

UCLA DEPARTMENT OF ELECTRICAL ENGINEERING

EE113: DIGITAL SIGNAL PROCESSING

Midterm 1 Exam Solution

Date: October 28, 2019, Duration: 1 hour

**INSTRUCTIONS:**

- The exam has 3 problems and 6 printed pages.
- The exam is closed-book.
- One double-sided cheat sheets of A4 size is allowed.
- Calculator is allowed.

**Your name:** \_\_\_\_\_

**Student ID:** \_\_\_\_\_

Table 1: Score Table

Problem	a	b	c	d	e	Total	Score
1	3	3	3	3	3	15	
2	5	5	5			15	
3	10					10	
Sum						40	

## Problem 1 (15 pts)

Consider the following system:  $y[n] = 0.5x[3n] + 0.5x[2n - 1]$

- (a) (3pts) Is the system linear? Prove your answer.
- (b) (3pts) Is the system time-invariant? Prove your answer.
- (c) (3pts) Is the system causal? Prove your answer.
- (d) (3pts) Is the system BIBO stable? Prove your answer.
- (e) (3pts) Suppose that  $x[n]$  is a periodic signal with a fundamental period of  $N_x = 110$ , what will be the fundamental period of  $y[n]$ ,  $N_y$ ?

### Solution:

- (a) The system is linear. (1 pt)

$$\begin{aligned} Sys\{\alpha_1 x_1[n] + \alpha_2 x_2[n]\} &= 0.5\alpha_1 x_1[3n] + 0.5\alpha_2 x_2[3n] + 0.5\alpha_1 x_1[2n - 1] + 0.5\alpha_2 x_2[2n - 1] \\ &= \alpha_1 Sys\{x_1[n]\} + \alpha_2 Sys\{x_2[n]\} \quad (2 \text{ pts}) \end{aligned}$$

- (b) The system is time-varying. (1 pt)

$$Sys\{x[n - k]\} = 0.5x[3n - k] + 0.5x[2n - k - 1] \quad (1 \text{ pt})$$

$$y[n - k] = 0.5x[3n - 3k] + 0.5x[2n - 2k - 1] \quad (1 \text{ pt})$$

- (c) The system is not causal. (1 pt)

For example, to obtain  $y[n]$  at  $n = 1$ , we need the values of  $x[n]$  at  $n > 1$ . Therefore, the system is not causal.

$$y[1] = 0.5x[3] + 0.5x[1] \quad (2 \text{ pts})$$

- (d) The system is BIBO stable. (1 pt)

$$|x[n]| \leq B_x \forall n \Rightarrow |y[n]| \leq B_x \forall n \quad (2 \text{ pts})$$

- (e) Let  $y_1[n] = 0.5x[3n]$  and  $y_2[n] = 0.5x[2n - 1]$ .

The period of  $y_2[n]$  is  $N_{y_2} = 110$ . (1 pt)

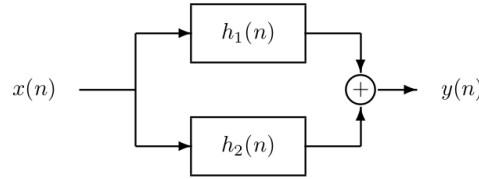
The period of  $y_1[n]$  is  $N_{y_1} = 55$ . (1 pt)

$N_y$  is the least common multiple of  $N_{y_1}$  and  $N_{y_2}$ , therefore  $N_y = 110$ . (1 pt)

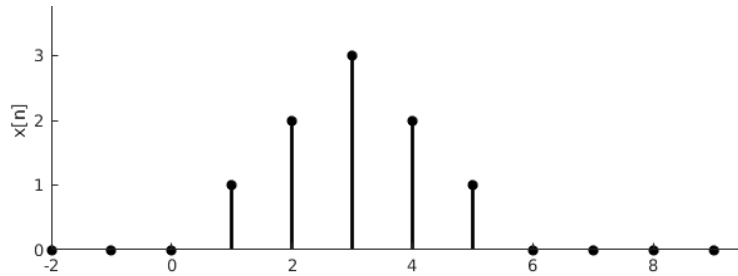
Full marks given for only stating the correct answer also.

## Problem 2 (15 pts)

Consider the system composed of parallel connection of two LTI systems.



- (a) (5 points) If unit-step response of the equivalent system (the response when the input is a unit-step function) is  $y[n] = r[n+1] - r[n-1]$  and  $h_1[n] = u[n] - 2u[n-1] + u[n-2]$ , find and sketch  $h_2[n]$ .
- (b) (5 points) Find the equivalent impulse response of the system  $h_{eq}[n]$ . The equivalent response is defined by the following relation:  $y[n] = x[n] * h_{eq}[n]$ .
- (c) (5 points) Find the response of the system  $y[n]$  for  $x[n]$  shown in the figure below.



**Solution:**

- (a) The system response to  $u[n]$  is

$$y[n] = u[n] * (h_1[n] + h_2[n]) = u[n] * h_1[n] + u[n] * h_2[n] \quad (1 \text{ pt})$$

Furthermore,

$$\begin{aligned} u[n] * h_1[n] &= u[n] * (u[n] - 2u[n-1] + u[n-2]) \\ &= u[n] * u[n] - 2u[n-1] * u[n] + u[n-2] * u[n] \\ &= r[n+1] - 2r[n] + r[n-1] \quad (2 \text{ pts}) \end{aligned}$$

Where we have used the fact that  $u[n] * u[n] = r[n+1]$ . Now we can find  $h_2[n]$ ,

$$\begin{aligned} u[n] * h_2[n] &= y[n] - u[n] * h_1[n] \\ u[n] * h_2[n] &= r[n+1] - r[n-1] - r[n+1] + 2r[n] - r[n-1] \end{aligned}$$

$$u[n] * h_2[n] = 2r[n] - 2r[n - 1]$$

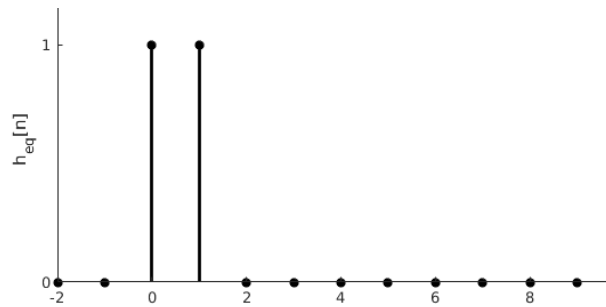
$$u[n] * h_2[n] = u[n] * 2u[n - 1] - u[n] * 2u[n - 2]$$

$$u[n] * h_2[n] = u[n] * (2u[n - 1] - 2u[n - 2])$$

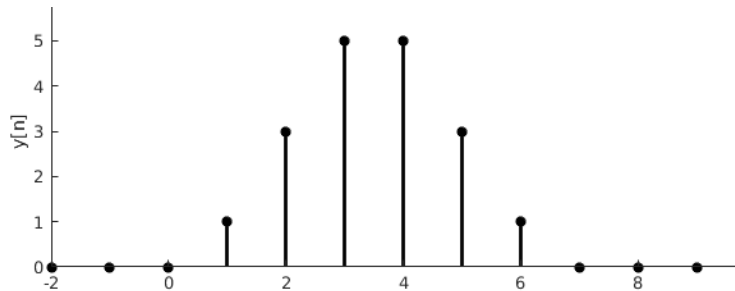
Hence,  $h_2[n] = 2u[n - 1] - 2u[n - 2] = 2\delta[n - 1]$ . (1 pt)

Sketch (1 pt)

- (b) Since this is an addition of the outputs of two parallel systems,  $h_{eq} = h_1[n] + h_2[n] = u[n] - 2u[n - 1] + u[n - 2] + 2u[n - 1] - 2u[n - 2] = u[n] - u[n - 2]$ . (Points awarded even if incorrect  $h_2[n]$  is used.) 2 points awarded for the equation. 3 points for the correct answer.



- (c) The response of the system is shown. Points awarded even if incorrect  $h_{eq}[n]$  is used.



### Problem 3 (10 pts)

Let  $\tilde{x}[n]$  be a periodic signal with period  $N$ . Its DTFS representation is given by

$$\tilde{x}[n] = \sum_{k=0}^{N-1} \tilde{c}_k e^{j\frac{2\pi}{N}kn},$$

where  $\tilde{c}_k$  are the DTFS coefficients.

Show that if  $\tilde{x}[n]$  is a complex signal and conjugate symmetric ( $\tilde{x}^*[n] = \tilde{x}[-n]$ ), then  $\text{Im}\{\tilde{c}_k\} = 0$ .

**Solution 1:**

$$\tilde{c}_k = \sum_{n=0}^{N-1} x[n] e^{-j\frac{2\pi}{N}kn}$$

We take the conjugate of both sides of the expression above

$$\tilde{c}_k^* = \sum_{n=0}^{N-1} x^*[n] e^{j\frac{2\pi}{N}kn} \quad (2 \text{ pts})$$

Using the fact that  $\tilde{x}^*[n] = \tilde{x}[-n]$

$$\tilde{c}_k^* = \sum_{n=0}^{N-1} x[-n] e^{j\frac{2\pi}{N}kn} \quad (2 \text{ pts})$$

We make the substitution  $m = -n$

$$\tilde{c}_k^* = \sum_{m=1-N}^0 x[m] e^{-j\frac{2\pi}{N}km} \quad (2 \text{ pts})$$

We can freely shift the starting index of the summation above without changing its value and we shift it by  $N - 1$  to the right

$$\tilde{c}_k^* = \sum_{m=0}^{N-1} x[m] e^{-j\frac{2\pi}{N}km} \quad (2 \text{ pts})$$

The expressions for  $\tilde{c}_k^*$  and  $\tilde{c}_k$  are identical, therefore  $\tilde{c}_k^* = \tilde{c}_k$ . This implies that  $\text{Im}\{\tilde{c}_k\} = 0$ . (2 pts)

**Solution 2:**

$$\tilde{x}[n] = \sum_{k=0}^{N-1} \tilde{c}_k e^{j\frac{2\pi}{N}kn}$$

$$\tilde{x}^*[n] = \sum_{k=0}^{N-1} \tilde{c}_k^* e^{-j\frac{2\pi}{N}kn}$$

$$\tilde{x}[-n] = \sum_{k=0}^{N-1} \tilde{c}_k e^{-j\frac{2\pi}{N}kn}$$

$$\tilde{x}[-n] - \tilde{x}^*[n] = \sum_{k=0}^{N-1} (\tilde{c}_k - \tilde{c}_k^*) e^{-j\frac{2\pi}{N}kn} = 0 \quad (4 \text{ pts})$$

$$2 \sum_{k=0}^{N-1} \text{Im}\{\tilde{c}_k\} e^{-j\frac{2\pi}{N}kn} = 0 \quad (3 \text{ pts})$$

This implies that either  $\text{Im}\{\tilde{c}_k\}$  is 0 or orthogonal to  $e^{-j\frac{2\pi}{N}kn}$ . However, since  $e^{-j\frac{2\pi}{N}kn}$  is an orthogonal basis, its null space is  $\{\emptyset\}$ . therefore  $\text{Im}\{\tilde{c}_k\} = 0$ . (3 pts)