

EE 121B Mid-term Examination  
Winter 2017

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2) 50

3) 24

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Name

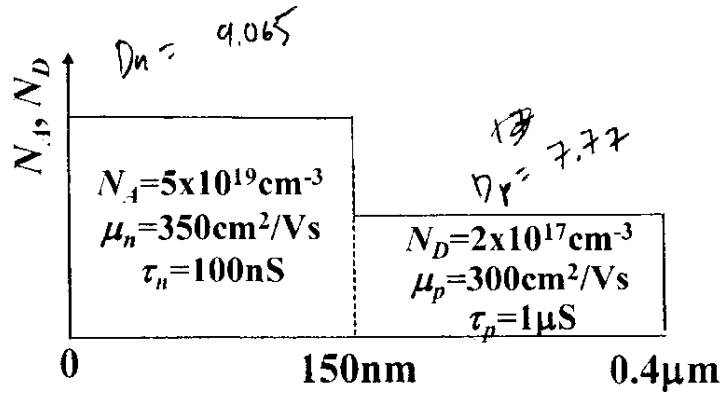

Open Book  
Assume  $T=300K$  and the substrate is silicon

$$n_i = 1.5 \times 10^{10} \text{ cm}^{-3}$$

1. (a) Consider a *pn* junction with uniformly doped *n* and *p* region as shown below. What are the built-in potential and the depletion width at 300K? (12 points)

Shut base  
 $\sqrt{D_n \tau_n} = 9.5 \times 10^{-4}$   
 $L_p \approx 7.7 \mu\text{m}$

$L_n = \sqrt{D_p \tau_p} \approx 7.7 \mu\text{m}$



$$V_{bi} = \frac{KT}{2} \ln \left( \frac{N_A N_D}{n_i^2} \right) = 0.0259 \text{ V} \ln \left( \frac{5 \times 10^{19} \text{ cm}^{-3} \cdot 2 \times 10^{17} \text{ cm}^{-3}}{1.5 \times 10^{10} \text{ cm}^{-3}} \right)$$

$$V_{bi} = 0.993 \text{ V}$$

$$W_D = \sqrt{\frac{2 \epsilon V_{bi}}{q \left( \frac{N_A N_D}{N_A + N_D} \right)}} = \sqrt{\frac{2 \cdot 11.7 \times 8.85 \times 10^{-14} \text{ F/cm} \cdot 0.993 \text{ V}}{1.6 \times 10^{-19} \text{ C} \left( \frac{5 \times 10^{19} \cdot 2 \times 10^{17}}{5 \times 10^{19} + 2 \times 10^{17}} \right)}} = 8.03 \times 10^{-6} \text{ cm}$$

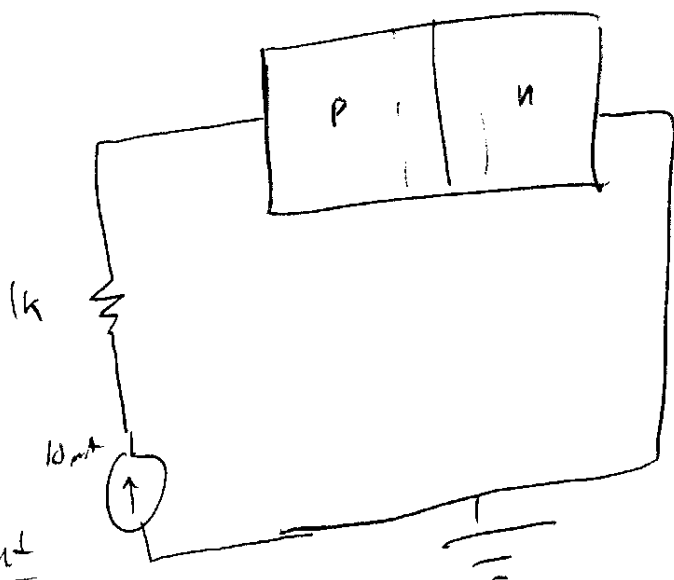
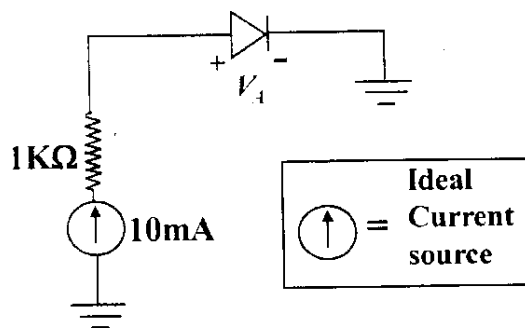
$$= 80.3 \text{ nm}$$

$$= W_D$$

$$W_{Dp} = W_D \left( \frac{N_D}{N_D + N_A} \right) = 3.2 \times 10^{-8} \text{ cm}$$

$$W_{Dn} = W_D \left( \frac{N_A}{N_D + N_A} \right) = 7.998 \times 10^{-6} \text{ cm}$$

(b) If the size (cross-section) of the *pn* diode,  $A=10^{-2}\text{cm}^2$ , and is put into the following circuit, what is  $V_A$  at 300K? (12 points)



- find  $V$  relation at

$$I = 10\text{mA}$$

$$W_A = 150\text{nm} - W_{DP} = \approx 150\text{nm}$$

$$W_{NB} = 250\text{nm} - W_{DN} = 250\text{nm} - 80\text{nm} \\ = 170\text{nm} \\ = 170 \times 10^{-7}\text{cm}$$

Start base current

$$I = A \left( \frac{q D_n n_i^2}{W_A N_A} + \frac{q D_p n_i^2}{W_{NB} N_D} \right) \left( e^{\frac{q V_A}{kT}} - 1 \right) = 10\text{mA}$$

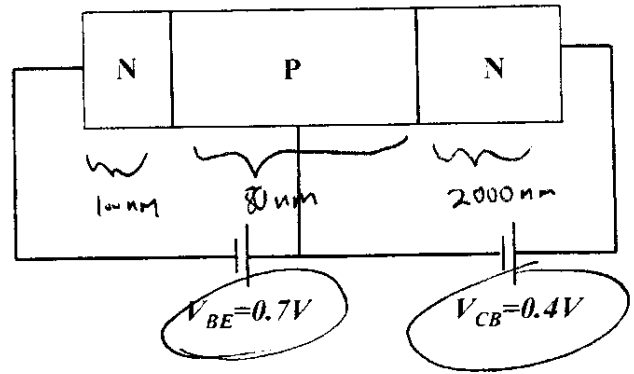
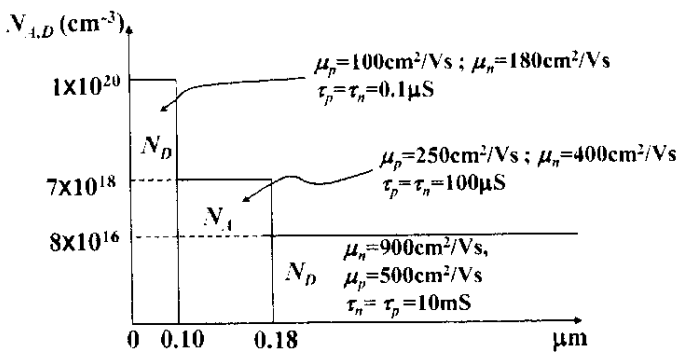
$$10^{-2}\text{A} \left( \frac{1.6 \times 10^{-19} \text{C} \cdot 9.07 \text{cm}^2/\text{s} \cdot (1.5 \times 10^{10} \text{cm}^{-3})^2}{150 \times 10^{-7} \text{cm} \cdot 5 \times 10^{19} \text{cm}^{-3}} + \frac{1.6 \times 10^{-19} \text{C} \cdot 7.77 \cdot (1.5 \times 10^{10})^2}{170 \cdot 10^{-7} \text{cm} \cdot 2 \times 10^{17} \text{cm}^{-3}} \right) \left( e^{\frac{V_A}{V_T}} - 1 \right) = 10\text{mA}$$

$$10^{-2} \left( 4.35 \times 10^{-13} + 8.23 \times 10^{-11} \right) \left( e^{\frac{V_A}{V_T}} - 1 \right) = 10\text{mA}$$

$$e^{\frac{V_A}{V_T}} - 1 = 1.209 \times 10^{10} \rightarrow e^{\frac{V_A}{V_T}} = 1.209 \times 10^{10}, \frac{V_A}{V_T} = 23.2$$

$$V_A = 0.601\text{V}$$

2) Consider the following bipolar transistor (the collector length is 2μm)



$$A = 100 \mu\text{m}^2$$

a) What is the neutral base width of the device with the DC biases as indicated? (12 points)

$$N_{DE} = 10^{20} \quad N_{AB} = 7 \times 10^{18} \quad N_{DC} = 8 \times 10^{16}$$

$$D_{pE} = \mu_p v_t = 2.59 \text{ cm}^2/\text{s} \quad D_{nB} = \mu_n v_t = 10.36 \text{ cm}^2/\text{s} \quad D_{pC} = \mu_p v_t = 12.95 \text{ cm}^2/\text{s}$$

$$L_{Ep} = \sqrt{D_{pE} \tau_{pE}} = \sqrt{2.59 \cdot 10^{-4}} = 5.09 \times 10^{-4} \text{ cm} = 5.09 \mu\text{m}$$

$$L_{Bn} = \sqrt{D_{nB} \tau_{nB}} = \sqrt{10.36 \cdot 10^{-4}} = 3.22 \times 10^{-2} \text{ cm} = 322 \mu\text{m}$$

$$V_{bibe} = V_t \ln \left( \frac{N_{DE} N_{AB}}{n_i^2} \right) = 1.103 \text{ V} \quad V_{bibe}$$

$$V_{bibe} = V_t \ln \left( \frac{N_{DC} N_{AB}}{n_i^2} \right) = 0.918 \text{ V} \quad V_{bibe}$$

$$W_{DBE} = \sqrt{\frac{2 \epsilon_s \epsilon_i}{q} \left( \frac{V_{bibe} - V_{BE}}{\frac{N_{DE} N_{AB}}{N_{DE} + N_{AB}}} \right)} = \sqrt{\frac{2 \epsilon_s \epsilon_i}{q} \left( \frac{1.103 - 0.7}{\dots} \right)} = 8.93 \text{ nm}$$

$$W_{DBEp} = W_{DBE} \left( \frac{N_{DE}}{N_{DE} + N_{AB}} \right) = 8.93 \times 10^{-7} \text{ cm} \left( \frac{10^{20}}{10^{20} + 7 \times 10^{18}} \right) = 8.35 \text{ nm}$$

$$W_{DBC} = \sqrt{\frac{2 \epsilon_s \epsilon_i (V_{bibe} + V_{BC})}{2 \left( \frac{N_{DC} N_{AB}}{N_{DC} + N_{AB}} \right)}} = \sqrt{\frac{2 \epsilon_s \epsilon_i (0.918 + 0.4)}{2 \left( \dots \right)}} = 147 \text{ nm}$$

$$W_{DBCp} = W_{DBC} \left( \frac{N_{DC}}{N_{DC} + N_{AB}} \right) = 147 \times 10^{-7} \text{ cm} \left( \frac{8 \times 10^{16}}{8 \times 10^{16} + 7 \times 10^{18}} \right) = 1.66 \text{ nm}$$

$$W_B = W_{BM} - W_{DBEp} - W_{DBCp} = 80 \text{ nm} - 8.35 \text{ nm} - 1.66 \text{ nm} = 70.0 \text{ nm} = W_B$$

$$D_{Bn} = 10^{-3} \text{ cm}^2$$

$$1 \text{ mm}^2 = 10^{-12} \text{ m}^2 =$$

$$1 \text{ cm}^2 = 10^{-8} \text{ m}^2 \quad A = 100 \text{ } \mu\text{m}^2 = \left( \frac{10^{-4} \text{ cm}}{\text{mm}} \right) \left( \frac{10^{-4} \text{ cm}}{\text{mm}} \right) = 10^{-10} \text{ cm}^2$$

b) What are the values of  $I_C$ ,  $I_E$ , and  $I_B$ ? (14 points)  $A = 10^{-6} \text{ cm}^2$

$$I_C \approx \frac{A q D_{Bn} n_i^2}{W_B N_{AB}} e \frac{V_{BE}}{V_T} = \frac{10^{-6} \text{ cm}^2 \cdot 1.6 \times 10^{-19} \text{ C} \cdot 1.236 \text{ cm}^2/\text{s} \cdot (1.5 \times 10^{10} \text{ cm}^{-3})^2}{70 \cdot 10^{-7} \text{ cm} \cdot 7 \times 10^{18} \text{ cm}^{-3}} e \frac{0.7}{0.0259}$$

$$= \frac{3.73 \times 10^{-4}}{4.9 \times 10^{13}} \text{ C}^{27.07} = 4.16 \times 10^{-6} \text{ A} = I_C$$

$$I_B = \frac{A q D_{EP} n_i^2}{W_E N_{DE}} e \frac{V_{BE}}{V_T} \quad ; \quad W_E = W_{EM} - W_{DBEn} = 100 \text{ nm} - (8.93 \text{ nm} - 8.35 \text{ nm}) = 99.4 \text{ nm}$$

$$= \frac{10^{-6} \text{ cm}^2 \cdot 1.6 \times 10^{-19} \text{ C} \cdot 2.59 \text{ cm}^2/\text{s} \cdot (1.5 \times 10^{10})^2}{99.4 \times 10^{-7} \text{ cm} \cdot 10^{20} \text{ cm}^{-3}} e \frac{0.7}{0.0259} = 5.13 \times 10^{-8} \text{ A} = I_B$$

$$I_E = I_C + I_B = 4.21 \times 10^{-6} \text{ A}$$

c) What are the common base current gain,  $\alpha_0$  and the common emitter current gain,  $\beta_0$  at these biases? (12 points)

$$\gamma \approx \frac{1}{1 + \frac{D_E W_B N_{AB}}{D_B W_E N_{DE}}} = \frac{1}{1 + \frac{2.59 \cdot 70 \times 10^{-7} \text{ cm} \cdot 4 \times 10^{18} \text{ cm}^{-3}}{10.36 \cdot 99.4 \times 10^{-7} \text{ cm} \cdot 10^{22} \text{ cm}^{-3}}} = 0.988$$

$$L_B = \sqrt{D_B \tau_B} = \sqrt{10.36 \text{ cm}^2/\text{s} \cdot 10^{-8} \text{ s}} = 3.22 \times 10^{-2} \text{ cm}$$

$$\alpha_T \approx 1 - \frac{W_B^2}{2L_B^2} = 1 - \frac{(70 \times 10^{-7} \text{ cm})^2}{2 \cdot (3.22 \times 10^{-2} \text{ cm})^2} = 0.999999976$$

$$\alpha_0 = \gamma \alpha_T M, M \approx 1 \rightarrow \alpha_0 = \gamma \alpha_T = 0.988$$

$$\beta = \frac{\alpha_0}{1 - \alpha_0} = 82.3 = \beta$$

(e) What are the Early Voltage and the base transit time of this transistor at these biases? (12 points)

$$V_a = \frac{q W_B N_A B}{C_{DBC}} = C_{DBC} = \frac{\epsilon}{W_{DBC}} = \frac{11.7 \cdot 8.85 \times 10^{-14} \text{ F/cm}}{1.7 \times 10^{-7} \text{ cm}} = 7.04 \times 10^{-8} \text{ F/cm}^2$$

$$V_a = \frac{1.6 \times 10^{-19} \text{ C} \cdot 70 \times 10^{-7} \text{ cm} \cdot 7 \times 10^{18} \text{ cm}^{-3}}{7.04 \times 10^{-8} \text{ F/cm}^2} = 111.4 \text{ V} = V_a$$

$$\tau_B = \frac{W_D^2}{2 D_n} = \frac{(70 \times 10^{-7} \text{ cm})^2}{2 \cdot 10.36 \text{ cm}^2/\text{s}} = 2.36 \times 10^{-12} \text{ s}$$

3 (a) For the transistor in (2) if we want to while maintaining the same current gain,  $\beta_0$ , but with the base width (metallurgical) reduced to  $0.06 \mu\text{m}$ , what is the new base doping concentration assuming the same minority carrier (i.e. electron) mobility? (same biases as in (2)) (14 points)

$$\beta = \frac{\alpha_0}{1 - \alpha_0} = 82.3 \rightarrow \alpha_0 = 0.988 \rightarrow \text{since } \alpha_T \approx 1, \gamma = 0.988$$

$$\gamma = 0.988 = \left( 1 - \frac{D_{EP} W_B' N_{AB}}{D_{BN} W_E N_{DE}} \right)^{-1}$$

$N_{AB}', W_B'$  : new base width & doping

same  $\rightarrow \frac{D_{EP} W_B N_{AB}}{D_{BN} W_E N_{DE}} = 0.01215$

same  $\rightarrow \frac{D_{BN} W_E N_{DE}}{D_{EP}}$

↑ same  
 $W_E$  will be approximately the same bc Base will still be doped much less

$$W_B N_{AB} = 0.01215 \cdot \frac{D_{BN} W_E N_{DE}}{D_{EP}} = \frac{4.83 \times 10^{13}}{\cancel{1}} \quad -2$$

The product of  $W_B$  at  $N_{AB}$  must maintain this value. If we approximate that  $W_B$  will shorten by as much as  $W_{BM}$ , then  $W_B' = 50 \times 10^{-7} \text{ cm}$ .

In this case,  $N_{AB}' = 4.83 \times 10^{13} / 50 \times 10^{-7} \text{ cm} = 9.66 \times 10^{18}$ , which is not that much greater than  $N_{AB}$ . Thus, we justify our approximation because now assuming  $N_{AB}'$  is valid, we look at depletion widths and they will not change very much.

So,  $\boxed{N_{AB}' \approx 9.66 \times 10^{18} \text{ cm}^{-3}}$



(b) What are the Early Voltage and the base transit time for the case in 3(a)?  
Discuss. (12 points)

$$V_A = \frac{2W_B N_A B}{C_{DBC}}$$

By definition  $W_B N_A B =$  same as for previous case.

$$4.9 \times 10^{13}$$

$$C_{DBC} = \frac{\epsilon}{W_{DBC}}$$

$$W_{DBC} =$$

$$\frac{2\epsilon (V_{bi, bc} - V_{bc})}{q \left( \frac{N_D N_A B}{N_{Dc} - N_{Ab}} \right)}$$

$$V_{bi, bc} - V_{bc} = 1.308V$$

$$= 1.471 \times 10^{-5} \text{ cm}$$

$$C_{DBC} = 7.04 \times 10^{-8}$$

$$V_A = \frac{2 \cdot 4.9 \times 10^{13}}{7.04 \times 10^{-8}}$$

$$111.4 V = V_A$$

$$T_B = \frac{W_B^2}{2D_{Bn}}$$

→ approximately  $W_B \approx 50 \times 10^{-7} \text{ cm}$  as per discussed

$$T_B = 1.21 \times 10^{-12} \text{ s}$$

Since  $V_A$  depends on  $W_B N_A B$ , that part will not change.  $C_{DBC}$  depends on the depletion width of BC junction, as per discussed, this does not change much. The result is that is very similar to previous  $V_A$ , which makes sense because current gain  $\beta$  remains constant.

$T_B$  becomes smaller because the base width shrinks, meaning carriers can travel across it much faster bc of shorter travel distance.