

Problem 1 (24 points): Consider the four signals:

$$x_1[n] = \delta[n] + \delta[n-1]$$

$$x_2[n] = \delta[n] - \delta[n-1]$$

$$x_3[n] = \delta[n-2] + \delta[n-3]$$

$$x_4[n] = \delta[n-2] - \delta[n-3]$$

(a) (8 points) Express the signal $x[n] = \{1, 3, -1, 0\}$ as a function of the signals $x_i[n]$.

$$x[n] = \delta[n] + 3\delta[n-1] + \delta[n-2]$$

Since $\delta[n] = \frac{x_1[n] + x_2[n]}{2}$,

$$\delta[n-1] = \frac{x_1[n-1] - x_2[n-1]}{2}$$

$$\delta[n-2] = \frac{x_3[n-2] + x_4[n-2]}{2}$$

$$\Rightarrow x[n] = \frac{x_1[n] + x_2[n]}{2} + \frac{3}{2}(x_1[n-1] - x_2[n-1]) + \frac{x_3[n-2] + x_4[n-2]}{2}$$

$$= \frac{1}{2}(4x_1[n] - 2x_2[n] - x_3[n-2] - x_4[n-2])$$

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- (b) (8 points) Consider again the signals $x_i[n]$ $i = 1, 2, 3, 4$ and $x[n]$ from part (a) of this problem. You are given that the convolution of the signals $x_1[n]$ with $y[n]$ gives the signal $z_1[n]$, that is,

$$z_1[n] = x_1[n] * y[n]$$

You are also given that

$$z_2[n] = x_4[n] * y[n]$$

Calculate the convolution $x[n] * y[n]$ as a function of $z_1[n]$ and $z_2[n]$.

~~$$\begin{aligned} z_1[n] &= x_1[n] * y[n] \\ &= (\delta[n] + \delta[n-1]) * y[n] = y[0] + y[n-1] \end{aligned}$$~~

~~$$\begin{aligned} z_2[n] &= x_4[n] * y[n] \\ &= (\delta[n-2] - \delta[n-3]) * y[n] = y[n-2] - y[n-3] \end{aligned}$$~~

~~$$\begin{aligned} \text{Therefore, } x[n] * y[n] &= (\delta[n] + 3\delta[n-1] - \delta[n-2]) * y[n] \\ &= y[0] + 3y[n-1] - y[n-2] \end{aligned}$$~~

Since $z_1[n] = x_1[n] * y[n]$, and $x_3[n] = x_1[n-2]$

$$\Rightarrow x_3[n] * y[n] = z_1[n-2]$$

Since $z_2[n] = x_4[n] * y[n]$, and $x_2[n] = x_4[n+2]$

$$\Rightarrow x_2[n] * y[n] = z_2[n+2]$$

$$\begin{aligned} \text{Therefore, } x[n] * y[n] &= (2x_1[n] - x_2[n] - \frac{1}{2}x_3[n] - \frac{1}{2}x_4[n]) * y[n] \\ &= \underline{2z_1[n] - z_2[n+2] - \frac{1}{2}z_1[n-2] - \frac{1}{2}z_2[n]} \end{aligned}$$

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- (c) (8 points) The periodic extension of a signal $x[n]$ that has length N , is defined to be the signal $x_p[n]$ that simply repeats $x[n]$ every N samples, that is,

$$x_p[n] = \sum_{k=-\infty}^{\infty} x[n - kN]$$

- (i) (4 points) Consider the periodic extension of the signals $x_1[n]$ and $x_2[n]$ that are given in part (a) of this problem. Are any of these signals even for $N = 2$? If yes which ones?

$$x_1[n] = \delta[n] + \delta[n-1]$$

$$x_2[n] = \delta[n] - \delta[n-1]$$

Both signal x_1 and x_2 have length of 2. ($n=0$ and $n=1$)

After periodic extension,

$$x_{p1}[n] = \sum_{k=-\infty}^{\infty} (\delta[n-2k] + \delta[n-2k-1])$$

meaning $x_{p1}[n] = 1$ for all $n \in \mathbb{Z}$.

$$x_{p2}[n] = \sum_{k=-\infty}^{\infty} (\delta[n-2k] - \delta[n-2k-1])$$

meaning $x_{p2}[n] = \begin{cases} 1 & \text{for } n \text{ is even} \\ -1 & \text{for } n \text{ is odd} \end{cases}$.

Since $x_{p1}[-n] = x_{p1}[n] = 1$, $x_{p1}[n]$ is even;

$$x_{p2}[-n] = x_{p2}[n] = \begin{cases} 1 & \text{for } n \text{ is even} \\ -1 & \text{for } n \text{ is odd} \end{cases}$$

Therefore, both x_{p1} and x_{p2} are even.

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(ii) (4 points) Now consider the periodic extension of the signals $x_3[n]$ and $x_4[n]$ that are given in part (a) of this problem. Are any of these signals even for $N = 4$? If yes which ones?

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$$x_3[n] = \delta[n-2] + \delta[n-3]$$

$$x_4[n] = \delta[n-2] - \delta[n-3]$$

$$\Rightarrow X_{p3}[n] = \sum_{k=-\infty}^{\infty} (\delta[n-2-4k] + \delta[n-3-4k])$$

$$X_{p4}[n] = \sum_{k=-\infty}^{\infty} (\delta[n-2-4k] - \delta[n-3-4k])$$

Since $X_{p3}[-1] = 1 \neq X_{p3}[1] = 0$,

$X_{p3}[n]$ is not even

Since $X_{p4}[-1] = -1 \neq X_{p4}[1] = 0$,

$X_{p4}[n]$ is not even.

→ Counter examples

Problem 2 (40 points) : The following questions are not related to each other.

(a) (8 points) A one-dimensional linear classifier takes as input a feature value x and outputs 1 if x is larger than a constant A and zero otherwise.

(i) (4 points) Is this system linear?

$$y[n] = \begin{cases} 1, & \text{if } x[n] > A \\ 0, & \text{if } x[n] \leq A \end{cases}$$

Consider $x_1[n] < A$, $x_2[n] < A$, but $x_1[n] + x_2[n] > A$:

$$\Rightarrow \text{Sys}\{x_1[n]\} + \text{Sys}\{x_2[n]\} = 0 + 0 = 0$$

$$\text{Sys}\{x_1[n] + x_2[n]\} = 1 \neq 0$$

Therefore, system is not linear.

Counter example.

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(ii) (4 points) Is the system time invariant?

$$y[n] = \begin{cases} 1, & \text{if } x[n] > A \\ 0, & \text{if } x[n] \leq A \end{cases}$$

Delayed input:

$$\text{Sys} \{x[n-m]\} = \begin{cases} 1, & \text{if } x[n-m] > A = y[n-m], \\ 0, & \text{if } x[n-m] \leq A \end{cases}$$

Therefore, the system is T.I.

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(b) (12 points) Assume that the system $y[n] = f(x[n])$, where $f()$ is some unknown function, is Linear Time Invariant (LTI). Can you determine if the following systems are LTI? Briefly explain why.

(i) (4 points) $z[n] = y[n-5] + y[n-2]$

$$\begin{aligned} z[n] &= y[n-5] + y[n-2] \\ &= f(x[n-5]) + f(x[n-2]) \end{aligned}$$

Since it is a linear combination of two LTI systems, $z[n]$ is also LTI.

Specifically,

$$\text{Linearity: } \text{sys} \{ \alpha y_1[n] + b y_2[n] \}$$

$$= a \text{sys} \{ y_1[n] \} + b \text{sys} \{ y_2[n] \}$$

$$\text{TI: delayed input: } y[n-5-m] + y[n-2-m]$$

$$\text{delayed output: } z[n-m] = y[n-5-m] + y[n-2-m]$$

\Rightarrow LTI

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(ii) (4 points) $z[n] = y^2[n]$

$$z[n] = y^2[n] = f^2(x[n])$$

① Linearity:

$$\begin{aligned} \textcircled{1} \text{ Sys } \{ax_1[n] + bx_2[n]\} &= f^2(ax_1[n] + bx_2[n]) \\ &= (ay_1[n] + by_2[n])^2 \end{aligned}$$

$$\textcircled{2} \text{ Sys } \{ax_1[n]\} + \text{Sys } \{bx_2[n]\} = (ay_1[n])^2 + (by_2[n])^2$$

Since $\textcircled{1} \neq \textcircled{2} \Rightarrow z[n]$ is not linear

② TI:

$$\text{delayed input: } y^2[n-m] = f^2(x[n-m])$$

$$\text{delayed output: } z[n-m] = y^2[n-m] = f^2(x[n-m])$$

Therefore, $z[n]$ is TI.

Overall, $z[n]$ is not LTI.

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(iii) (4 points) $z[n] = y[2n]$

$$z[n] = y[2n] = f(x[2n])$$

① Linearity:

$$\begin{aligned} \text{Sys}\{ax_1[n] + bx_2[n]\} &= f(ax_1[2n] + bx_2[2n]) = ay_1[2n] + by_2[2n] \\ &= \text{Sys}\{ax_1[n]\} + \text{Sys}\{bx_2[n]\} = ay_1[2n] + by_2[2n] \end{aligned}$$

$\Rightarrow z[n]$ system is linear;

② T.I.:

delayed input: $z[n] = y[2n-m]$

delayed output: $z[n-m] = y[2(n-m)] = y[2n-2m] \neq y[2n-m]$

Therefore, system $z[n]$ is not T.I.

Overall, $z[n]$ system not LTI.

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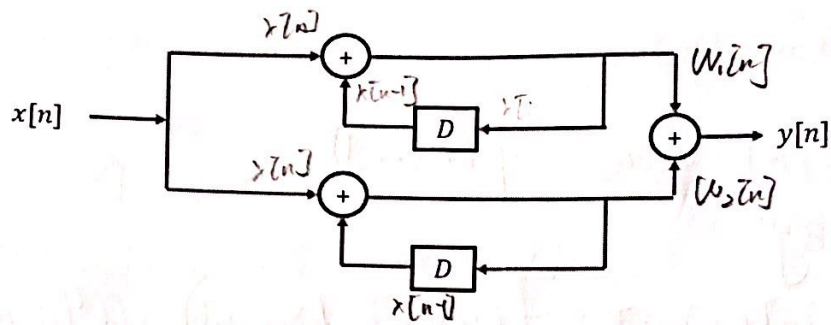


Figure 1: System for Problem 2(c)

(c) (20 points) Consider a relaxed system as shown in Fig. 1

(i) (8 points) Can you write the input-output equations for this system?

$$\begin{aligned}
 y[n] &= w_1[n] + w_2[n] \\
 &= (x[n] + x[n-1]) + (x[n] + x[n-1]) \\
 &= 2(x[n] + x[n-1]) \quad y[n] = 2x[n] + y[n-1]
 \end{aligned}$$

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(ii) (4 points) Is this a BIBO stable system? (explain why?)

Since system relaxed:

$$x[-\infty] = 0.$$

Therefore, for both $w_1[n]$ and $w_2[n]$, they act like accumulators:

$$w_1[n] = w_2[n] = \sum_{k=-\infty}^{\infty} x[n-k]$$

While $x[n]$ is bounded, k is unbounded
 \Rightarrow both $w_1[n]$ and $w_2[n]$ are unbounded.

Therefore, $y[n] = w_1[n] + w_2[n]$
is not bounded.

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(iii) (8 points) Assume you connect two of these relaxed systems, shown in Fig. 1, in series, what is the impulse response of the overall equivalent system?

$$y[n] = 2(x[n] + x[n-1])$$

$$\Rightarrow h[n] = 2(\delta[n] + \delta[n-1])$$

$$\Rightarrow \text{eq. 1} \quad h[n] * h[n]$$

$$= (2\delta[n] + 2\delta[n-1]) * (2\delta[n] + 2\delta[n-1])$$

$$= 4\delta[n] + 4\delta[n-1] + 4\delta[n-1] + 4\delta[n-2]$$

$$= 4\delta[n] + 8\delta[n-1] + 4\delta[n-2]$$

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Problem 3 (36 points) : The following questions are not related to each other.

- (a) (14 points) True or false: Let $x[n]$ be a real periodic signal with even period N , and c_k be the coefficients of its associated discrete-time Fourier series. Then, there always exist j_1, j_2 such that c_{j_1} and c_{j_2} are purely real. Mathematically justify your answer.

$x[n] : \text{real} \Rightarrow C_k$ is conjugate symmetric

$$\Rightarrow x^*[n] = x[n], C_k^* = C_{-k} = C_{N-k}$$

where $C_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi k}{N} n}$

Consider $k=0$:

$$C_0 = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^0 = \frac{1}{N} \sum_{n=0}^{N-1} x[n]$$

Since $x[n]$ is real, C_0 is real.

Since $C_k^* = C_{-k} = C_{N-k}$.

We always have C_0 being real.

with $a \in \mathbb{Z}$.
Therefore, it is true.

C_{j_1} and C_{j_2} are purely real

when j_1 and j_2 are multiples of N .

j_1 & j_2 ~~should~~ ^{can} be in the same period & still C_{j_1} & C_{j_2} will be real.

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(b) (8 points) Find the discrete-time Fourier Series coefficients for the following signal:

$$x[n] = \sin\left(\frac{2\pi n}{3}\right) \cos\left(\frac{\pi n}{3}\right)$$

$$X[n] = \sin\left(\frac{2\pi n}{3}\right) \cos\left(\frac{\pi n}{3}\right)$$

$$= \frac{1}{4} \left(e^{j\frac{2\pi}{3}n} - e^{-j\frac{2\pi}{3}n} \right) \left(e^{j\frac{\pi}{3}n} + e^{-j\frac{\pi}{3}n} \right)$$

$$= \frac{1}{4} \left(e^{j\pi n} + e^{j\frac{\pi}{3}n} - e^{-j\frac{\pi}{3}n} - e^{-j\pi n} \right)$$

Since $\sin\left(\frac{2\pi}{3}n\right)$ has period $N_1 = 3$,

$\cos\left(\frac{\pi n}{3}\right)$ has period $N_2 = 6$

$\Rightarrow X[n]$ has period of 6.

$$\text{Since } X[n] = \sum_{k=0}^{5} C_k e^{j\frac{\pi}{3}kn}$$

$$\Rightarrow C_0 = 0$$

$$C_2 = 0$$

$$C_4 = 0$$

$$C_1 = \frac{1}{4}$$

$$C_3 = 0$$

$$C_5 = C_{-1} = -\frac{1}{4}$$

- (c) (14 points) Assume that for a periodic signal $x[n]$ with period N you find the DTFS coefficients c_k . As we discussed in class, the Fourier Series coefficients c_k can also be thought as the values of a periodic signal with the same period N . Let's call this periodic signal $c[k]$, that is,

$$c[k] = c_k$$

Your friend claims that if you take the Fourier Series expansion of this periodic signal $c[k]$, you will get coefficients d_m that will be sufficient to retrieve the original signal $x[m]$. Are they right? Mathematically justify your answer.

$$C_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j \frac{2\pi k}{N} n}$$

$$\Rightarrow d_m = \frac{1}{N} \sum_{k=0}^{N-1} c[k] e^{-j \frac{2\pi m}{N} k} = d[m]$$

$$\text{Since } x[n] = \sum_{k=0}^{N-1} C_k e^{j \frac{2\pi k}{N} n}$$

$$N d[m] = \sum_{k=0}^{N-1} c[k] e^{-j \frac{2\pi k}{N} m}$$

We can see that $x[n] = N d[-m]$.

Therefore, $x[m]$ has values of $N d[-m]$.

The friend is right.

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