

# **EE 115A**

**Final Exam**

**Fall 2006**

**Your Name:**

Solutions

**Name of Person to Your Left:**

**Name of Person to Your Right:**

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**Time Limit: 3 Hours**

**Where applicable, place answers inside designated boxes.**

**Use all approximations specified in each problem.**

**1. 5**

**2. 5**

**3. 10**

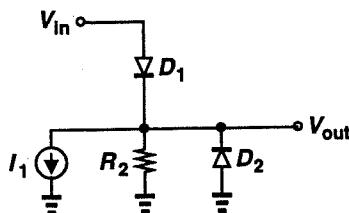
**4. 8**

**5. 12**

**6. 10**

**Total: 50**

1. Plot  $V_{out}$  and  $I_{D2}$  as a function of  $V_{in}$  as  $V_{in}$  goes from  $-\infty$  to  $+\infty$ . Determine the coordinates of each break point in the plots. Assume a constant-voltage diode model and  $I_1 R_2 > V_{D,ON}$ .



$$V_{in} \rightarrow -\infty$$

$D_1$  off

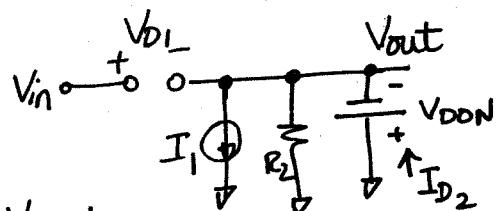
$$V_{out} = -V_{D,ON}$$

$D_2$  ON

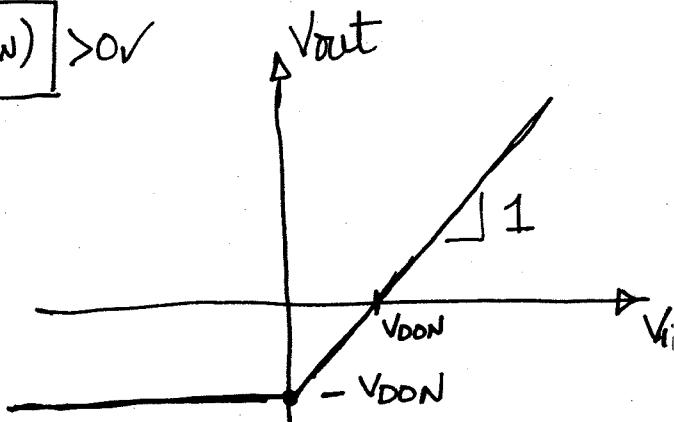
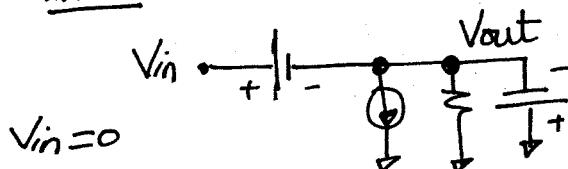
$$V_D < V_{D,ON} \rightarrow V_{in} - V_{out} < V_{D,ON}$$

$$\rightarrow V_{in} < 0 \quad (1)$$

$$I_1 = I_{D2} + \frac{V_{D,ON}}{R_2} \rightarrow I_{D2} = \frac{1}{R_2}(I_1 R_2 - V_{D,ON}) > 0$$

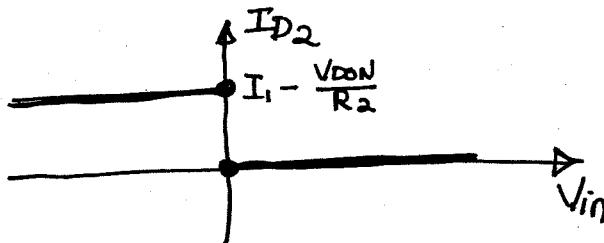
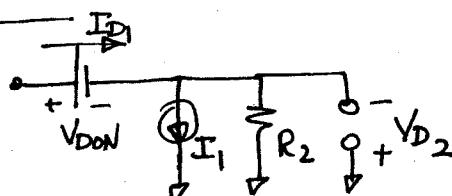


at  $V_{in}=0$   $D_1$  turns on:



$D_1 : ON$

$D_2 : OFF$



$$V_{out} = V_{in} - V_{D,ON}$$

$$I_{D2} = 0$$

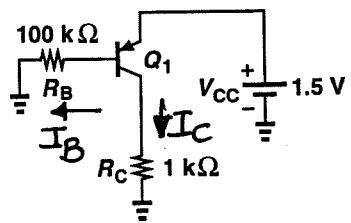
$$V_{D2} = -V_{out} < V_{D,ON} \rightarrow -V_{in} + V_{D,ON} < V_{D,ON}$$

$$\rightarrow V_{in} > 0$$

$$I_{D1} = I_1 + \frac{V_{out}}{R_2} > 0$$

$$I_{D1} = I_1 + \frac{V_{in}}{R_2} - \frac{V_{D,ON}}{R_2} > 0 \rightarrow V_{in} > \underbrace{(I_1 R_2 - V_{D,ON})}_{\text{negative}}$$

2. Determine the value of  $\beta$  that places  $Q_1$  at the edge of saturation. Assume  $V_A = \infty$  and  $I_S = 8 \times 10^{-16}$  A.



edge of saturation  $\rightarrow V_C = V_B$

$$V_B = 100K \cdot I_B$$

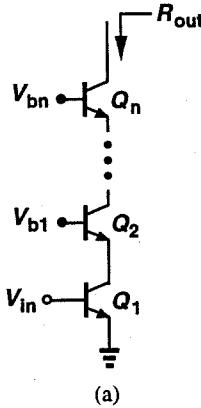
$$V_C = 1K \cdot I_C$$

$$I_C = \beta \cdot I_B$$

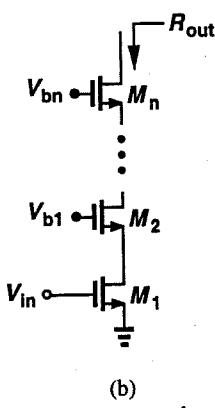
$$\Rightarrow 100K \frac{I_C}{\beta} = 1K I_C \Rightarrow \boxed{\beta = 100}$$

$$\beta = \boxed{100}$$

3. Compute the output impedance of each circuit as  $n \rightarrow \infty$ . Assume  $V_A < \infty$ ,  $\lambda > 0$ , and  $\beta < \infty$ .



(a)



(b)

a)  $R_n$ :  $R_{out}$  when we have 'n' transistors.

$$R_1 = r_o, R_2 = r_o (1 + g_m (r_{o\parallel} r_\pi)) + r_o \parallel r_\pi, \dots, R_n = r_o (1 + g_m (R_{n-1} \parallel r_\pi)) + R_{n-1} \parallel r_\pi$$

as  $n \rightarrow \infty$ ;  $R_n = R_{n-1}$  so

$$R_\infty = r_o (1 + g_m (R_\infty \parallel r_\pi)) + R_\infty \parallel r_\pi$$

$$\Rightarrow R_\infty = r_o \cdot \frac{1 + \beta + \sqrt{(1 + \beta)^2 + 4 \frac{r_\pi}{r_o}}}{2} \simeq r_o (1 + \beta)$$

b)  $R_1 = r_o$ ;  $R_2 = r_o (1 + g_m r_o) + r_o$ ; ...;  $R_n = r_o (1 + g_m R_{n-1}) + R_{n-1}$

$$R_n = r_o + (1 + g_m r_o) R_{n-1}$$

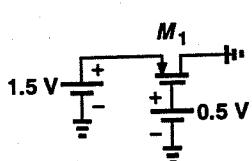
Since  $1 + g_m r_o > 1$ ;  $R_n$  keeps increasing as  $n \rightarrow \infty$

so  $R_\infty = \infty$

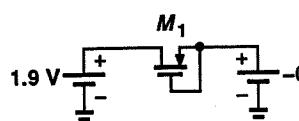
(a):  $R_{out} = \boxed{\simeq r_o (1 + \beta)}$

(b):  $R_{out} = \boxed{\infty}$

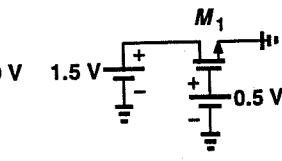
4. Determine the region of operation of each transistor (off, triode region, or saturation region). Assume a threshold voltage of 0.4 V for NMOS devices and -0.4 V for PMOS devices.



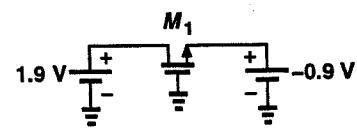
(a)



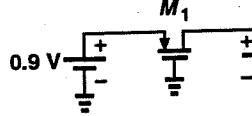
(b)



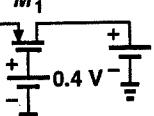
(c)



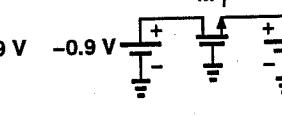
(d)



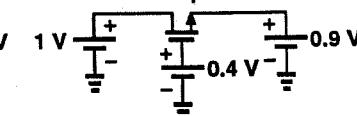
(e)



(f)



(g)



(h)

a) PMOS

$$V_{GD} = 0.5 > V_{THP}$$

$$V_{GS} = -1 < V_{THP}$$

SAT

b) PMOS

$$V_{GD} = 0 > V_{THP}$$

$$V_{GS} = -1 < V_{THP}$$

SAT

c) NMOS

$$V_{GD} = -1 < V_{THN}$$

$$V_{GS} = 0.5 > V_{THN}$$

SAT

d) NMOS

$$V_{GD} = -1.9 < V_{THN}$$

$$V_{GS} = 0.9 > V_{THN}$$

SAT

e) PMOS

$$V_{GS} = -0.9 < V_{THP}$$

$$V_{GD} = -0.4 = V_{THP}$$

Edge of Sat

f) PMOS

$$V_{GS} = -0.5 < V_{THP}$$

$$V_{GD} = -0.5 < V_{THP}$$

TRIODE

g) NMOS

$$V_{GD} = -0.4 < V_{THN}$$

$$V_{GS} = 0.9 > V_{THN}$$

SAT

h) NMOS

$$V_{GS} = -0.5 < V_{THN}$$

OFF

(a)  SAT

(b)  SAT

(c)  SAT

(d)  SAT

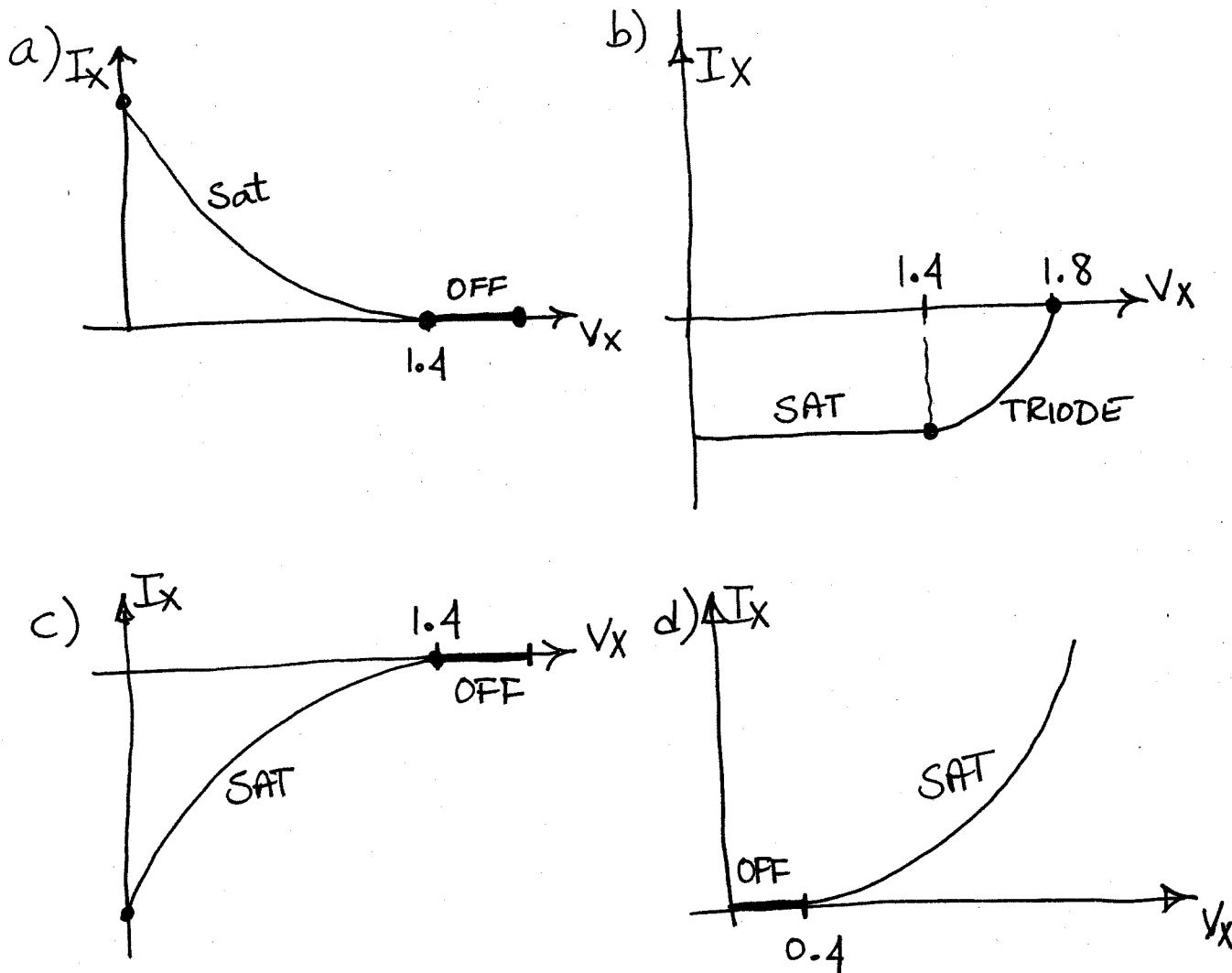
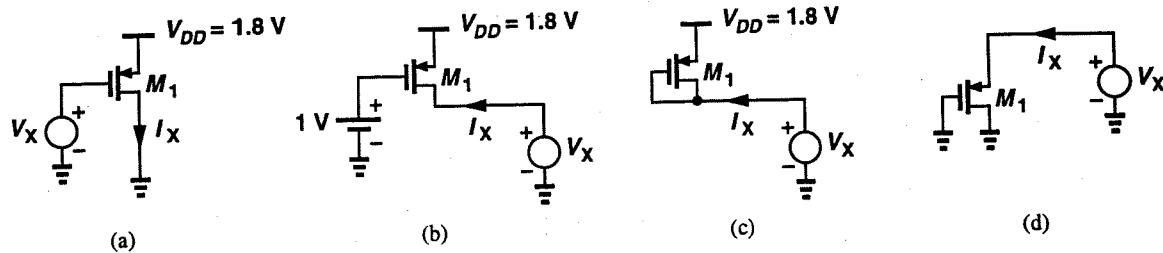
(e)  EDGE

(f)  TRIODE

(g)  SAT

(h)  OFF

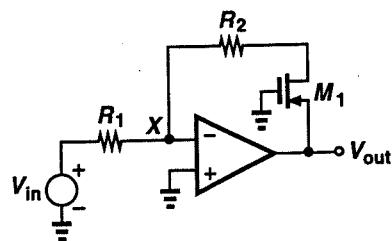
5. Plot  $I_X$  as a function of  $V_X$  as  $V_X$  varies from 0 to 1.8 V. Assume  $\lambda = 0$  and  $V_{THP} = -0.4$  V. Determine at what value of  $V_X$  the device changes its region of operation.



6. In the circuit shown below, assume the op amp is ideal and has an infinite gain. Also,  $\lambda = 0$ .

(a) Assuming  $R_2 = 0$ , determine  $V_{out}$  as a function of  $V_{in}$ . For what range of  $V_{in}$  does the circuit produce a meaningful output, i.e.,  $M_1$  remains on?

(b) Now assume  $R_2 > 0$  and calculate the value of  $V_{in}$  that places  $M_1$  at the edge of saturation.

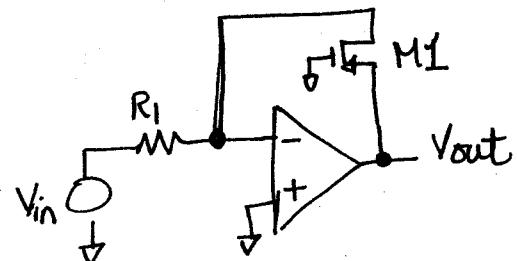


a)

$$\frac{1}{2} \mu_p C_{ox} \frac{w}{L} (V_{SG} - |V_{THP}|)^2 = - \frac{V_{in}}{R_1}$$

$$V_{SG} = V_{out}$$

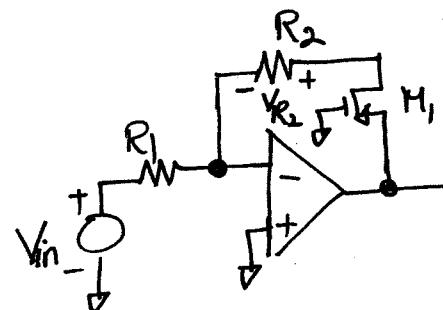
$$\Rightarrow V_{out} = \sqrt{\frac{-2V_{in}}{R_1 \mu_p C_{ox} \frac{w}{L}}} + |V_{THP}| \quad V_{in} \leq 0$$



b)

$$V_{R_2} = -\frac{V_{in}}{R_1} \cdot R_2$$

$$\text{EDGE OF SAT: } V_{R_2} = |V_{THP}|$$



$$\Rightarrow V_{in} = -\frac{R_1}{R_2} |V_{THP}|$$

$$(a) V_{out} = \sqrt{\frac{-2V_{in}}{R_1 \mu_p C_{ox} \frac{w}{L}}} + |V_{THP}| \quad V_{in} \leq 0$$

$$(b) V_{in} = -\frac{R_1}{R_2} |V_{THP}|$$