

UCLA DEPARTMENT OF ELECTRICAL ENGINEERING

EE102: SYSTEMS & SIGNALS

Midterm Examination I

February 6, 2018

Duration: 1 hr 50 mins.

INSTRUCTIONS:

- The exam has 6 problems and 14 pages.
- The exam is closed-book.
- One cheat sheet of A4 size is allowed.
- Calculator is NOT allowed.
- Write your discussion session in the top-right corner. ↗↗

Your name: _____**Student ID:** _____

DISCUSSION

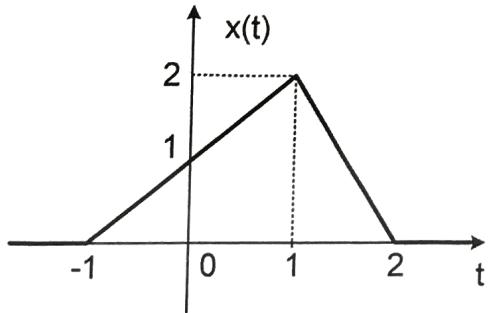
Table 1: Score Table

Problem	a	b	c	d	Score
1	4	4	4		12
2	2	2	8		12
3	6	2	4	6	18
4	8	8			16
5	10	6			16
6	5	5	6		16
Total					90

12
12
18
9
15
9
75

12

Problem 1 (12 pts) Consider the following signal $x(t)$ for (a), (b)



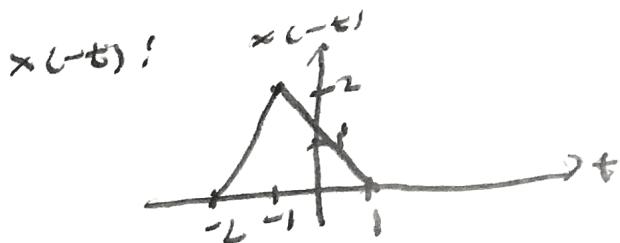
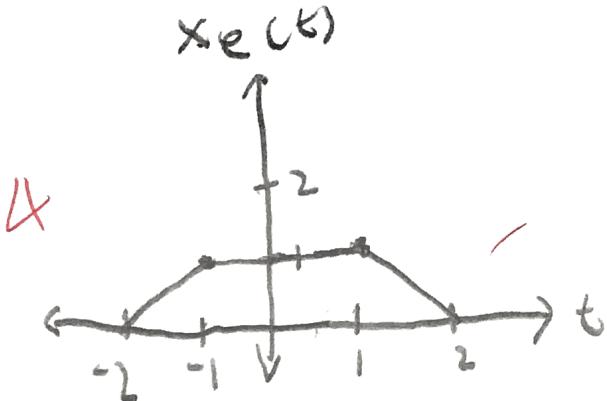
(a) (4 pts) Sketch even and odd decompositions $x_e(t)$ and $x_o(t)$.

(b) (4 pts) Sketch $x(-2t + 3)$.

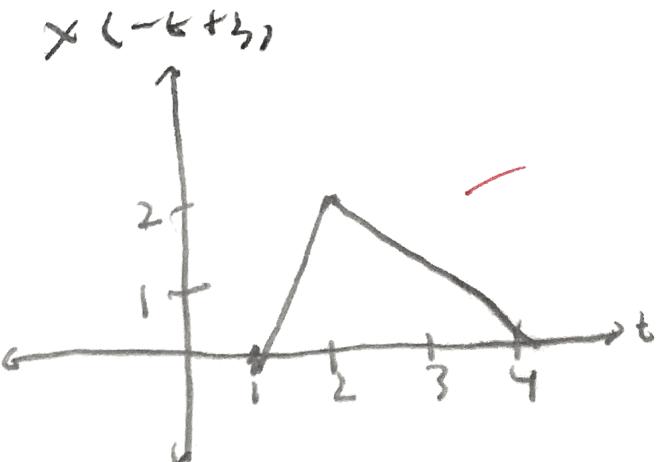
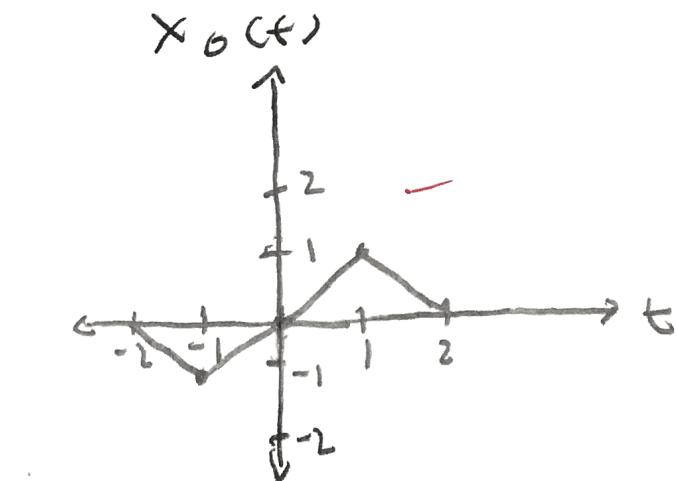
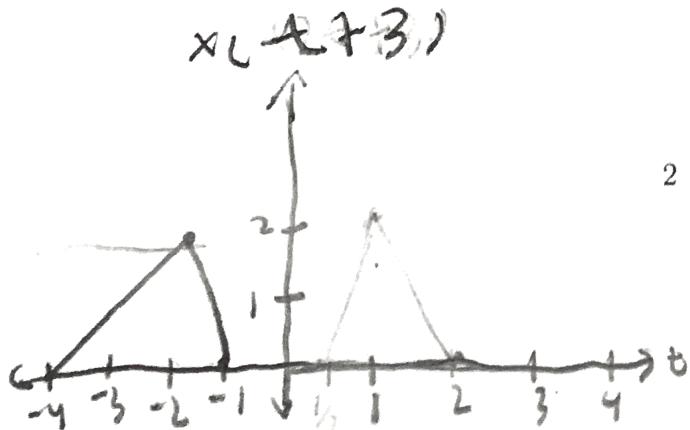
(c) (4 pts) Sketch $x(t/3 + 2)$.

$$a. \quad x_e(t) = \frac{1}{2}[x(t) + x(-t)]$$

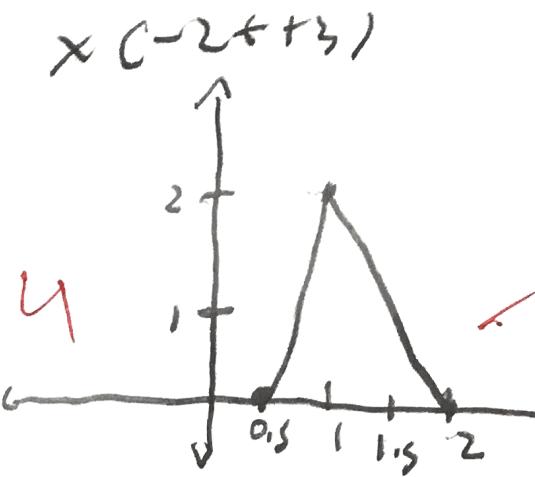
$$x_o(t) = \frac{1}{2}[x(t) - x(-t)]$$



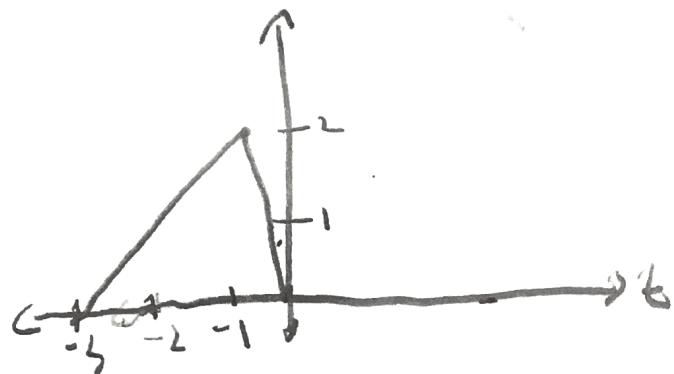
b.



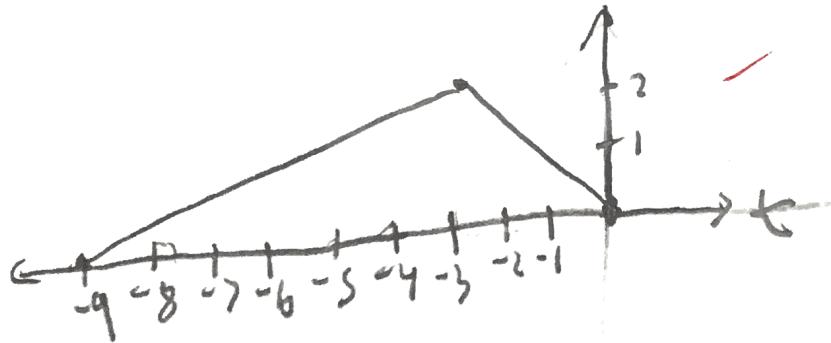
b.
cont.



c. $x(t+2)$



$x(\frac{1}{3}t+2)$ u



12

Problem 2 (12 pts) In this problem, we identify system properties from the impulse response function:

$$h(t, \tau) = e^{-(t-\tau)} u(t - \tau) u(t) \quad (1)$$

- (a) (2 pts) Is the system TV or TI? Explain.
- (b) (2 pts) Is it C or NC? Explain.
- (c) (8 pts) Find the output $y(t)$ if the input is $x(t) = (t - 2)u(t - 2)$.

a. $h(t, \tau)$ cannot be expressed as a function of $t - \tau$, therefore, the system is TV.

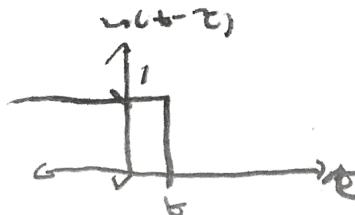
b. $h(t, \tau) h(t-\tau) = e^{-(t-\tau)} u(t - \tau) u(t) u(t - \tau)$
2 $= e^{-(t-\tau)} u(t - \tau) u(t)$ since $u(t - \tau) u(t - \tau) = u(t - \tau)$
 $= h(t, \tau).$

This means that $h(t, \tau) = 0$ for $t < \tau$, which means the system is causal (C)

c. $y(t) = \int_{-\infty}^{\infty} x(\tau) h(t, \tau) d\tau = \int_{-\infty}^{\infty} (t-2) u(t-2) e^{-(t-\tau)} u(t-\tau) u(t) d\tau$
2 $= \int_2^{\infty} (t-2) u(t-2) e^{-(t-\tau)} u(t-\tau) u(t) d\tau$ if $t < 2$, 0.0.
2 if $t < 2$:

$$\int_2^{\infty} (t-2) e^{-(t-\tau)} u(t-2) u(t) d\tau = 0$$

if $t < 2$:



if $t \geq 2$:

$$\int_2^\infty (t-2) e^{-(t-\tau)} u(t-\tau) u(\tau) d\tau$$

$$= \int_2^t (t-2) e^{-(t-\tau)} d\tau$$

$$= \int_2^t \tau e^{\tau-t} d\tau - \int_2^t \tau e^{-(t-\tau)} d\tau$$

$$\int_2^t \tau e^{\tau-t} d\tau = \tau e^{\tau-t} \Big|_2^t - \int_2^t e^{\tau-t} d\tau = \int_2^t$$

$\begin{matrix} 1 & 1 \\ u & dv \end{matrix}$

$$= t e^{\circ} - 2e^{-(t-2)} - e^{\tau-t} \Big|_2^t$$

$$\int_2^t \tau e^{-(t-\tau)} d\tau = t - \tau e^{-(t-\tau)} \Big|_2^t - 1$$
$$2e^{-(t-2)} \Big|_2^t = 2 - 2e^{-(t-2)}$$

$$\int_2^t (t-2) e^{-(t-\tau)} d\tau = t + e^{-(t-2)} - 3$$

$$y(t) \int_{-\infty}^\infty x(\tau) u(t, \tau) d\tau = \begin{cases} t + e^{-(t-2)} - 3 & t \geq 2 \\ 0 & t < 2 \end{cases}$$

Q

$$y(t) = [t + e^{-(t-2)} - 3] u(t-2)$$

Problem 3 (18 pts)Consider IPOP relation for an LTI system S :

$$y(t) = e^{-t} \int_{-\infty}^t e^{\tau} [\cos(\tau) \cos(t) + \sin(\tau) \sin(t)] x(\tau) d\tau$$

where $x(t)$ and $y(t)$ are input and output of the system, respectively.

- (a) (6 pts) Find the impulse response function $h(t)$.
- (b) (2 pts) Is the system C or NC? Provide justification.
- (c) (4 pts) Is this system BIBO stable? Provide justification.
- (d) (6 pts) Find Laplace transform $H(s)$ and ROC.

(Hint: Use the identity $\cos(A+B) = \cos(A)\cos(B) - \sin(A)\sin(B)$)

$$\begin{aligned} a. \quad x(t) &= y(t) \\ h(t) &= e^{-t} \int_{-\infty}^t e^{\tau} [\cos(\tau) \cos(t) + \sin(\tau) \sin(t)] \delta(\tau) d\tau \\ &= e^{-t} \int_{-\infty}^t e^{\tau} [\cos(t-\tau)] \delta(\tau) d\tau \\ &= e^{-t} \int_{-\infty}^t e^{\tau} \cos(t-\tau) \delta(\tau) d\tau \end{aligned}$$

$$\begin{aligned} b. \quad &\approx e^{-t} \cos(t) \int_{-\infty}^t \delta(\tau) d\tau \Rightarrow \begin{cases} 1 & \text{if } t > 0 \\ 0 & \text{if } t \leq 0 \end{cases} \\ &\approx e^{-t} \cos(t) u(t) \quad h(t) = e^{-t} \cos(t) u(t) \end{aligned}$$

b. The system is C (causal) This is because

$$\begin{aligned} h(0) \cdot u(0) &= e^{-0} \cos(0) u(0) u(0) \\ &= e^{-0} \cos(0) u(0) = h(0) \text{ since } u(0) \cdot u(0) = 0 \end{aligned}$$

In other words, when $t < 0$, $h(t) = 0$ this means the system is causal.

$$L. \int_{-\infty}^{\infty} h(t) dt = \int_{-\infty}^{\infty} e^{-t} \cos(t) u(t) dt$$

$$= \int_0^{\infty} e^{-t} \cos(t) dt = \left. e^{-t} \sin(t) \right|_0^{\infty}$$

$$= e^{-\infty} \sin(\infty) - e^0 \sin(0)$$

Since $\int_{-\infty}^{\infty} h(t) dt < \infty \Rightarrow \underline{z} 0 - 0 = 0 < \infty$

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d. $h(t) = e^{-t} \cos(t) u(t)$ shows, as $h(t) = h(u(t))$,

$$\text{Note: } L\{\cos(t) u(t)\} = \frac{s}{s^2 + 1}$$

from the frequency shifting property:

if $f(t) \xrightarrow{L} F(s)$

$$\text{then. } e^{-at} f(t) \xrightarrow{L} F(s+a)$$

therefore

$$H(s) = L\{e^{-t} \cos(t) u(t)\} = \frac{s+1}{(s+1)^2 + 1} = \frac{s+1}{s^2 + 2s + 2}$$

consider denominator polynomial.

$$s^2 + 2s + 2 : \text{find roots: } s = -2 \pm \sqrt{4-8} = -2 \pm \sqrt{-4}$$

$$= -2 \pm 2j = -1 \pm j$$

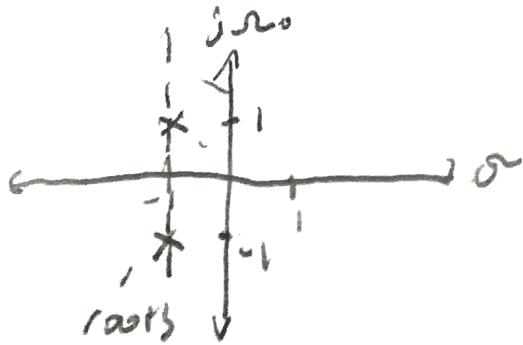
$$s = 0 + j\pi/2$$

rightmost root is at

$$s = -1$$

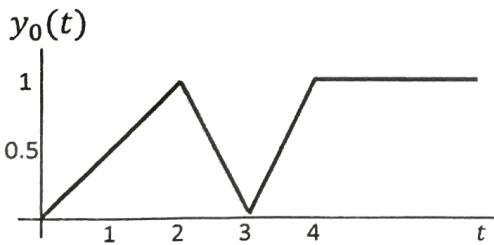
therefore the ROC is:

$$\text{ROC: } \text{Re}\{s\} > -1$$



9

Problem 4 (16 pts) Consider an LTI system S_0 with input $x_0(t)$ and the impulse response function $h_0(t)$. The corresponding output $y_0(t)$ is shown below:



8

(a) (8 pts) Consider an LTI system S_1 with input $x_1(t) = x_0(t+2)$ and IRF $h_1(t) = h_0(t-1)$. Express the output $y_1(t)$ as a function of $y_0(t)$ and then plot it.

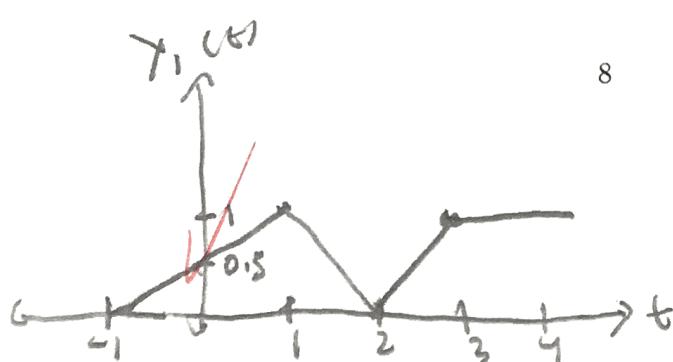
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(b) (8 pts) Consider an LTI system S_2 with input $x_2(t) = x_0(-t)$ and IRF $h_2(t) = h_0(-t)$. Express the output $y_2(t)$ as a function of $y_0(t)$ and then plot it.

a.

$$\begin{aligned}
 Y_1(t) &\approx \int_{-\infty}^{\infty} x_1(\tau) h_1(t-\tau) d\tau \\
 &= \int_{-\infty}^{\infty} x_0(\tau+2) h_0(t-1)-\tau d\tau \quad \text{Set } T=t-1 \\
 &= \int_{-\infty}^{\infty} x_0(\tau+2) h_0(T-\tau) d\tau \\
 &\approx y_0(T+2) \text{ since system is LTI} \\
 &= y_0((t-1)+2) = y_0(t+1)
 \end{aligned}$$

Y_1(t) = y_0(t+1)



$$b. y_2(t) = \int_{-\infty}^{\infty} x_0(\tau) h_0(t-\tau) d\tau$$

$$= \int_{-\infty}^{\infty} x_0(-\tau) h_0(t-(t-\tau)) d\tau$$

1

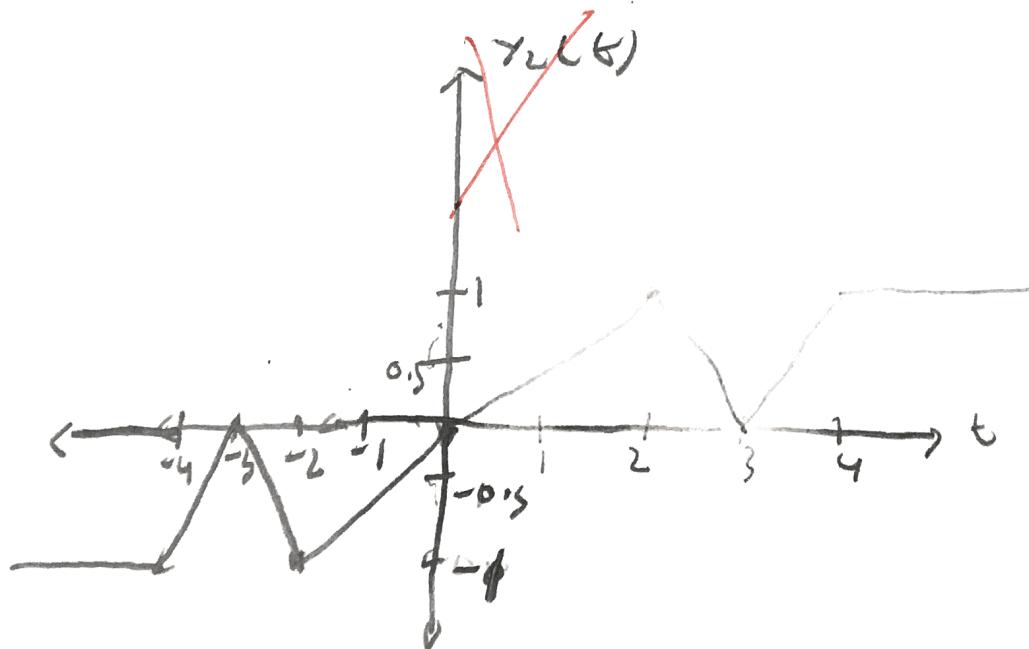
$$= \int_{-\infty}^{\infty} x_0(-\tau) h_0(t+\tau) d\tau$$

if $z = t$
 set $Tz - t$
 if $Pz = \tau$
 $dPz = dz$

$$= - \int_{-\infty}^{\infty} x_0(p) h_0(T-p) dp$$

$= -y_0(T) = -y_0(-t)$

$$\boxed{y_2(t) = -y_0(-t)}$$



15 Problem 5 (16 pts) Consider a cascade combination of two systems S_1 and S_2 : $x(t)$ is input to S_1 and $y(t)$ is the output, while the output of S_2 is $z(t)$.

$$x(t) \rightarrow [S_1] \rightarrow y(t) \rightarrow [S_2] \rightarrow z(t)$$

The IPOP relation for S_1 and S_2 is:

$$\begin{aligned} S_1 : y(t) &= e^{-t} x(t) u(t), \\ S_2 : z(t) &= \int_0^t e^{-(t-\sigma)} y(\sigma) u(\sigma) d\sigma. \end{aligned}$$

10 (a) (10 pts) Compute impulse response function $h_{12}(t, \tau)$ of the cascaded system $S_1 S_2$.

5 (b) (6 pts) Compute the output $z(t)$ if the input is
 $x(t) = e^{-3t} [u(t) - u(t-3)]$.

$$S_1: x(t) = \delta(t-\tau)$$

$$h_{12}(t, \tau) = h_2[h_1(t, \tau)]$$

Signal only matters
for $t \geq 0, \tau \geq 0$. $\therefore w.$

$$h_{12}(t, \tau) = \int_0^t e^{-(t-\sigma)} h_1(\sigma, \tau) u(\sigma) d\sigma$$

$$= \int_0^t e^{-(t-\sigma)} e^{-\sigma} \delta(\sigma-\tau) u(\sigma) d\sigma$$

$$= \int_0^t e^{-(t-\sigma)} e^{-\sigma} \delta(\sigma-\tau) u(\sigma) d\sigma$$

\downarrow
- consider $t \geq 0$ and $\tau \geq 0$

$$= e^{-t} u(\tau) \int_0^t \delta(\sigma-\tau) d\sigma = e^{-t} u(\tau) u(t-\tau)$$

\downarrow
0 if $t < \tau$
1 if $t > \tau$

$$h_{12}(t, \tau) = e^{-t} u(\tau) u(t-\tau)$$

for $t \geq 0$
 $\tau \geq 0$

$$b. Z(t) = \int_{-\infty}^{\infty} x(\tau) h_1(t, \tau) d\tau$$

$$= \int_{-\infty}^{\infty} e^{-3\tau} [u(\tau) - u(\tau-3)] e^{-t} u(\tau) u(t-\tau) d\tau$$

if
 $0 \leq \tau \leq 3$

$$= \int_0^3 e^{-3\tau} e^{-t} u(t-\tau) d\tau = INT$$

if $t < 0$

then

$$INT = 0$$

if $0 < t < 3$

$$\text{then } INT = \int_0^t e^{-3\tau} e^{-t} d\tau = e^{-t} \left[-\frac{e^{-3\tau}}{3} \right]_0^t$$

$$= e^{-t} \left[\frac{1}{3} - \frac{e^{-3t}}{3} \right]$$

if $t > 3$

$$\text{then } INT = \int_0^3 e^{-3\tau} e^{-t} d\tau = e^{-t} \left[-\frac{e^{-3\tau}}{3} \right]_0^3$$

$$= e^{-t} \left[\frac{1}{3} - \frac{e^{-9}}{3} \right]$$

$$Z(t) = \begin{cases} 0 & \text{if } t < 0 \\ 1 - \frac{e^{-3t}}{3} & \text{if } 0 \leq t \leq 3 \\ 1 - \frac{e^{-9}}{3} & \text{if } t > 3 \end{cases}$$

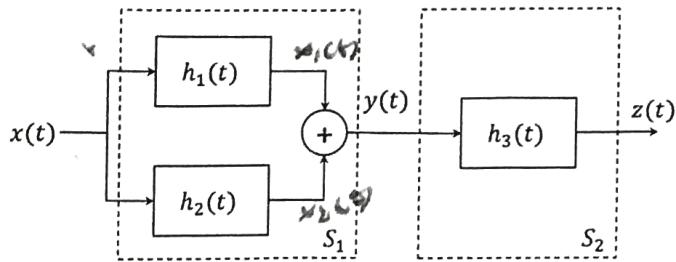
$$Z(t) = \boxed{\left[1 - \frac{e^{-3t}}{3} \right] [u(t) - u(t-3)] + \left[1 - \frac{e^{-9}}{3} \right] u(t-3)}$$

9 Problem 6 (16 pts)

Consider a cascaded LTI system S_1S_2 as follows

$$x(t) \rightarrow [S_1] \rightarrow y(t) \rightarrow [S_2] \rightarrow z(t)$$

The cascade system is shown below.



where $h_1(t) = \delta(t-1)$, $h_2(t) = \delta(t-2)$, and $h_3(t) = \delta(t-1) - \delta(t-2) + \delta(t-3)$. Let $x(t) = 2(u(t) - u(t-2))$, then

5 (a) (5 pts) Find the IPOP between $x(t)$ and $y(t)$. Plot $y(t)$.

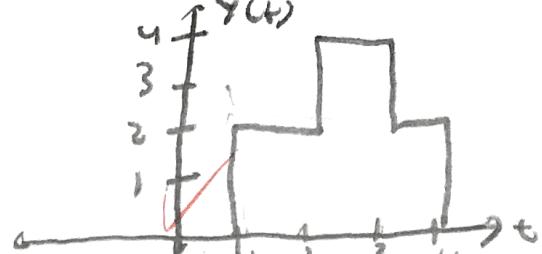
3 (b) (5 pts) Write the impulse response of the cascade system S_1S_2 .

1 (c) (6 pts) Compute and plot $z(t)$ for the specified input.

$$\begin{aligned} a. \quad x_1(t) &= \int_{-\infty}^{\infty} x(\tau) h_1(t-\tau) d\tau \\ &= \int_{-\infty}^{\infty} x(\tau) \delta(t-1-\tau) d\tau \quad \text{if } p = t-1 \\ &= \int_{-\infty}^{\infty} x(\tau) \delta(p-\tau) d\tau = x(p) \\ &= x(t-1) \end{aligned}$$

$$\begin{aligned} x_2(t) &= \int_{-\infty}^{\infty} x(\tau) h_2(t-\tau) d\tau \quad \text{if } q = t-2 \\ &= \int_{-\infty}^{\infty} x(\tau) \delta(t-2-\tau) d\tau \\ &= \int_{-\infty}^{\infty} x(\tau) \delta(p-\tau) d\tau = x(p) = x(t-2) \end{aligned}$$

$$\begin{aligned} &\approx \int_{-\infty}^{\infty} x(\tau) \delta(p-\tau) d\tau = x(p) = x(t-2) \\ Y(t) &= x(t-1) + x(t-2) \\ \text{Since } x(t) &= 2[u(t) - u(t-2)] \end{aligned}$$



$$\text{b. } h_{1S}(t) = h_1(t) + h_2(t)$$

$$= \delta(t-1) + \delta(t-2)$$

$$h_{2S}(t) = \delta(t-1) + \delta(t-2) + \delta(t-3)$$

$$h_{12}(t) = \int_{-\infty}^{\infty} h_{1S}(\tau) h_{2S}(t-\tau) d\tau$$

$$= \int_{-\infty}^{\infty} [\delta(\tau-1) + \delta(\tau-2)] [\delta(t-\tau-1) + \delta(t-\tau-2) + \delta(t-\tau-3)] d\tau$$

$$= \int_{-\infty}^{\infty} \delta(\tau-1)\delta(t-\tau-2) + \delta(\tau-1)\delta(t-\tau-3) + \delta(\tau-1)\delta(t-\tau-4) d\tau$$

$$+ \cancel{\delta(\tau-2)\delta(t-\tau-3)} + \cancel{\delta(\tau-2)\delta(t-\tau-4)} + \cancel{\delta(\tau-2)\delta(t-\tau-5)}$$

$$= \delta(t-2) + 2\delta(t-3) + 2\delta(t-4) + \delta(t-5)$$

$$\boxed{h_{12}(t) = \cancel{\delta(t-2) + 2\delta(t-3) + 2\delta(t-4)} + \delta(t-5)}$$

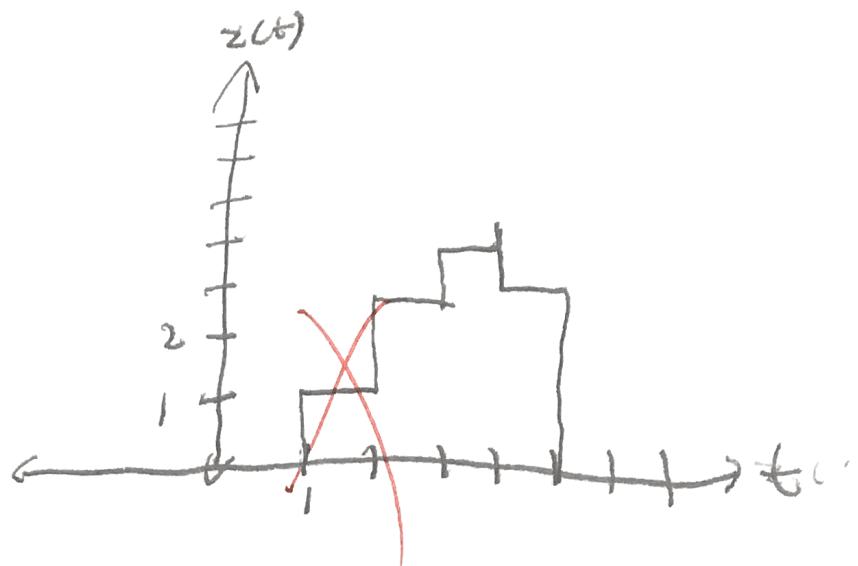
$$Z(k) = \int_{-\infty}^{\infty} x(\tau) h_{12}(t-\tau) d\tau$$

$$= \int_{-\infty}^{\infty} 2[u(\tau) - u(\tau-2)] h_{12}(t-\tau) d\tau = 2 \int_0^2 m_2(t-\tau) d\tau$$

$$= 2 \int_0^2 \delta(t-\tau-2) + \cancel{2\delta(t-\tau-3)} + 2\delta(t-\tau-4) + \delta(t-\tau-5) d\tau$$

$$0 \leq t-2 \leq 2$$

$$= \boxed{2[u(t-2) - u(t-4)] + 2[u(t-3) - u(t-5)] + 2[u(t-4) - u(t-6)] + u(t-5) - u(t-7)}$$



$$+\sum_{j=2}^L I_j + V_0 + \sum_{j=1}^L I_j^2 - I_{L+1}^2 = 0$$
$$\cancel{\left(-I_2 + \frac{1}{2} I_1 \right)} = -1$$

$$\left(2 + \frac{1}{2} I_1 \right) I_1 - I_2 - I_3 = -1$$

2. $I_2 \cdot \frac{1}{j_8} + (I_2 - I_3) \mapsto (I_2 - I_3) \cdot 1 \mapsto I_2 = 6$

$$-1 + I_1 + \left(2 + \frac{1}{j_8} \right) I_2 - I_3 = 1$$

$$I_3 - I_1 + I_3 - I_2 + 2j_8 I_3 = 0$$
$$-I_1 - I_2 + (2 + j_8) I_3 = 0$$