

EE 10H, Fall 2017, Midterm Exam #1 – May 2, 2017

Instructions: This exam booklet consists of problems, blank sheets for the solutions, reference sheets with mathematical identities, and additional blank sheets. Please follow these instructions while answering your exam:

1. Write your name and student identification number below.
2. Write the names of students to your left and right as well.
3. You have 1 hour 45 minutes to finish your exam.
4. Write your solutions in the provided blank sheets after each problem.
5. The sheets marked "Scratch..." will NOT be graded. These sheets are provided for your rough calculations only.
6. Write your solutions clearly. You may box in your final answer. Illegible solutions will NOT be graded.
7. Be brief.
8. Open text and open notes. NO homework or homework solutions!

NAME _____

STUDENT ID _____

NAME _____

LEFT _____

RIGHT _____

Problem	Score
#1	25/25
#2	35/40
#3	25/35
Total	85 / 100

$q = CV$
 $v = C \frac{dq}{dt}$
 $v = \frac{1}{C} \int_{-\infty}^t i_C(t) dt$
 $v = L \frac{di}{dt}$

Problem 1: Consider the circuit shown in Figure 1.

- (a) Draw a graph for this circuit.
- (b) Identify (draw) a tree that does **not** include the branches C_2 and R . Choose it such that you have the minimum number of equations to solve.
- (c) Use mesh current method to write the equations for this circuit. Make sure that $i_C(t)$ and $i_R(t)$ are two of the unknowns. Assume that no energy is stored in any of the inductors or capacitors.

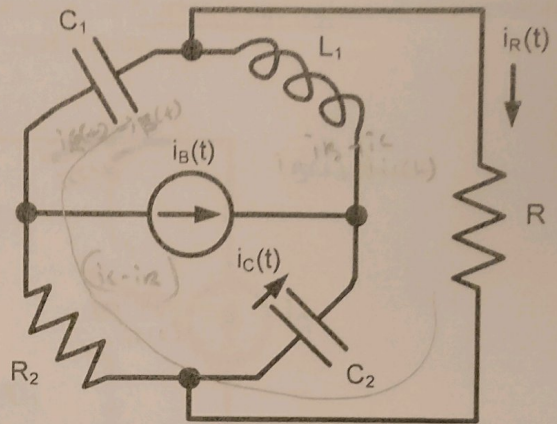
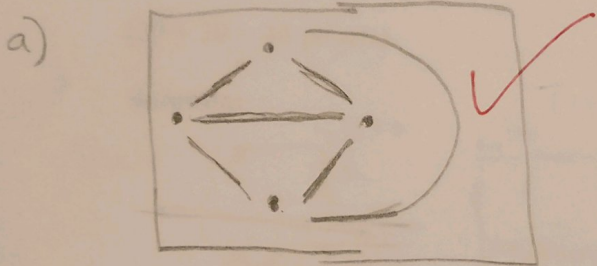


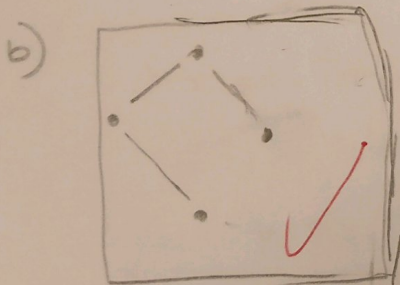
Figure 1.

(5 + 10 + 10 = 25 points)

Solution:

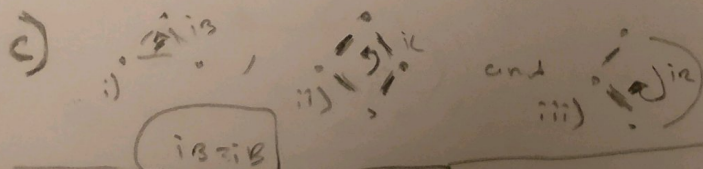


Graph for this circuit



Tree that does not include C_2 or R

We have 2 choices, node voltage = $4 - 1 = 3$ eq or
 Mesh current = $6 - (4 - 1) = 3$
 But $i_B(t)$ is known, so we just need 2 equations using mesh current.



3 such loops, first one we already know that i is just i_B

Thus: ii) $\frac{1}{C_2} \int_{-\infty}^t i_C(t) dt + L_1 \frac{d}{dt} (i_C(t) + i_B(t)) + \frac{1}{C_1} \int_{-\infty}^t (i_C(t) + i_B(t)) dt + R_2 (i_C(t) - i_R(t)) = 0$

iii) $i_R(t) \cdot R + R_2 \cdot (i_R(t) - i_C(t)) + \frac{1}{C_1} \int_{-\infty}^t (i_R(t) - i_B(t) - i_C(t)) dt = 0$

* Continued on back *

Problem 2: Refer to Figure 2 for this problem. Calculate the Norton's equivalent of this network looking into the terminals 1-1'. Use any method of your choice.

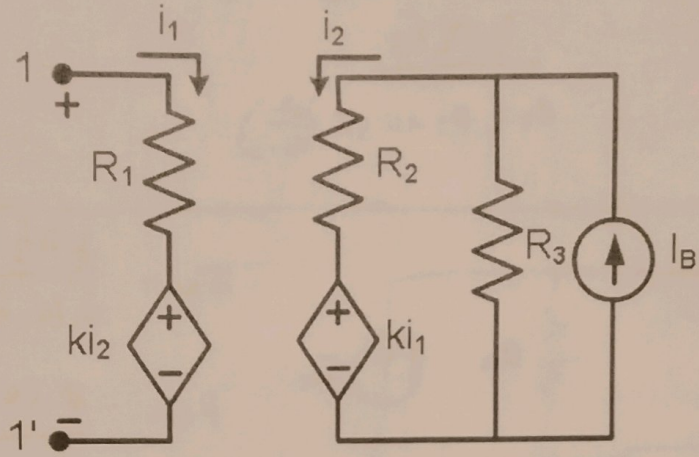
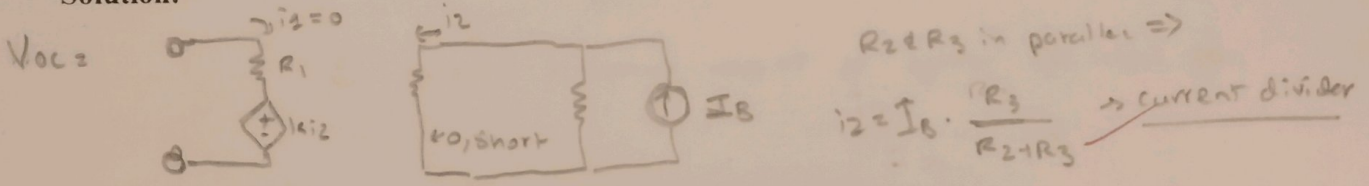


Figure 2.

(20 + 20 = 40 points)

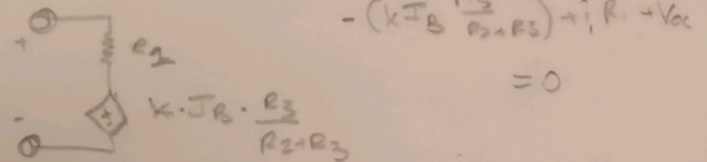
Solution:



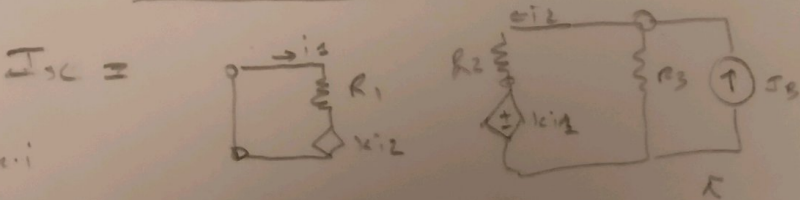
R_2 & R_3 in parallel \Rightarrow
 $i_2 = I_B \cdot \frac{R_3}{R_2 + R_3}$ \rightarrow current divider

We can then write this as:

$i_2 = 0$, so just



$V_{oc} = k \cdot I_B \cdot \frac{R_3}{R_2 + R_3}$



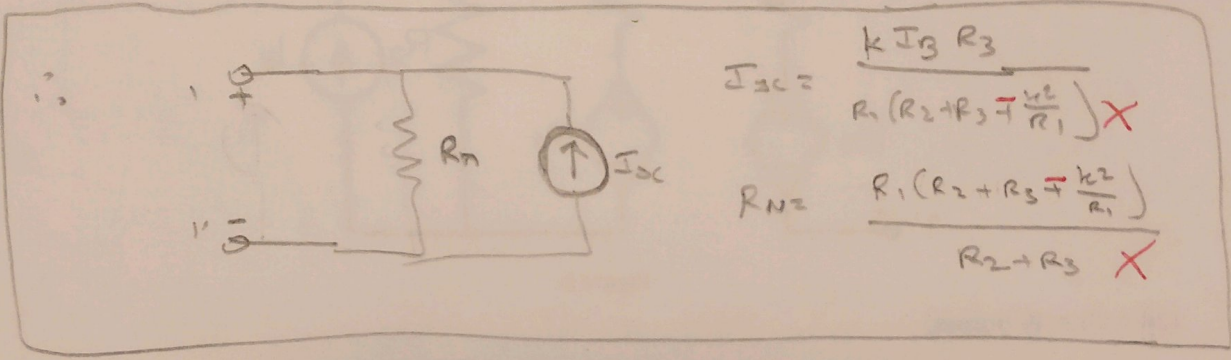
$i_2 \neq 0$ in this case because not OC
 so $V=iR$, $i_2 = -\frac{k i_2}{R_1} = -I_{sc}$

$V_{2x} = i$
 \rightarrow looks like a resistor
 \rightarrow k_2 ohms
 so $k i_2 = k \cdot \frac{k I_2}{R_1} = -\frac{k^2 I_2}{R_1}$

~~$\frac{V_2 - k i_2}{R_2} + \frac{V_1}{R_3} - I_3 = 0$~~
 ~~$V_1 = (R_2 + R_3) \cdot k i_2 + I_3 R_3$~~
 looks like resistor w/ value $-\frac{k^2}{R_1} \rightarrow$ so $(R_2 + \frac{k^2}{R_1}) \parallel R_3 \Rightarrow$
~~NOT .5~~
 $\Rightarrow i_2 = \frac{k}{R_1} \cdot \frac{I_B R_3}{R_2 + \frac{k^2}{R_1} + R_3} = I_{sc}$
 $\frac{R_2}{2} = R \checkmark$

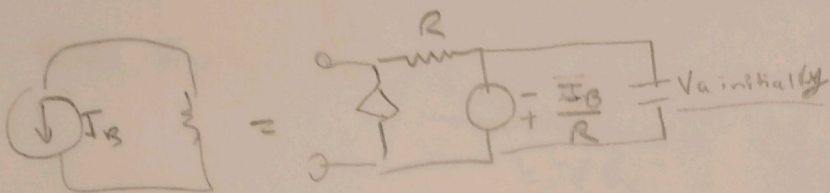
$$R_N = \frac{V_{oc}}{I_{sc}} = \frac{k I_3 R_3}{R_2 + R_3} = \frac{R_1 (R_2 + R_3 + \frac{k^2}{R_1})}{R_2 + R_3}$$

$$R_1 (R_2 + R_3 + \frac{k^2}{R_1})$$



object w/ voltage \propto to own current & resistor !!

* Continued on next page *



Caps hate change in voltage,
Inductors hate change in current

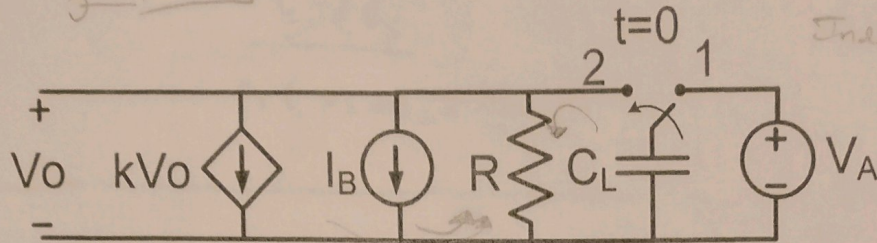


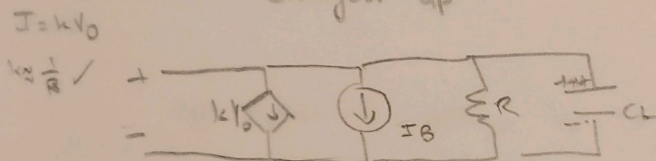
Figure 3
don't combine!!

Current leaving is positive

all will have same potential after ∞

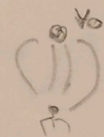
Problem 3: Refer to Figure 3 for this problem. Find $V_0(t)$ for all $t \geq 0$. (35 points)

Solution: C_L has been at one for a very long time, and so charged up



so all components in parallel, we can find

graphs:



$q = CV$
 $i = C \frac{dV}{dt}$

$V(0^-) = V_A$
 \therefore Node voltage tells us: $kV_0 + I_B + \frac{V_0}{R} + C_L \frac{dV_0}{dt} = 0$ ✓

$V(0^+) =$ will start discharging,
 $\frac{dV_0}{dt} + \frac{V_0}{R C_L} + \frac{kV_0}{C_L} = -\frac{I_B}{C_L} \Rightarrow V_0 \left(\frac{1}{R C_L} + \frac{k}{C_L} \right)$

Integrating factor = $e^{\left(\frac{1}{R C_L} + \frac{k}{C_L}\right)t} = e^{\left(\frac{1+kR}{R C_L}\right)t}$ $\frac{1}{R C_L} + \frac{k}{C_L} = \frac{1+kR}{R C_L}$

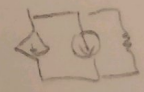
Integrate both sides and use fundamental theorem of calculus

$\Rightarrow e^{\frac{1+kR}{R C_L} t} \cdot V_0(t) - V_0(0) = \frac{R C_L}{1+kR} \cdot \left[e^{\frac{1+kR}{R C_L} t} \left(\frac{I_B}{C_L} \right) \right]_{0^+}$

$= e^{\frac{1+kR}{R C_L} t} V_0(t) - V_0(0) = \frac{R C_L}{1+kR} \left[e^{\frac{1+kR}{R C_L} t} \left(\frac{I_B}{C_L} \right) - \frac{I_B}{C_L} \right]$

$1+kR$ is just constant since $k = \frac{1}{R} \cdot \omega = \text{const}$

$V_0(0) = ? \Rightarrow$ well @ 0^- equals



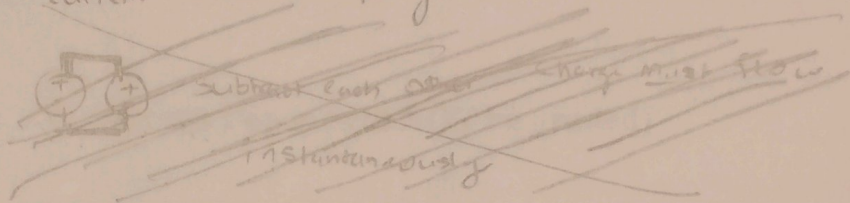
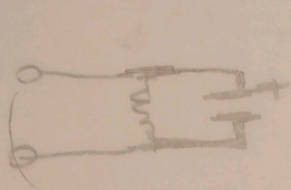
$R \cdot (I_B + V_0) = V_0 \Rightarrow R I_B + R V_0 = V_0$

$V_0 - R V_0 = R I_B \Rightarrow V_0(0^-) = \frac{R I_B}{1-R}$ ✓

What's $V_0(0^+)$?

at 0^+ , the capacitor is attached so \Rightarrow

current starts dissipating, ~~but~~



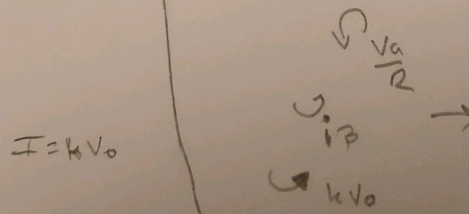
instantaneously

\Rightarrow $V_0(t) = \dots$

$$\Rightarrow V_0(t) = e^{-\frac{(1+kR)t}{RCL}} (V_0(0^+)) + e^{-\frac{(1+kR)t}{RCL}} \left[\frac{RCL}{1+kR} \left(e^{\frac{(1+kR)t}{RCL}} \frac{I_B}{CL} - \frac{I_B}{CL} \right) \right]$$

$$V_0(t) = \left[\left(\frac{V_A - I_B R}{1+kR} \right) + \frac{I_B \cdot RCL}{(1+kR) \cdot CL} \right] e^{-\frac{(1+kR)t}{RCL}} + \frac{RCL I_B}{(1+kR) \cdot CL}$$

Assuming that is $V_0(0^+)$, else, the rest stays the same, just the initial voltage will change. $\Rightarrow \tau = \frac{RCL}{1+kR}$
 $RCL / \text{constant}$



$$R \cdot \left[\frac{V_A}{R} - I_B - kV_0 \right] = V_0(0^+)$$

$$V_0(0^+) = R \left[\frac{V_A}{R} - I_B - \frac{k \cdot R I_B}{1-kR} \right]$$

we know $V_0(0^-) = 0$ but, can't use since will change!

Current source depending on voltage (changing voltage)...



these two are constant (currents don't really need to worry)

$\downarrow \frac{V_A}{R} \uparrow$ (other two currents), now $R \frac{V_A}{R} - R I_B - R k V_0 = V_0$
 $\Rightarrow k \cdot \text{the} \Rightarrow$

What if just $V_0(0^+) + k \cdot R \cdot V_0(0^+) = \frac{V_A}{R} - I_B R$

$$V_0(0^+) = \frac{V_A - I_B R}{1+kR}$$

$$V_0(0^+) = V_A$$

one of these