

**EE 110, Fall 2009, Midterm Exam – October 27, 2009**

**Instructions:** This exam booklet consists of four problems, blank sheets for the solutions, reference sheets with mathematical identities, and additional blank sheets. Please follow these instructions while answering your exam:

1. Write your name and student identification number below.
2. Write the names of students to your left and right as well.
3. You have 1 hour 45 minutes to finish your exam.
4. You are not allowed any cheat sheets.
5. Write your solutions in the provided blank sheets after each problem.
6. The sheets marked “Additional Sheets” at the end of the blanket will NOT be graded. These sheets are provided for your rough calculations only.
7. Write your solutions clearly. Illegible solutions will NOT be graded.
8. Be brief.

NAME: \_\_\_\_\_

STUDENT ID: \_\_\_\_\_

NAMES OF ADJACENT STUDENTS:

LEFT: \_\_\_\_\_

RIGHT: \_\_\_\_\_

<b>Problem</b>	<b>Score</b>
<b>#1</b>	<b>/40</b>
<b>#2</b>	<b>/20</b>
<b>#3</b>	<b>/20</b>
<b>#4</b>	<b>/20</b>
<b>Total</b>	<b>/100</b>

**Problem 1:** Refer to the circuit schematic shown in Figure 1.

(a) Obtain the phasor domain representation for the circuit shown in the figure.

(b) Determine the sinusoidal steady state expressions for  $v_A(t)$ ,  $i_B(t)$ , and  $i_R(t)$ .

(c) Determine the Thevenin's equivalent of the circuit seen by the 10 Volts sinusoidal source i.e. the circuit to the right of terminals "x" and "y".

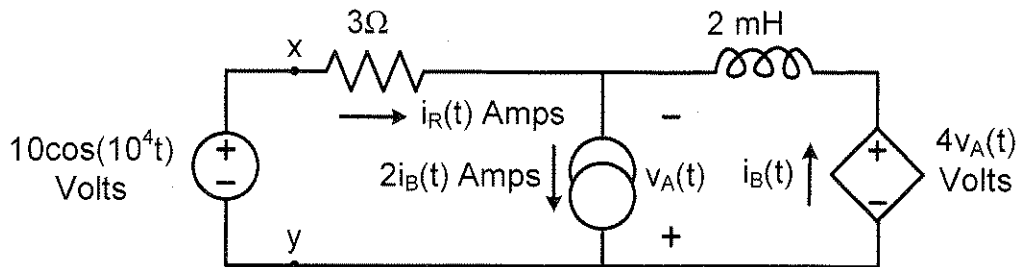
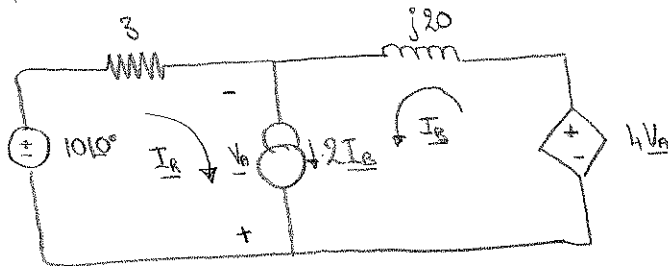


Figure 1.

(8 + 20 + 12 = 40 points)

A)



B)

Summing Currents.

$$\underline{I_R} + \underline{I_B} = 2\underline{I_B}$$

$$\Rightarrow \underline{I_R} = \underline{I_B}$$

MESH ANALYSIS

$$\textcircled{1} \quad 10\angle 0^\circ = 3\underline{I_B} - v_A$$

$$\textcircled{2} \quad 4v_A = j20\underline{I_B} - v_A$$

$$\Rightarrow v_A = j4\underline{I_B} = j4\underline{I_R}$$

$$10\angle 0^\circ = (3 - j4)\underline{I_B}$$

$$\underline{I_B} = \underline{I_R} = 1.2 + 1.6j = 2\angle 53.13^\circ$$

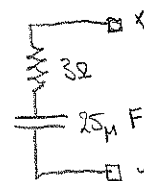
$$\underline{v_A} = -6.4 + 4.8j = 8\angle 143.13^\circ$$

$$i_R(t) = i_B(t) = 2\cos[10^4 t + 53.13^\circ]$$

$$v_A(t) = 8\cos[10^4 t + 143.13^\circ]$$

c)  $V_{th} = 0$  [no IND. CURRENT/VOLTAGE SOURCES]

$$Z_{th} = \frac{V_{oc}}{I_{sc}} = \frac{10\angle 0^\circ}{\underline{I_R}} = 3 - 4j$$







**Problem 2:** Refer to Figure 2 for this problem. A complex power of  $3000 + 4000j$  Volt-Amperes is delivered to the load under maximum power transfer conditions.

- (a) Determine the value of  $X$ . Is the reactance,  $jX$ , capacitive or inductive?  
 (b) Determine the value of  $V_{in}$ .

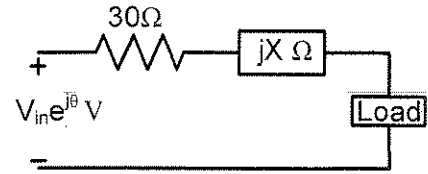
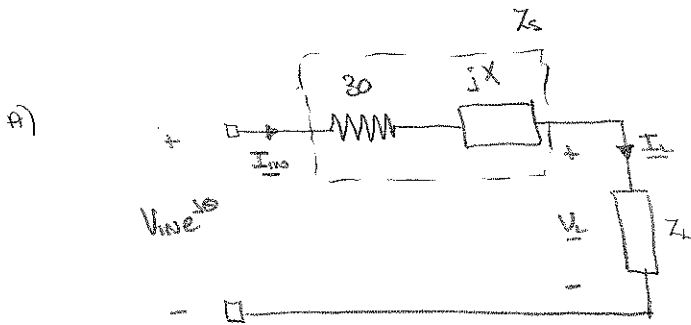


Figure 2

(10 + 10 = 20 points)



MAXIMUM POWER TRANSFER

$$Z_L = Z_s^*$$

$$\Rightarrow Z_L = 30 - jX$$

Complex Power

$$S = \frac{1}{2} V_L I_L^* = \frac{1}{2} |I_L|^2 Z_L$$

$$\Rightarrow \angle S = \angle Z_L$$

$$\angle 3000 + j4000 = \angle 30 - jX$$

$\Rightarrow$  ANGLES CAN ONLY BE THE SAME WHEN  $X = -40$

$\Rightarrow jX = -j40$ . SINCE THE IMPEDANCE IS IMAGINARY AND NEGATIVE IT MUST BE CAPACITIVE. (C)

B)

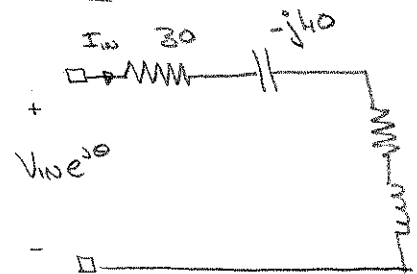
$$|I_L| = \sqrt{\frac{2S}{Z_L}} = \sqrt{\frac{2(3000 + j4000)}{30 + j40}} = \sqrt{200}$$

NOTE THAT  $Z_{in} = 60$

$$\Rightarrow I_{in} = \sqrt{200} e^{j\omega t}$$

$$Z_{in} = 60 = \frac{V_{in} e^{j\omega t}}{\sqrt{200} e^{j\omega t}} \Rightarrow V_{in} = 60\sqrt{200} \text{ V}$$

$$= 848.528 \text{ Volts.}$$



**Problem 3:** Refer to Figure 3 for this problem. It is observed that the input and output voltages of the linear electrical network in Figure 3(a) are related by the following differential equation:

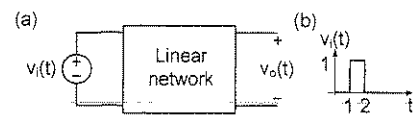


Figure 3

$$v_o(t) = v_i(t) - \frac{dv_o(t)}{dt}$$

The input to the network,  $v_i(t)$ , is shown in Figure 3(b). Assume that  $v_i(t) = v_o(t) = 0$  for all  $t < 0$ .

- (a) Derive an expression for the Laplace transform of  $v_i(t)$ .
  - (b) Derive an expression for the Laplace transform of  $v_o(t)$ .
- (8 + 12 = 20 points)

**Solution:**

A) 
$$v_i(t) = u(t-1) - u(t-2)$$

$$V_i(s) = \frac{e^{-s}}{s} - \frac{e^{-2s}}{s}$$

B) 
$$v_o(t) = v_i(t) - \frac{dv_o(t)}{dt}$$

$$V_o(s) = V_i(s) - [sV_o(s) - \cancel{v_o(0^+)}]$$

$$\Rightarrow V_o(s) = \frac{V_i(s)}{1+s} = \frac{1}{s(s+1)} [e^{-s} - e^{-2s}]$$

**Problem 4:** Refer to Figure 4 for this problem. The ideal transformers have zero initial conditions.

- (a) Determine the equivalent impedance  $Z_{xy}$  for an angular frequency of  $\omega = 10^4$  rad/s.  
 (b) How does  $Z_{xy}$  change if the angular frequency is changed to  $\omega = 10^6$  rad/s?

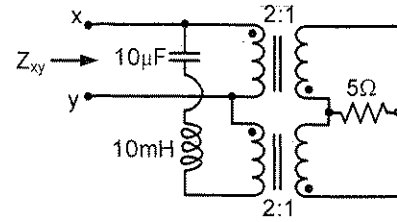
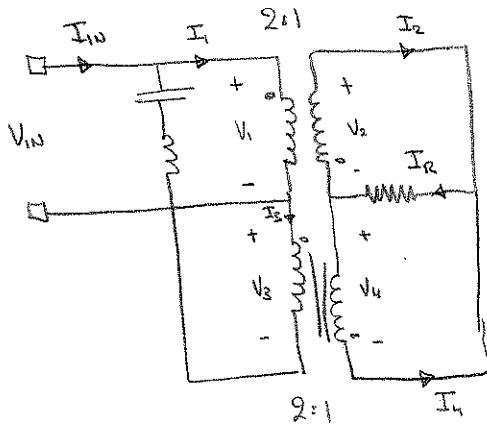


Figure 4

(15 + 5 = 20 points)

**Solution:**

A)



XFRM VOLTAGE EQNS

$$\frac{V_1}{2} = -V_2 \quad ; \quad \frac{V_3}{2} = -V_4 \quad ; \quad V_2 = -V_4$$

$$\begin{aligned} V_1 &= V_{IN} \\ V_2 &= -V_{IN}/2 \\ V_3 &= -V_{IN} \\ V_4 &= V_{IN}/2 \end{aligned}$$

XFRM CURRENT EQNS

$$2I_1 = -I_2$$

$$2I_3 = I_4$$

$$I_R = I_2 + I_4 = -2I_1 + 2I_3$$

I/P CURRENT

$$\textcircled{2} \quad I_{IN} = I_1 - I_3$$

$$\textcircled{1} \quad R = S = \frac{V_4}{I_2 + I_4} = \frac{-V_{IN}/2}{-2I_1 + 2I_3}$$

Solving  $\textcircled{1}$  +  $\textcircled{2}$  FOR  $V_{IN}$  AND  $I_{IN}$

$$S = \frac{-V_{IN}/2}{-2(I_{IN})} = \frac{V_{IN}}{4I_{IN}}$$

$$8 \quad \Rightarrow \quad Z_{IN} = 20$$

B) INPUT IMPEDANCE IS INDEPENDENT OF FREQUENCY



## Reference Sheet #1

### Trigonometric Identities:

$$\sin A = \cos(A - 90^\circ) = \cos(A - \pi/2)$$

$$\cos A = \sin(A + 90^\circ) = \sin(A + \pi/2)$$

$$\cos(A \pm B) = \cos A \cos B \mp \sin A \sin B$$

$$\sin(A \pm B) = \cos A \cos B \pm \sin A \sin B$$

$$\cos A + \cos B = 2 \cos((A+B)/2) \cos((A-B)/2)$$

$$\cos A - \cos B = -2 \sin((A+B)/2) \sin((A-B)/2)$$

$$\sin A + \sin B = 2 \sin((A+B)/2) \cos((A-B)/2)$$

$$\sin A - \sin B = 2 \cos((A+B)/2) \sin((A-B)/2)$$

$$\cos 2A = 2 \cos^2 A - 1 = \cos^2 A - \sin^2 A = 1 - 2 \sin^2 A$$

$$\sin 2A = 2 \sin A \cos A$$

$$a \cos A + b \sin A = \sqrt{a^2 + b^2} \cos(A - \tan^{-1}(b/a))$$

### Complex Arithmetic:

$$\operatorname{Re}\{z_1 \pm z_2\} = \operatorname{Re}\{z_1\} \pm \operatorname{Re}\{z_2\}$$

$$\operatorname{Im}\{z_1 \pm z_2\} = \operatorname{Im}\{z_1\} \pm \operatorname{Im}\{z_2\}$$

$$\operatorname{Re}\{z_1 z_2\} = \operatorname{Re}\{z_1\} \operatorname{Re}\{z_2\} - \operatorname{Im}\{z_1\} \operatorname{Im}\{z_2\}$$

$$\operatorname{Im}\{z_1 z_2\} = \operatorname{Re}\{z_1\} \operatorname{Im}\{z_2\} + \operatorname{Im}\{z_1\} \operatorname{Re}\{z_2\}$$

$$e^{j\theta} = \cos \theta + j \sin \theta$$

$$x + jy = r e^{j\theta} \text{ where } r = \sqrt{x^2 + y^2}, \theta = \tan^{-1}(y/x)$$

$$r e^{j\theta} = x + jy \text{ where } x = r \cos \theta, y = r \sin \theta$$

$$|z_1 z_2| = |z_1| |z_2|, \text{ angle}(z_1 z_2) = \text{angle}(z_1) + \text{angle}(z_2)$$

$$|1/z| = 1/|z|, \text{ angle}(1/z) = -\text{angle}(z)$$

$$(x + jy)^* = x - jy, \text{ angle}(z^*) = -\text{angle}(z)$$

$$(z_1 z_2)^* = (z_1)^* (z_2)^*$$

### Quadratic Equations:

$$\text{The roots of } ax^2 + bx + c = 0 \text{ are } x = \frac{-b \pm \sqrt{b^2 - 4ac}}{2}$$

## Reference Sheet #2

### Laplace Transforms:

$$\delta(t) \leftrightarrow 1$$

$$u(t) \leftrightarrow 1/s$$

$$tu(t) \leftrightarrow 1/s^2$$

$$e^{-at}u(t) \leftrightarrow \frac{1}{s+a}$$

$$te^{-at}u(t) \leftrightarrow \frac{1}{(s+a)^2}$$

$$\sin(\omega t)u(t) \leftrightarrow \frac{\omega}{s^2 + \omega^2}$$

$$\cos(\omega t)u(t) \leftrightarrow \frac{s}{s^2 + \omega^2}$$

$$e^{-at} \sin(\omega t)u(t) \leftrightarrow \frac{\omega}{(s+a)^2 + \omega^2}$$

$$e^{-at} \cos(\omega t)u(t) \leftrightarrow \frac{s+a}{(s+a)^2 + \omega^2}$$

### Operational Laplace Transforms:

$$k_1 f_1(t) + k_2 f_2(t) \leftrightarrow k_1 F_1(s) + k_2 F_2(s)$$

$$\frac{df(t)}{dt} \leftrightarrow sF(s) - f(0^-)$$

$$\frac{d^2 f(t)}{dt^2} \leftrightarrow s^2 F(s) - sf(0^-) - \frac{df(0^-)}{dt}$$

$$\int_0^t f(x) dx \leftrightarrow \frac{F(s)}{s}$$

$$f(t-a)u(t-a), a > 0 \leftrightarrow e^{-as}F(s)$$

$$e^{-at}f(t) \leftrightarrow F(s+a)$$

$$tf(t) \leftrightarrow -\frac{dF(s)}{ds}$$

$$t^n f(t) \leftrightarrow (-1)^n \frac{d^n F(s)}{ds^n}$$

$$\frac{f(t)}{t} \leftrightarrow \int_s^\infty F(u) du$$

## Additional Sheets

