

EE 115A Analog Electronic Circuits

Winter 2013

Test 1

Closed Book

Name: Solution

Student ID No.: Solution

1) _____ / 10

2) _____ / 12

3) _____ / 10

4) _____ / 8

TOTAL _____ / **40**

10 pts

1) Provide a short answer to each of the following.

2 pts

a) Why/How is the depletion region formed in a PN junction when there is nothing connected to the terminals?

Electrons tend to diffuse from N-type to P-type while holes tend to diffuse from P-type to N-type. Diffusion leaves immobile atoms which forms depletion region. (1pt)
 Depletion region creates electrical field which opposes diffusion so that net current is zero at equilibrium. (1pt)

2 pts

b) Name 4 Operations that are possible with analog electronics but are not possible if you were constrained to the tools and circuit elements used in EE10 and EE110?

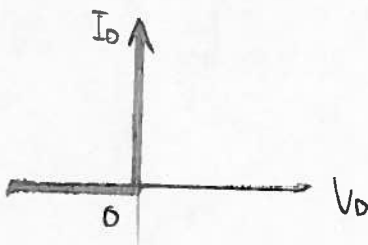
- (1) AC-DC converter (Voltage doubler, tripler...)
- (2) Rectifier (Full wave, half wave...)
- (3) Switch (Frequency translation, create digital logic...)
- (4) Data Converters (Analog to digital, digital to analog)

2 pts

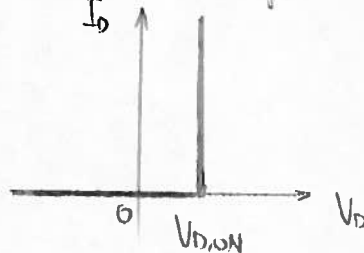
c) Name the three different diode models/equations. Why do we have three different diode models?



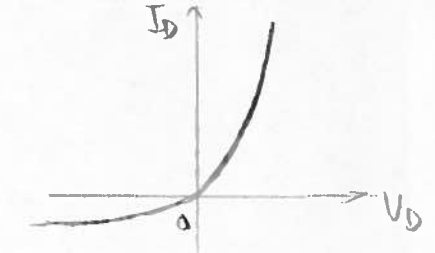
(1) Ideal diode



(2) Constant Voltage model



(3) Exponential model



$$I_D = I_S \left(e^{\frac{V_D}{V_T}} - 1 \right)$$

$$V_T = \frac{kT}{q}$$

(c) & (d) graded together, 4 pts in total

2 pts

d) When should we use each of the three different diode models?

Ideal diode model provides a quick understanding of circuit behavior.

Constant voltage model is a much better approximation to real model.

Exponential model is the most accurate, but takes much more effort to solve. It is only use when great accuracy is demanded.

2 pts

e) Name 4 of the factors that determine the amount of current flowing in a PN-junction

(1) Doping density

(2) cross section area of PN-junction

(3) voltage applied to PN-junction

(4) temperature, determines V_T

2) Find the value of the current flowing in the 1K resistance.

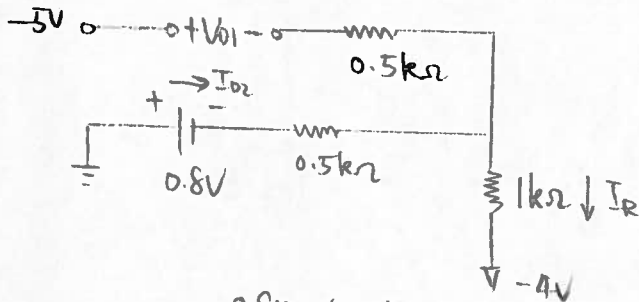
(a) Use the constant voltage drop model for the diodes, with $V_{D-on} = 0.8V$

(b) Use the real-diode equation with $I_s = 1.7 \times 10^{-17} A$. Feel free to make "engineering" assumptions to solve the problem. Be sure to state all your assumptions.

12 pts

(a) Assume D_1 OFF, D_2 ON

Equivalent circuit is shown below

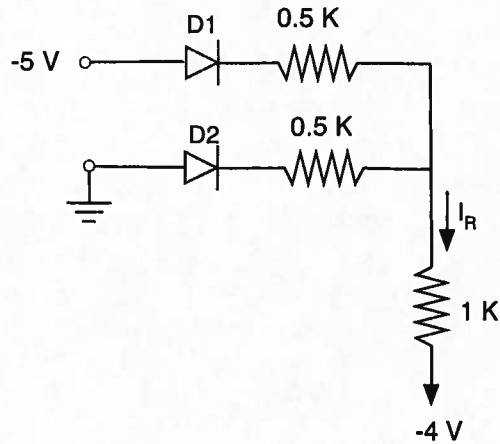


$$I_{D2} = \frac{0.8V - (-4V)}{0.5k\Omega + 1k\Omega} = 2.13mA > 0$$

$$V_{b1} = -5V - (-4V + 2.13mA \times 1k\Omega) = -3.13V < 0.8V$$

\Rightarrow Assumption is correct
 D_1 OFF, D_2 ON

Thus, $I_R = I_{D2} = 2.13mA$

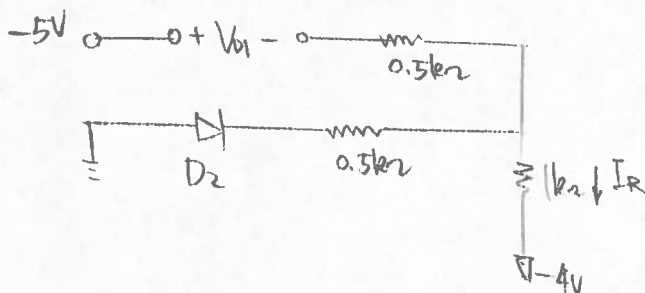


(b) From (a), we know that V_{b1} is very negative, so

$$I_{D1} = I_s (e^{\frac{V_{b1}}{V_T}} - 1) \approx -I_s \approx 0 \text{ with } V_{b1} \ll 0 \text{ and } I_{s1} = 1.7 \times 10^{-17} A$$

With $I_{D1} \approx 0$, D_1 can be regarded as open circuit. This assumption only creates ignorable error with $I_{D2} \gg I_s$.

Equivalent circuit is shown below



$$I_s (e^{\frac{V_{D2}}{V_T}} - 1) = \frac{0 - V_{D2} - (-4V)}{1.5k\Omega}$$

$$V_T = 26mV @ 300K, \text{ room temperature}$$

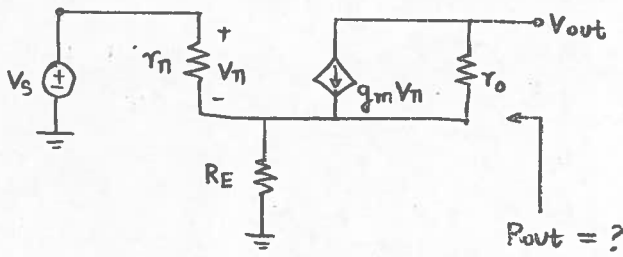
$$\Rightarrow I_R = I_s (e^{\frac{V_{D2}}{V_T}} - 1) = 2.105mA$$

$$\Rightarrow V_{b1} = -5V - (-4V + 2.105mA \times 1k\Omega) = -3.1V$$

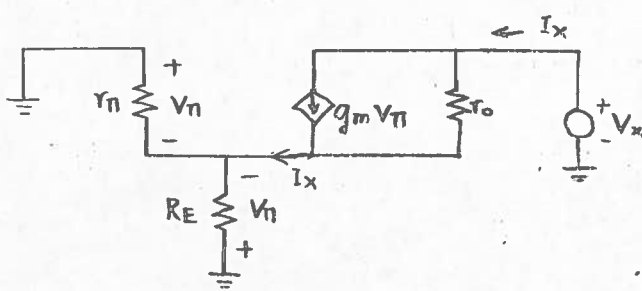
It further verifies that our assumption that $I_{D1} \approx -I_{s1} = -1.7 \times 10^{-17} A \ll I_{D2}$

So $I_R = I_{D2} = 2.105mA$

(3) Excluding R_c ,



To calculate R_{out} , we suppress all independent sources



$$R_{out} = V_x / I_x$$

$$I_x = \frac{-V_{\pi}}{(R_E \parallel r_{\pi})}$$

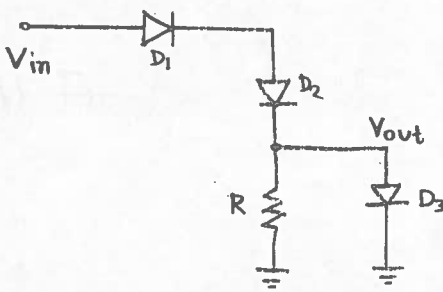
$$V_x = V_{r_o} - V_{\pi}$$

$$\therefore V_x = (I_x - g_m V_{\pi}) r_o - V_{\pi}$$

$$\therefore V_x = I_x \{ 1 + g_m (R_E \parallel r_{\pi}) \} r_o + (R_E \parallel r_{\pi}) I_x$$

$$R_{out} = \frac{V_x}{I_x} = (R_E \parallel r_{\pi}) + r_o + g_m r_o (R_E \parallel r_{\pi})$$

(4)



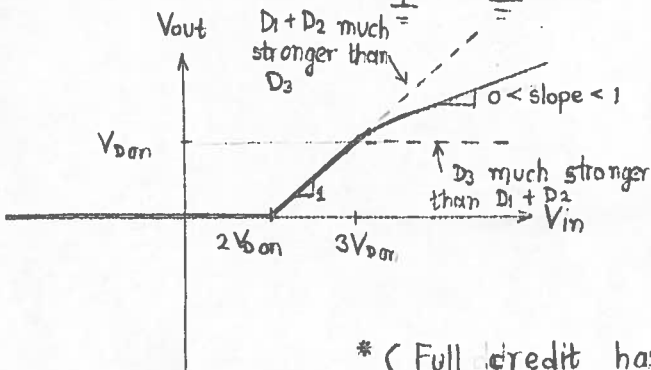
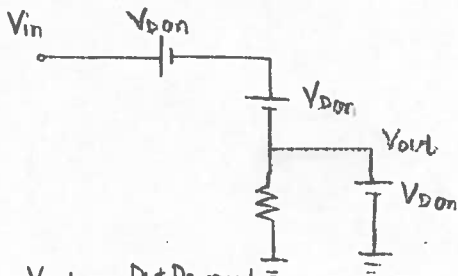
Plot. V_{out} vs V_{in} .

For $V_{in} < 2V_{D_{on}}$, D_1 & D_2 off $\Rightarrow V_{out} = 0 \Rightarrow D_3$ off

For $2V_{D_{on}} \leq V_{in} < 3V_{D_{on}}$, D_1 & D_2 on
 $\Rightarrow V_{out} = V_{in} - 2V_{D_{on}}$

But when $V_{out} \geq V_{D_{on}}$, D_3 turns on
 & now we have following situation

At $V_{out} \geq V_{D_{on}}$, i.e. $V_{in} \geq 3V_{D_{on}}$



$V_{out} \stackrel{?}{=} V_{in} - 2V_{D_{on}} \stackrel{?}{=} V_{D_{on}}$ which is not possible. Hence constant voltage model fails. So, at this point we can make following argument. If D_3 was absent, $V_{out} = V_{in} - 2V_{D_{on}}$. If $D_1 + D_2$ were absent, $V_{out} = V_{D_{on}}$. But since all 3 are present V_{out} will be some intermediate voltage as shown (a first order approximation) & slope of V_{out} will depend on relative strength of D_3 & $D_1 + D_2$.

* (Full credit has been given if you got either of the dotted lines)