- 1. Consider the following systems.
	- (a)  $S_1$ :  $y(t) = \sin(x(t))$ . Is it linear? Time-invariant? Causal? Explain. Nonlinear:  $sin(k_1x_1(t) + k_2x_2(t)) \neq k_1 sin(x_1(t)) + k_2 sin(x_2(t))$ . Time-invariant:  $\sin(x(t-\tau)) = y(t-\tau)$ . Causal and memoryless:  $y(t_0)$  depends only on  $x(t_0)$ .
	- (b)  $S_2$ :  $y(t) = \int_{t-2}^{t} (t-\sigma)x(\sigma)d\sigma$ . Is it linear? Time-invariant? Causal? Explain. O<sub>2</sub>.  $y(t) - J_{t-2}(t - \sigma)x(\sigma)d\sigma$ . Is it inteat: The invariant: Causar: Explain.<br>One can write  $y(t) = \int_{-\infty}^{\infty} [U(\sigma - t + 2)U(t - \sigma)](t - \sigma)x(\sigma)d\sigma = x(t) * h(t)$ , with  $h(t) = [U(-t + 2)U(t)]t$ . Note that  $h(t)$  can be also written as  $h(t) =$  $[U(t)-U(t-2)]t$ . The system is then a linear, time-invariant system with impulse response  $h(t)$ . Because  $h(t) = 0, t < 0$ , the system is also causal.
	- (c) Consider now the **time-invariant** system  $S_3$ , and assume that the responses to the two signals  $x_1(t) = U(t)$  and  $x_2(t) = U(t) - U(t-2)$  are respectively  $y_1(t) = U(t-1)$  and  $y_2(t) = U(t-1) - U(t-2)$ . Is  $S_3$  linear? If the system were linear, the output corresponding to  $x_3(t) = x_1(t) - x_2(t)$  would be equal to  $y_3(t) = y_1(t) - y_2(t)$ . We have that  $x_3(t) = U(t-2)$  and  $y_3(t) = U(t-2)$ . This contradicts the fact that the system is time-invariant:  $x_3(t)$  is simply equal to  $x_1(t-2)$ , therefore the corresponding output should be equal to  $y_1(t-2) = U(t-3)$ . [If you are curious to know what the input-output relation for this system is, here it is:  $y(t) = x(t)x(t-1)$ .]
- 2. Consider the system  $S_2$  from the previous question.
	- (a) Compute its impulse response,  $h(t)$ . See above:  $h(t) = t[U(t) - U(t-2)].$
	- (b) Compute the output of the system when  $x(t) = U(t-1)$ . Use the convolution integral. Because the system is time-invariant, one can compute the output corresponding to  $\bar{x}(t) = x(t+1) = U(t)$  and then shift it to the right by 1.  $\bar{y}(t) = \int_{t-2}^{t} (t-\sigma)\bar{x}(\sigma)d\sigma = \int_{t-2}^{t} (t-\sigma)U(\sigma)d\sigma$ . If  $t < 0$ ,  $\bar{y}(t) = 0$ . If  $\sum_{t=2}^{t} (t-\sigma) U(\sigma) d\sigma$ . If  $t < 0$ ,  $\bar{y}(t) = 0$ . If  $0 < t < 2$ , then  $\bar{y}(t) = \int_0^t (t - \sigma) d\sigma = [t\sigma - \sigma^2/2]_{\sigma=0}^t = t^2/2$ . If  $t > 2$ , then  $\bar{y}(t) = \int_{t-2}^{t} (t-\sigma)d\sigma = [t\sigma - \sigma^2/2]_{\sigma=t-2}^{t} = 2$ . The desired output  $y(t)$  is equal to  $y(t) = \bar{y}(t-1).$
- 3. Consider now the system  $S_4$ , described by the following input-output relation:

$$
\mathcal{S}_4: y(t) = t^2 x(t).
$$

(a) Suppose that the output of system  $S_2$  above is sent as input to system  $S_4$ ,

$$
\stackrel{a(t)}{\rightarrow} \boxed{\mathcal{S}_2} \stackrel{b(t)}{\rightarrow} \boxed{\mathcal{S}_4} \stackrel{c(t)}{\rightarrow}.
$$

What is the output of the overall system when  $a(t) = \delta(t-1)$ ? The output of  $S_2$  is equal to  $b(t) = h(t-1) = (t-1)[U(t-1)-U(t-3)]$ , because the system is linear, time-invariant. Therefore  $c(t) = t^2(t-1)[U(t-1)-U(t-3)]$ .

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(b) The two systems are now switched:

$$
\stackrel{a(t)}{\rightarrow} \boxed{\mathcal{S}_4} \stackrel{b(t)}{\rightarrow} \boxed{\mathcal{S}_2} \stackrel{c(t)}{\rightarrow}.
$$

What is the output of this combined system when  $a(t) = \delta(t-1)$ ? The output of  $S_4$  is now equal to  $b(t) = t^2 \delta(t-1) = \delta(t-1)$ . Therefore  $c(t) = h(t-1) =$  $(t-1)[U(t-1)-U(t-3)].$ 

- 4. Compute the Laplace transform of the following signals:
	- (a)  $f_1(t) = e^{-t} \cos(t-1)U(t-1);$ Because  $f_1(t) = e^{-1}g(t-1)$ , where  $g(t) = e^{-t}\cos(t)U(t)$ , then  $F_1(s) = e^{-1}e^{-s}G(s)$  $e^{-(s+1)}(s+1)/[(s+1)^2+1]$ , with a pole in  $s=-1$ , therefore the region of convergence is given by  $\Re\{s\} > -1$ .
	- (b)  $f_2(t) = \int_0^t (t \sigma) e^{-\sigma} d\sigma;$  $J_2(t) = \int_0^t (t - \sigma)e^{-\sigma} d\sigma;$ <br>Re-write  $f_2(t)$  as  $f_2(t) = \int_{-\infty}^{\infty} [(t - \sigma)U(t - \sigma)][U(\sigma)e^{-\sigma}] d\sigma = [tU(t)] * [U(t)e^{-t}],$ then  $F_2(s) = (1/s^2)[1/(s+1)] = 1/[s^2(s+1)]$ , with poles in  $s = 0$  and  $s = -1$ , therefore the ROC is  $\Re\{s\} > 0$ .
	- (c)  $f_3(t) = \int_0^t \sin(\sigma) \cos(\sigma) d\sigma$ .<br>
	Re-write  $f_3(t)$  as  $f_3(t) = \int_{-\infty}^{\infty} [(1/2) \sin(2\sigma) U(\sigma)] [U(t-\sigma)] d\sigma = [(1/2)U(t) \sin(2t)]*$  $U(t)$ , then  $F_3(s) = (1/2)[2/(s^2+4)](1/s) = 1/[s(s^2+4)]$  with poles in  $s = 0$ ,  $s = \pm 2i$ , therefore the ROC is  $\Re\{s\} > 0$ .