=-\infty}^{\infty} e^{-|t-n|} \cos(t-n),$$

$$x_2(t+T) = \sum_{n=-\infty}^{\infty} e^{-|t+T-n|} \cos(t+T-n),$$

Let $\tau = n - T$:

$$x_2(t+T) = \sum_{n=-\infty}^{\infty} e^{-|t-\tau|} \cos(t-\tau).$$

Note that the indices of the summation are $n = \dots, -2, -1, 0, 1, 2, \dots$, which correspond to $\tau = \dots, -2 - T, -1 - T, -T, 1 - T, 2 - T, \dots$. For $x_2(t)$ to be equal to $x_2(t+T)$, T should be equal to 1.

$$y_2(t) = \int_t^{t+1} x_2(\tau) d\tau - \int_{t-1}^t x_2(\tau) d\tau$$

Since $x_2(t)$ is periodic with period 1, then $\int_t^{t+1} x_2(\tau) d\tau = \int_{t-1}^t x_2(\tau) d\tau$.

$$\therefore y_2(t) = 0.$$