

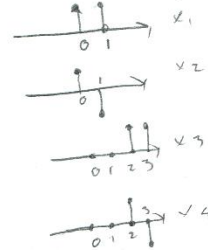
Problem 1 (24 points): Consider the four signals:

$$x_1[n] = \delta[n] + \delta[n-1]$$

$$x_2[n] = \delta[n] - \delta[n-1]$$

$$x_3[n] = \delta[n-2] + \delta[n-3]$$

$$x_4[n] = \delta[n-2] - \delta[n-3]$$



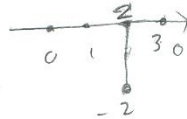
(a) (8 points) Express the signal  $x[n] = \begin{cases} 1, & n=0 \\ 3, & n=1 \\ -1, & n=2 \\ 0, & n=3 \end{cases}$  as a function of the signals  $x_i[n]$ .

$$x_1[n] - x_2[n] + x_3[n] - x_4[n] \rightarrow$$

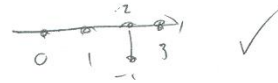
$$\Rightarrow 2x_1[n] - x_2[n]$$



$$-x_3[n] - x_4[n] \rightarrow$$



$$\rightarrow -\frac{1}{2}(x_3[n] + x_4[n]) \rightarrow$$



Thus  $x[n] = 2x_1[n] - x_2[n] - \frac{1}{2}(x_3[n] + x_4[n])$

8

$$x[n] = \{1, 3, -1, 0\}$$

(b) (8 points) Consider again the signals  $x_i[n]$   $i = 1, 2, 3, 4$  and  $x[n]$  from part (a) of this problem. You are given that the convolution of the signals  $x_1[n]$  with  $y[n]$  gives the signal  $z_1[n]$ , that is,

$$z_1[n] = x_1[n] * y[n] = \overbrace{\delta[n] * y[n]}^{y[n] + y[n-1]} + \delta[n-2] * y[n]$$

You are also given that

$$z_2[n] = x_2[n] * y[n] = \overbrace{y[n-2] - y[n-3]}^{\delta[n-2] * y[n] - \delta[n-3] * y[n]}$$

Calculate the convolution  $x[n] * y[n]$  as a function of  $z_1[n]$  and  $z_2[n]$ .

$$\begin{aligned} x[n] * y[n] &= (x_1 - x_2 - \frac{1}{2}(x_3 + x_4)) * y[n] \\ &= (2(\delta[n] + \delta[n-1]) - (\delta[n] - \delta[n-1]) - \frac{1}{2}(2\delta[n-2])) * y[n] \\ &= (2\delta[n] + 2\delta[n-1] - \delta[n] + \delta[n-1] - \delta[n-2]) * y[n] \\ &= (\delta[n] + 3\delta[n-1] - \delta[n-2]) * y[n] \\ &= y[n] + 3y[n-1] - y[n-2] \\ &= z_1[n](1 + 3\delta[n-1]) + z_2[n](-\delta[n-2]) \end{aligned}$$

$$\begin{aligned} z_1[n] &= y[n] + y[n-1] \\ z_2[n] &= y[n-2] - y[n-3] \end{aligned}$$

⚡

- (c) (8 points) The periodic extension of a signal  $x[n]$  that has length  $N$ , is defined to be the signal  $x_p[n]$  that simply repeats  $x[n]$  every  $N$  samples, that is,

$$x_p[n] = \sum_{k=-\infty}^{\infty} x[n - kN]$$

- (i) (4 points) Consider the periodic extension of the signals  $x_1[n]$  and  $x_2[n]$  that are given in part (a) of this problem. Are any of these signals even for  $N = 2$ ? If yes which ones?



$x_{1p}[-n] = x_{1p}[n] \rightarrow$  even for all  $n \neq 0$   
 but @  $n=0, x_{1p}[-0] \neq -x_{1p}[0] \neq 0$   
not even

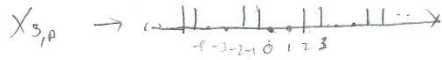


$x_{2p}[0] \neq -x_{2p}[-0],$  not even

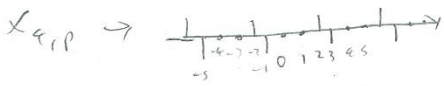
neither even

(ii) (4 points) Now consider the periodic extension of the signals  $x_3[n]$  and  $x_4[n]$  that are given in part (a) of this problem. Are any of these signals even for  $N = 4$ ? If yes which ones?

4



$x[1] \neq x[-1]$ , not even



$x[1] \neq x[-1]$ , not even

Neither even

**Problem 2 (40 points) :** The following questions are not related to each other.

(a) (8 points) A one-dimensional linear classifier takes as input a feature value  $x$  and outputs 1 if  $x$  is larger than a constant  $A$  and zero otherwise.

(i) (4 points) Is this system linear?

$$y = \begin{cases} x & \text{for } x > A \\ 0 & \text{otherwise} \end{cases} \rightarrow y[n] = u[x-A] \quad \text{where } u[0] = 0$$

$$\text{Let } x_1 = \left(\frac{2}{3}A\right) \text{ and } x_2 = \left(\frac{2}{3}A + 5\right)$$

such that  $y_1 = 0, y_2 = 0$

$$y[n] = u[x[n]-A]$$

$$\rightarrow \text{Let } x_3 = \alpha x_1 + \beta x_2 = \frac{2}{3}\alpha A + \frac{2}{3}\beta A + 5$$

where  $\alpha$  and  $\beta$  are positive integers

$$\rightarrow \text{Then } y_3 = u\left(\frac{2}{3}A(\alpha + \beta) + 5\right) \rightarrow = 1 \text{ since } \alpha + \beta \geq 2 \rightarrow y_3 = 1$$

$$\neq \alpha y_1 + \beta y_2 = 0$$

Not Linear

4

(ii) (4 points) Is the system time invariant?

$$\text{Let } x_k = x[n-k]$$

$$\text{so } y_k = u[x_k - A] \text{ where } u[0] = 0$$

$$\rightarrow y = u[x[n-k] - A] = y[n-k]$$

Thus, the system is Time Invariant

4

(b) (12 points) Assume that the system  $y[n] = f(x[n])$ , where  $f()$  is some unknown function, is Linear Time Invariant (LTI). Can you determine if the following systems are LTI? Briefly explain why.

(i) (4 points)  $z[n] = y[n-5] + y[n-2]$

Linearity: Let  $x_1 = ax[n] \xrightarrow{f()} y_1[n]$   
 $x_2 = bx[n] \xrightarrow{f()} y_2[n]$   
 $x_3[n] = x_1 + x_2 \xrightarrow{LTI} y_3[n] = ay_1 + by_2$

$y_3[n] \rightarrow z_3[n] = y_3[n-5] + y_3[n-2]$   
 $= ay_1[n-5] + by_2[n-5] = \underline{az_1 + bz_2}$

where  $ay_1[n] \rightarrow az_1[n]$ ,  $ay_2[n] \rightarrow az_2[n]$

Linear

Time Invariant: Let  $x_k = x[n-k]$   
 $\xrightarrow{f()} y_k \stackrel{LTI}{=} y[n-k]$

$y[n-k] \rightarrow y[n-k-5] + y[n-k-2]$   
 $= z[n-k] \checkmark \quad TI$

The sum of two LTI system outputs in itself

is also LTI.

(ii) (4 points)  $z[n] = y^2[n] = [f(x)]^2$

Linearity: Let  $y_1 = a y[n] \rightarrow z_1[n] = a^2 y^2[n]$

$y_2 = b y[n] \rightarrow z_2[n] = b^2 y^2[n]$

As can be observed,  $z_1[n] \neq a z[n] =$

$z_2[n] \neq b z[n]$

$a \cdot x[n] \rightarrow ay[n] \rightarrow \underline{a^2 z[n]}$

Not linear

Not Linear

Thus, not LTI

4



(iii) (4 points)  $z[n] = y[2n]$

Linearity: Let  $y_1 = ay[n]$

Let  $y_2 = by[n]$

$$\rightarrow y_3 = y_1 + y_2 \iff z_3 = ay[2n] + by[2n] \\ = az_1[n] + bz_2[n]$$

Linear ✓

Thus,  $ax_1 + bx_2 \xrightarrow{f(\cdot)} ay_1 + by_2 \rightarrow az_1 + bz_2$

Time Variance: Let  $y_k = y[n-k]$

$$y_k \rightarrow z_k = y_k[2n] = y[2n-k] \neq z[n+k] = y[2n-2k]$$

$$\text{Thus } x[n-k] \xrightarrow{f(\cdot)} y[n-k] \not\rightarrow z[n-k]$$

Not TI,

Therefore Not LTI

⚡

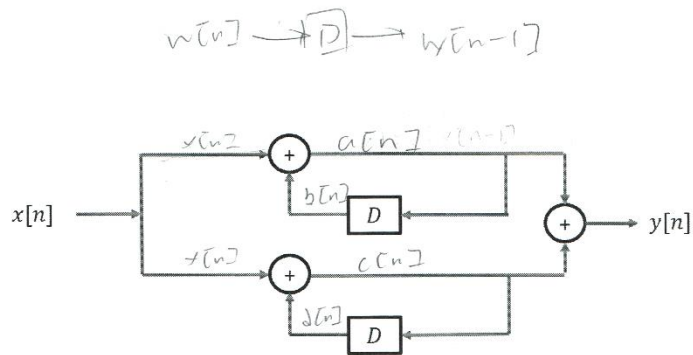


Figure 1: System for Problem 2(c)

(c) (20 points) Consider a relaxed system as shown in Fig. 1

(i) (8 points) Can you write the input-output equations for this system?

$$\begin{aligned}
 y[n] &= a[n] + c[n] \\
 &= x[n] + b[n] + x[n] + d[n] \\
 &= 2x[n] + a[n-1] + c[n-1] \\
 &= 2x[n] + (x[n-1] + b[n-1]) + (x[n-1] + d[n-1]) \\
 &= 2x[n] + 2x[n-1] + a[n-2] + c[n-2] \\
 &\quad \dots \\
 &= 2x[n] + 2x[n-1] + 2x[n-2] \dots
 \end{aligned}$$

$$y[n] = 2 \sum_{k=0}^{\infty} x[n-k]$$

✓ 3+5

(ii) (4 points) Is this a BIBO stable system? (explain why?)

if  $|x[n]| \leq B$  for all  $n$ ,

$$\text{then } y[n] = 2 \sum_{k=0}^{\infty} x[n-k]$$

$$\hookrightarrow \text{Maximum}_{h=0} \rightarrow 2 \sum_{k=0}^{\infty} B = \infty \cdot B = \infty$$

Not BIBO stable

4

- (iii) (8 points) Assume you connect two of these relaxed systems, shown in Fig. 1, in series, what is the impulse response of the overall equivalent system?

$$x[n] \rightarrow y[n] \rightarrow z[n]$$

$$\downarrow = 2 \sum_{k=0}^{\infty} x[n-k] \quad \downarrow = 2 \sum_{j=0}^{\infty} y[n-j] \cdot \left( \sum_{k=0}^{\infty} \delta[n-j-k] \right)$$

$$z[n] = \sum_{j=0}^{\infty} y[n-j] = 2 \sum_{j=0}^{\infty} \left( 2 \sum_{k=0}^{\infty} x[n-j-k] \right) = 4 \sum_{j=0}^{\infty} \sum_{k=0}^{\infty} x[n-j-k]$$

$$h_1[n] = 2 \sum_{k=0}^{\infty} \delta[n-k] = 2u[n]$$

$$h_2[n] = 2 \sum_{k=0}^{\infty} 2u[n-k] = 4 \sum_{k=0}^{\infty} u[n-k]$$

$$= \boxed{4r[n]} = h_2[n]$$

2

**Problem 3 (36 points) :** The following questions are not related to each other.

- (a) (14 points) True or false: Let  $x[n]$  be a real periodic signal with even period  $N$ , and  $c_k$  be the coefficients of its associated discrete-time Fourier series. Then, there always exist  $j_1, j_2$  such that  $c_{j_1}$  and  $c_{j_2}$  are purely real. Mathematically justify your answer.

a.  $c_k = \langle x[n], \phi_k[n] \rangle$

$$\phi_{j_1} = e^{j \frac{2\pi}{N} j_1 n}, \quad \phi_{j_2} = e^{j \frac{2\pi}{N} j_2 n}$$

$$\frac{2\pi}{N} j_1 = \pi \rightarrow 2\pi j_1 = N\pi$$

$$j_1 = \frac{N}{2} = \text{int}$$

for  $\phi_{j_1}, \phi_{j_2}$  to be real,

then  $\frac{2\pi}{N} j_1 = k_1 \pi, \frac{2\pi}{N} j_2 = k_2 \pi$  where  $k_1, k_2$  are integers

$$\rightarrow 2\pi j_1 = N k_1 \pi \quad \rightarrow 2\pi j_2 = N k_2 \pi$$

$$\rightarrow j_1 = \frac{N k_1}{2} \quad \rightarrow j_2 = \frac{N k_2}{2}$$

10

Since  $k_1, k_2$  are integers and  $N$  is even,

there will always exist  $\phi_{j_1}, \phi_{j_2}$  which are purely real,

and b/c  $c_k = \frac{1}{N} \sum_{n=0}^{N-1} x[n] \phi_k^*[n]$  where

$x[n]$  is real, there will always exist

$c_{j_1}, c_{j_2}$  which are purely real!

*You also need to find  $j_1$  &  $j_2$  such that  $x[n]$  is real, so that  $c_{j_1}, c_{j_2}$  are purely real!*

$\frac{2\pi}{T} = 3$       $\frac{2\pi}{T} = 3$       $\frac{2\pi}{T} = 3$

(b) (8 points) Find the discrete-time Fourier Series coefficients for the following signal:

Period:  $T=3$       $x[n] = \sin\left(\frac{2\pi n}{3}\right) \cos\left(\frac{\pi n}{3}\right)$       $\frac{2\pi}{T} = 3$  per

$$\sin\left(\frac{2\pi}{3}n\right) \cos\left(\frac{\pi}{3}n\right)$$

$$= \frac{1}{4j} \left( e^{j\frac{2\pi}{3}n} - e^{-j\frac{2\pi}{3}n} \right) \left( e^{j\frac{\pi}{3}n} + e^{-j\frac{\pi}{3}n} \right)$$

$$= \frac{1}{4j} \left( e^{j\pi n} + e^{j\frac{\pi}{3}n} - e^{-j\frac{\pi}{3}n} + e^{-j\pi n} \right)$$

$$\frac{1}{4j} \left( e^{j\pi n} + e^{j\frac{\pi}{3}n} - e^{-j\frac{\pi}{3}n} + e^{-j\pi n} \right)$$

$c_0 = 0$   
 $c_1 = \frac{1}{4j}$   
 $c_{-1} = \frac{1}{4j}$   
 $c_4 = \frac{1}{4j}$   
 $c_{-4} = -\frac{1}{4j}$

6

- (c) (14 points) Assume that for a periodic signal  $x[n]$  with period  $N$  you find the DTFS coefficients  $c_k$ . As we discussed in class, the Fourier Series coefficients  $c_k$  can also be thought as the values of a periodic signal with the same period  $N$ . Lets call this periodic signal  $c[k]$ , that is,

$$c[k] = c_k$$

Your friend claims that if you take the Fourier Series expansion of this periodic signal  $c[k]$ , you will get coefficients  $d_m$  that will be sufficient to retrieve the original signal  $x[m]$ . Are they right? Mathematically justify your answer.

$$c[k] = \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}kn}$$

$$d_m = \frac{1}{N} \sum_{k=0}^{N-1} c[k] e^{-j\frac{2\pi}{N}km}$$

$$= \frac{1}{N} \sum_{k=0}^{N-1} \left( \frac{1}{N} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}mn} \right) e^{-j\frac{2\pi}{N}km}$$

8

$$d_m[n] = \frac{1}{N^2} \sum_{k=0}^{N-1} \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}n(m+k)}$$

$$C[c[k]] = \sum_{m=0}^{N-1} d_m e^{-j\frac{2\pi}{N}mn}$$

$$x[n] = \sum_{k=0}^{N-1} c_k e^{-j\frac{2\pi}{N}kn}$$