

ECE 102 Midterm

TOTAL POINTS

88.5 / 106

QUESTION 1

Problem 1 19 pts

1.1 a.i 1.5 / 3

+ **3 pts** Correct (stating this is false) with appropriate justification

+ **3 pts** Correct (stating this is false) with appropriate justification, but with a minor algebra mistake somewhere (e.g., the integrand should be squared, or other trivial algebra mistake).

+ **2 pts** Correct (stating this is false) with correct calculations, but an incorrect or insufficient justification.

+ **2 pts** Incorrect (stating this is true) due to correct calculations but incorporating a bound. Note, $E_y < \infty$ means it's an energy signal; you want to show E_y is infinite.

✓ + **1.5 pts** Incorrect (stating that this is true) because of miscalculation of the integral of 1 across all time, but had the problem set up correct, making the correct calculation and indicating what conclusion would have led to the right answer.

+ **0 pts** Incorrect (stating that this is true) or no statement with incorrect or no justification.

1.2 a.ii 3 / 3

✓ + **3 pts** Correct with appropriate justification.

+ **0 pts** Incorrect with incorrect justification.

1.3 a.iii 2 / 3

+ **3 pts** Correct with appropriate justification (either through using the properties of an LTI system such as convolution; or stating time invariance preserves the period; or the result from class that complex exponentials are eigenfunctions of LTI systems)

✓ + **2 pts** Correct, with a time invariance statement, but that isn't directly related to why this leads to

periodicity.

+ **2 pts** Incorrect, but with some work that is almost along the right lines (correct setup) but makes a mistake.

+ **1 pts** Correct, but with inadequate justification (see comments); most common inadequate justification was that this is true because of linearity, but this isn't true, consider $y(t) = x(t) * \exp(-t)$.

+ **0 pts** Correct with no justification or a restatement of the question as fact.

+ **0 pts** Incorrect with incorrect justification; or no answer.

1.4 b 7.5 / 10

✓ + **5 pts** The system is linear with correct work.

+ **5 pts** The system is NOT time-invariant with correct work.

+ **4 pts** The system is time-invariant with correct work. (Most often it was not recognizing the $t - \tau \geq 1$ statement made the system time-variant, not time-invariant.)

+ **3 pts** Showed linear, but concluded it was not linear.

✓ + **2.5 pts** The system is time-invariant without time-shifting the condition (for output) or time-shifting the condition (for input).

+ **2.5 pts** Tried to show system was not linear with counter example; but did algebra incorrectly.

+ **1 pts** Stated definition of linearity; did not show that the system was linear or did so incorrectly.

+ **1 pts** Stated definition of time invariance; did not show the system was time variant.

+ **0 pts** No proof or inadequate proof of linearity; or incorrect statement of what linearity is.

+ **0 pts** No proof of time variance.

QUESTION 2

Problem 2 17 pts

2.1 a 10 / 11

- 0 pts Correct
- 9 pts No attempt to find $h_2(t)$

✓ - 1 pts Did provide the general definition of causality but without specifically explaining it for the given system, or did not use enough/right argument.

- 2 pts Did provide the simplified expression of $h_1(t)$ but without explaining how

- 5 pts Incomplete/wrong attempt to find $h_2(t)$
- 3 pts Did not simplify $h_1(t)$ / wrong $h_1(t)$
- 1 pts Wrong time argument of one of $h_2(t)$ terms
- 2 pts No or wrong answer for the causality of the system.

- 2 pts Wrong use of convolutions properties.
- 3 pts Wrong $h_2(t)$ because of wrong $h_1(t)$
- 3 pts Wrong use of properties of elementary signals

- 1 pts Wrong copying of $h_1(t)$
- 2 pts Mathematical mistake when taking the derivative of the integration

2.2 b 4 / 6

- 0 pts Correct
- 6 pts No answer or wrong answer.
- 2 pts One of the answer term is wrong or incomplete.
- 4 pts Writing the answer as convolution without further simplification or solving incorrectly the convolution.

- 1 pts Wrong simplification of $x(t)$
- 3 pts Incomplete attempt
- 1 pts Approach is correct but wrong copying/shifting of function

- 3 pts Wrong use of shifting or sifting properties

✓ - 2 pts Convolution with wrong impulse response.

QUESTION 3

Problem 3 16 pts

3.1 a 5 / 8

- 0 pts Correct
- 8 pts No or wrong answer

✓ - 3 pts Wrong value of b or a .

- 4 pts Answer with no clear explanation.
- 5 pts Incomplete attempt.
- 4 pts Partially correct answer.
- 1 pts Wrong flipping limits.
- 2 pts wrong shifting limits

3.2 b 8 / 8

✓ - 0 pts Correct

- 5 pts Missing the sign of the function, or not considering the right shift or flip, (which leads to a wrong answer).
- 1 pts Wrong value for the area.
- 2 pts Missing the maximum value.
- 1 pts Wrong limits of the function after flipping and shifting it.
- 8 pts No or wrong answer.
- 6 pts Incomplete answer
- 2 pts Wrong or missing shifting
- 3 pts Had the right intuition, but found the wrong t and the wrong area.

QUESTION 4

Problem 4 20 pts

4.1 a 6 / 6

✓ - 0 pts Correct steps toward answer, using Fourier expansion, and proving correct relation between f_k and g_k .

- 4.5 pts Wrong steps from f_k to g_k , the integration for Fourier coefficients is written. However, the extracting g_k from integration or definition is not correct

- 1.5 pts wrong final step/missing final step
- 0.5 pts Missing a negative sign
- 2 pts wrong final presentation
- 4 pts wrong definition for f_k
- 1 pts wrong interpretation
- 6 pts no answer

4.2 b 6 / 6

✓ - 0 pts Correct

- 6 pts no answer
- 5 pts Wrong answer despite correct initial steps

(writing the fourier expansion)

- **4 pts** missing final answer with correct path
- **1 pts** missing a negative sign results in odd answers instead of even k s

4.3 C.i 4 / 4

✓ - **0 pts** Correct

- **3 pts** Wrong answer, writing the answer not in form of Fourier series, wrong attempt to write the even part
- **2.5 pts** correct initial steps but errors leading to incorrect final answer
- **4 pts** No answer
- **1 pts** minor error in answer (for example wrong negative sign), or missing final representation

4.4 C.ii 4 / 4

✓ - **0 pts** Correct

- **4 pts** No answer
- **3 pts** Wrong answer, taking initial steps toward answer but missing final answer

QUESTION 5

Problem 5 28 pts

5.1 a 10 / 10

✓ - **0 pts** Correct, writing the correct relation between $x(t)/y(t)$ and comparing output of delayed input and delayed output.

- **4 pts** Correct justification of not time-invariance but not complete reasoning (missing the convolution in input/output relation, wrong comparing of two output cases, taking $e^{j\omega t}$ out of convolution, not all correct convolution definition)
- **7 pts** No right format of delayed output and output of delayed input despite writing similar form of convolution, invalid identities leading to wrong forms (Although stating that the $e^{j\omega t}$ is main reason but is not a proof of non-TI)
- **2 pts** One part of solution (either the delayed output or output of delayed inout) is missing, wrong definition of convolution
- **10 pts** No answer
- **3 pts** Wrong format of delaying the convolution,

for delaying convolution, just one side of convolution needs to be shifted

- **3 pts** Correct answer but Comparing of integrations are not right because two or more functions don't have same format in integration

5.2 b 8.5 / 12

- **0 pts** Correct, considering all parts of system and also reasoning for real and even.
- **12 pts** No/ Wrong answer
- ✓ - **3.5 pts** Answer is for only output of $h(t)$ given the input $x(t)$, however the question asks for all system including the multiplier at the beginning, which shifts the $X(j\omega)$ and therefore the answer is the half left of $X(j(\omega-\omega_0/2))$
- **0.5 pts** no reasoning (or wrong) for even/real
- **2 pts** not mentioning whether the signal is real/even or not or both wrong answers
- **4 pts** correct path toward $H(j\omega)$ but not completely right (plotting the $x(j(\omega-\omega_0))$), however minor errors leading to wrong answer
- **1 pts** wrong acknowledging real or even (one of them)
- **1 pts** Minor error in value of $Y(i\omega)$
- **2.5 pts** Error in sketching $Y(j\omega)$ given X and $H(j\omega)$ / no clear plot for $Y(j\omega)$
- + **2 pts** partial credit for initial steps toward $H(j\omega)$

5.3 C 6 / 6

✓ - **0 pts** Correct

- **6 pts** No/Wrong Answer
- **2 pts** Wrong power of e (negative sign or the value), the correct power is $-j\omega 2/3$
- **2 pts** Wrong argument of Y (the correct form is $Y(j\omega/3)$)
- **5 pts** Correct path toward finding shift/scale, however error in steps
- **2 pts** Wrong scale factor
- **1 pts** wrong sign in either factor or power

QUESTION 6

Problem 6 6 pts

6.1 a 1 / 4

+ **4 pts** Correct; the most elegant way was to notice this is a real and even signal multiplied by $e^{jt/2}$ and so the phase is t .

+ **3.5 pts** Very minor mistake on modulation theorem (wrote ω instead of t , or forgot a factor of 2, or both); or other unsimplified answers.

+ **3 pts** Wrote the signal as a sum of rects and took the inverse Fourier transform, applied modulation correctly, didn't get to the final answer.

+ **2.5 pts** Wrote the signal as a sum of rects and took the inverse Fourier transform, but did not apply the modulation theorem correctly.

+ **2.5 pts** Noticed modulation but did not get answer

+ **2.5 pts** Manually took the inverse Fourier transform but did not extract the phase.

✓ + **1 pts** Wrote the signal as a sum of rects and recognized the inverse FT would be sincs; or attempted to convolve them.

+ **0 pts** Incorrect or unattempted. (Note, phase of $X(j\omega)$ is not the phase of $x(t)$).

6.2 b 2 / 2

✓ + **2 pts** Correct (or essentially correct)

+ **0.5 pts** Answer does not recognize $x(t) = 0$ for $t < 0$.

+ **0 pts** Incorrect or unattempted (even if you wrote the definition of the even component).

ECE 102, Fall 2018
Department of Electrical and Computer Engineering
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Midterm
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TAs: H. Salami, S. Shahshavari

UCLA True Bruin academic integrity principles apply.
Open: Two pages of cheat sheet allowed.
Closed: Book, computer, internet.
2:00-3:50pm.
Wednesday, 14 Nov 2018.

State your assumptions and reasoning.
No credit without reasoning.
Show all work on these pages.

3:52

Name: 

Signature: 

ID#: 

Problem 1	_____	/	19
Problem 2	_____	/	17
Problem 3	_____	/	16
Problem 4	_____	/	20
Problem 5	_____	/	28
BONUS	_____	/	6 bonus points
Total	_____	/	100 points + 6 bonus points

Problem 1 (19 points)

(a) (9 points) Determine if each of the following statements is true or false. Briefly explain your answer to receive full credit.

i. (3 points) If $x(t)$ is an energy signal, then $y(t) = x(t) + 1$ is also an energy signal.

if $x(t)$ energy,

$$\int_{-\infty}^{\infty} |x(t)|^2 dt \text{ is finite}$$

$$\int_{-\infty}^{\infty} |y(t)|^2 dt = \int_{-\infty}^{\infty} |x(t) + 1|^2 dt$$

will also be finite,

so $y(t)$ is an energy signal, **TRUE**

ii. (3 points) If $x(t)$ is an even signal, then $y(t) = x(t - 1)$ is also an even signal.

let $x(t) = t^2$

$$x(-t) = (-t)^2 = t^2 = x(t)$$

$$x(-t) = x(t) \text{ even}$$



if shifted, it will no longer be even.

false

iii. (3 points) If the input to an LTI system is periodic, then its output is also periodic.

LTI needs to hold:

(TI) delay input \rightarrow delay output

Homogeneity/Superposition

True

- (b) (10 points) Is the following system linear? Is it time invariant? (Check both properties).
Explain your answer.

$$y(t) = \begin{cases} x(t-1), & t \geq 1 \\ 0, & \text{otherwise} \end{cases}$$

TI: delay input = $\begin{cases} x(t-t_0-1), & t \geq 1 \\ 0, & \text{else} \end{cases}$

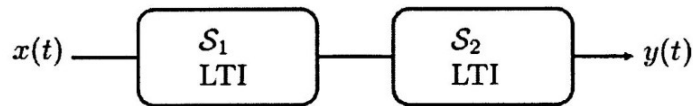
Same as delayed output, \checkmark

Linearity: $a x_1(t-1) + b x_2(t-1) = a y_1(t) + b y_2(t)$

\checkmark linear

piecewise isn't changing superposition or homogeneity

Problem 2 (17 points) Consider the series cascade of the following two systems:



The system S_1 is LTI with impulse response

$$h_1(t) = \int_{-\infty}^t u(\tau) \delta(\tau - 2) d\tau$$

The system S_2 is also LTI, with unknown impulse response $h_2(t)$ that we need to find. We are also given that, when the input $x(t)$ is $\delta(t)$, the output $y(t)$ is $r(t-3) + u(t-2)$.

Note: $r(t-3)$ is the ramp function delayed by 3.

This question continues on the next page.

$$h_1 * h_2 = h_{eq}$$

$$(h_1 * \delta) * h_2 = h_{eq} * \delta = h_{eq} \text{ (impulse response)}$$

$$r(t-3) + u(t-2) = \int_{-\infty}^t u(\tau) \delta(\tau-2) d\tau * h_2$$

$$r(t-3) + u(t-2) = \int_{-\infty}^t u(\tau) \delta(\tau-2) d\tau * h_2$$

$$= \int_{-\infty}^t \delta(\tau-2) d\tau * h_2$$

$$= [u(t-2) - u(-\infty)] * h_2$$

$\frac{d}{dt}$ both sides

$$r(t-3) + u(t-2) = u(t-2) * h_2$$

$$u(t-3) + \delta(t-2) = \delta(t-2) * h_2$$

$$u(t-3) + \delta(t-2) = h_2(t-2)$$

$$\boxed{h_2(t) = u(t-1) + \delta(t)}$$

- (a) (11 points) Find the impulse response $h_2(t)$ of the system \mathcal{S}_2 and determine if the system \mathcal{S}_2 is causal.

$$h_2(t) = u(t-1) + \delta(t)$$

\mathcal{S}_2 is causal, it does not depend on future time.

(b) (6 points) Find the output $y(t)$ to the following input:

$$x(t) = (1 + e^{-t})\delta(t+1) = (1 + e^{-1})\delta(t+1) \quad \text{sampling}$$
$$= (1+e)\delta(t+1)$$

$$y(t) = h_{eq}(t) * x(t)$$

$$y(t) = u(t-1) * x(t) + \delta(t) * x(t)$$

$$y(t) = u(t-1) * [(1+e)\delta(t+1)] + x(t)$$

$$y(t) = u(t-1) * [(1+e)\delta(t+1)] + (1+e)\delta(t+1)$$

$$= u(t+1-1)(1+e) + (1+e)\delta(t+1)$$

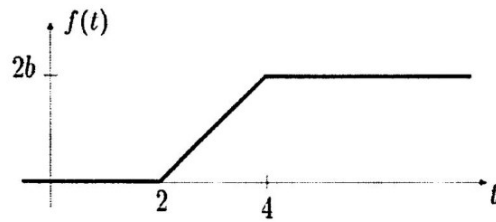
$$= u(t)(1+e) + (1+e)\delta(t+1)$$

$$y(t) = (1+e)(u(t) + \delta(t+1))$$

input x

Problem 3 (16 points)

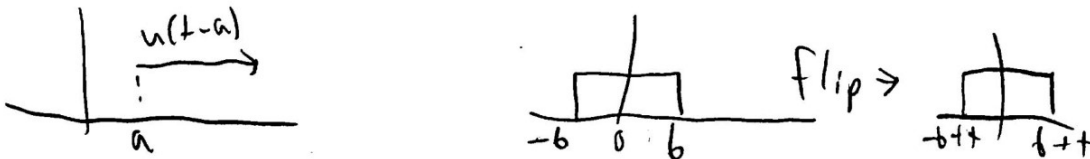
(a) (8 points) Consider the signal $f(t)$ shown below:



This signal can be written as

$$u(t-a) * \text{rect}\left(\frac{t}{2b}\right)$$

where $a > 0$ and $b > 0$. Find a and b . (Hint: use the flip and drag technique.)



Case 1: $b+t < a, 0$

occurs until $t=2$

$$b+2 < a$$

$b+t = a$ start intersect

$$b+2 = a$$

$$\frac{1}{2} + 2 = a$$

$$\boxed{a = 2.5}$$

Case 2: $\int u(\tau-a) \text{rect}\left(\frac{\tau}{2b}\right) d\tau$

norm

$2b = 1$, max height of rect = 1

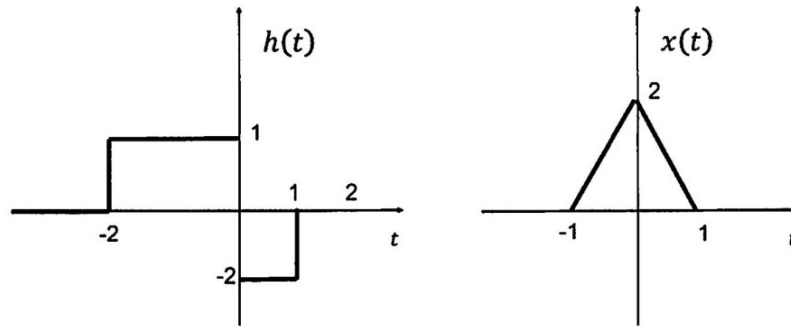
$$\boxed{b = \frac{1}{2}}$$

Case 3: $a < -b+t$

$$a+b < t$$

$$a+b < 4$$

- (b) (8 points) An input, $x(t)$, is given to an LTI system with impulse response $h(t)$. Both $x(t)$ and $h(t)$ are shown below.



Let $y(t)$ denote the output of the system, i.e., $y(t) = x(t) * h(t)$. Find the value of t at which the output $y(t)$ reaches its maximum value. Determine this maximum value.

Note: to answer this question, you do **not** need to find $y(t)$ for all t .

maximum value is when the most of each overlap, so when $x(t)$ goes from $[-2, 0]$

Flip and drag



$$1 + t = 0$$

$$t = -1$$

$$\int_{-\infty}^{\infty} h(\tau) \cdot x(t - \tau) d\tau$$

$$\int_{-\infty}^{\infty} h(\tau) \cdot x(-1 - \tau) d\tau$$

exists from $[-2, 0]$

$$\int_{-2}^0 h(\tau) \cdot x(-1 - \tau) d\tau$$

in this interval = 1

$$\int_{-2}^0 x(-1 - \tau) d\tau$$

$$= \text{Area} = \frac{1}{2} \cdot 2 \cdot 2 = 2$$

$$a \cdot t = -1$$

Problem 4 (20 points)

Consider the following two periodic signals $f(t)$ and $g(t)$. They both have the same period T_0 . Let f_k and g_k respectively denote the Fourier series coefficients of $f(t)$ and $g(t)$.

(a) (6 points) If $f(t) = -g(t + \frac{T_0}{2})$, how is f_k related to g_k ? Same $T_0 \rightarrow$ same ω_0

$$g(t) = \sum_{k=-\infty}^{\infty} g_k e^{jk\omega_0 t}$$
$$= \sum_{k=-\infty}^{\infty} g_k e^{jk \frac{2\pi}{T_0} t}$$

$$\omega_0 = \frac{2\pi}{T_0}$$

$$-g\left(t + \frac{T_0}{2}\right) = - \sum_{k=-\infty}^{\infty} g_k e^{jk\omega_0 \left(t + \frac{T_0}{2}\right)}$$

$$f(t) = -g\left(t + \frac{T_0}{2}\right) = - \sum_{k=-\infty}^{\infty} g_k e^{jk\omega_0 t} e^{jk\pi}$$

$\frac{2\pi}{T_0} \cdot \frac{T_0}{2}$

$$f(t) = \sum_{k=-\infty}^{\infty} \left[-g_k (-1)^k \right] e^{jk\omega_0 t}$$

So $f_k = -g_k (-1)^k$

(b) (6 points) If $f(t) = -f(t + \frac{T_0}{2})$, for what k are the coefficients f_k zero? $\omega_0 = \frac{2\pi}{T_0}$

$$\begin{aligned} \sum_{k=-\infty}^{\infty} f_k e^{jk\omega_0 t} &= - \sum_{k=-\infty}^{\infty} f_k e^{jk\omega_0 (t + \frac{T_0}{2})} \\ &= - \sum_{k=-\infty}^{\infty} f_k e^{jk\omega_0 t} e^{jk\omega_0 \frac{T_0}{2}} \\ &= - \sum_{k=-\infty}^{\infty} [(-f_k) (-1)^k] e^{jk\omega_0 t} \end{aligned}$$

$$f_k = (-f_k) (-1)^k$$

when k even: $f_k = -f_k$

the only number this works with is 0

\therefore even k makes f_k zero

(c) (8 points) This question has two parts. Note: part (c) is independent of parts (a) and (b).

i. (4 points) Let $f_e(t)$ denote the even part of $f(t)$. Express the Fourier series coefficients of $f_e(t)$ in terms of f_k .

$$f(t) = \sum_{-\infty}^{\infty} f_k e^{jk\omega_0 t} \quad \dots \quad f_e(t) = \frac{1}{2} [f(t) + f(-t)]$$

$$\begin{aligned} f_e(t) &= \frac{1}{2} \left[\sum_{-\infty}^{\infty} f_k e^{jk\omega_0 t} + \sum_{-\infty}^{\infty} f_k e^{jk\omega_0 (-t)} \right] \\ &= \frac{1}{2} \left[\sum_{-\infty}^{\infty} f_k e^{jk\omega_0 t} + \sum_{n=-\infty}^{\infty} f_{-n} e^{jn\omega_0 t} \right] \quad \begin{array}{l} \text{(bring negative to } k \text{ and} \\ \text{change variables)} \\ n = -k \end{array} \\ &= \frac{1}{2} \left[\sum_{-\infty}^{\infty} f_k e^{jk\omega_0 t} + \sum_{-\infty}^{\infty} f_{-k} e^{jk\omega_0 t} \right] \end{aligned}$$

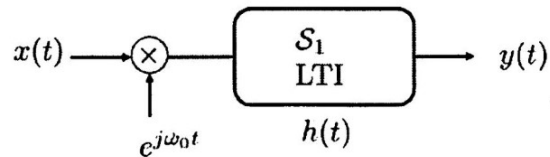
$$\boxed{f_e = \frac{f_k + f_{-k}}{2}}$$

ii. (4 points) Determine the DC component of $f_o(t)$, the odd part of $f(t)$.

If its odd, the average value has to be 0 [equal area on both sides of x-axis], so DC component is 0.

Problem 5 (28 points)

Consider the following system ($\omega_0 > 0$):



The system S_1 is LTI and $h(t)$ represents its impulse response.

- (a) (10 points) Show that the overall system, with input $x(t)$ and output $y(t)$, is not time-invariant.

$$y(t) = [x(t) \cdot e^{j\omega_0 t}] * h(t)$$

shift input, $x(t)$:

$$[x(t-t_0) e^{j\omega_0 t}] * h(t)$$

this is not the same as a shift in output:

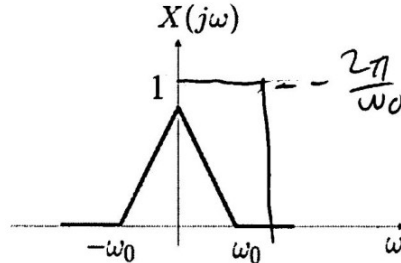
$$y(t-t_0) = [x(t-t_0) e^{j\omega_0(t-t_0)}] * h(t)$$

\therefore the system is not time invariant.

(b) (12 points) Consider the following impulse response for system S_1 :

$$h(t) = e^{j\frac{\omega_0}{2}t} \text{sinc}\left(\frac{\omega_0}{2\pi}t\right)$$

We give the system an input $x(t)$, where $x(t)$ has the following Fourier transform $X(j\omega)$:



Find and sketch the Fourier transform $Y(j\omega)$ of the corresponding output $y(t)$. After this, determine (i) if $y(t)$ is real and (ii) if $y(t)$ is even. Note: you do not need to give an expression for $Y(j\omega)$, a sketch of it is enough. There is space on the next page if needed.

~~$$Y(j\omega) = X((\omega - \omega_0)j) H(j\omega)$$~~

$$x(t) e^{j\frac{\omega_0}{2}t} \longleftrightarrow X(j(\omega - \frac{\omega_0}{2}))$$

$$\text{sinc}(t) \longleftrightarrow \text{rect}\left(\frac{\omega}{2\pi}\right)$$

$$x(at) \longleftrightarrow \frac{1}{|a|} X\left(j\frac{\omega}{a}\right)$$

$$\text{sinc}\left(\frac{\omega_0}{2\pi}t\right) \longleftrightarrow \frac{2\pi}{\omega_0} \text{rect}\left(\frac{2\pi}{\omega_0} \cdot \frac{\omega}{2\pi}\right) = \frac{2\pi}{\omega_0} \text{rect}\left(\frac{\omega}{\omega_0}\right)$$

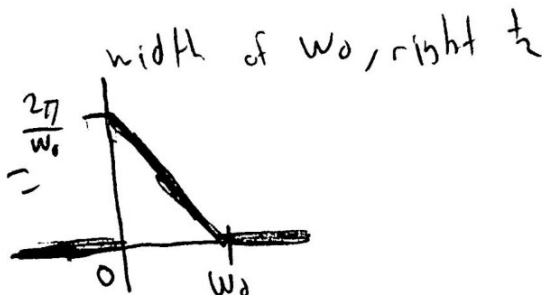
$$\text{now shift } \text{sinc}\left(\frac{\omega_0}{2\pi}t\right) e^{j\frac{\omega_0}{2}t} \longleftrightarrow \frac{2\pi}{\omega_0} \text{rect}\left(\frac{\omega - \frac{\omega_0}{2}}{\omega_0}\right) = \frac{2\pi}{\omega_0} \text{rect}\left(\frac{\omega}{\omega_0} - \frac{1}{2}\right)$$

$$H(j\omega) = \frac{2\pi}{\omega_0} \text{rect}\left(\frac{\omega}{\omega_0} - \frac{1}{2}\right)$$



$$Y(j\omega) = H(j\omega) X(j\omega)$$

$$\text{so } Y(j\omega) = \frac{2\pi}{\omega_0}$$



A multiplier does not change, here see next page

$$h_1 = e^{j\frac{\omega_0}{2}t} \text{sinc}\left(\frac{\omega_0}{2\pi}t\right)$$

$$h_{eq} = \delta(t) \cdot e^{j\omega_0 t} \times e^{j\frac{\omega_0}{2}t} \text{sinc}\left(\frac{\omega_0}{2\pi}t\right)$$

$$= h_{eq} \cdot \delta(t)$$

= // h_{eq} is the same \Downarrow

my answer to the previous

part stands

for $y(t)$ real, $Y^*(j\omega) = Y(-j\omega)$

$Y(j\omega)$ $Y^*(j\omega) \neq Y(-j\omega)$ because Y is
undefined for $j\omega < 0$.

so not real

also not wholly evn or odd.

(c) (6 points) Suppose

$$z(t) = y(3t - 2)$$

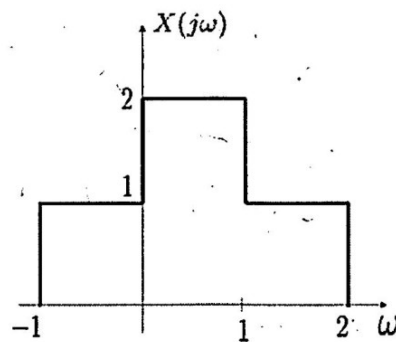
Express $Z(j\omega)$ in terms of $Y(j\omega)$. Note: part (c) is independent of parts (a) and (b).

shift: $Z(j\omega) \leftrightarrow Y(j\omega) e^{-j\omega 2}$

Scale: $Z(j\omega) \leftrightarrow \frac{1}{3} Y\left(\frac{j\omega}{3}\right) e^{-\frac{2}{3}j\omega}$

BONUS (6 points)

(a) (4 points) The Fourier transform $X(j\omega)$ of a signal $x(t)$ is given as follows:



Find the phase of $x^2(t)$.

$$X(j\omega) = \text{rect}\left(\frac{t}{2}\right) + \text{rect}\left(\frac{t}{2} - \frac{1}{2}\right)$$

X(j\omega)
use inverse fourier transform and it's sinc!

(b) (2 points) If a signal $x(t)$ is causal with $x(0) = 0$, how can we retrieve $x(t)$ from its even component $x_e(t)$?

$$x_e(t) = \frac{1}{2} [x(t) + x(-t)]$$

$\therefore x(t)$ for $t < 0$ is 0

so $x_e(t) = \frac{1}{2} [x(t)]$ ^{$[t > 0]$}

for $t > 0$, $x_e(t) = \frac{x(t)}{2} \rightarrow \boxed{x(t) = 2x_e(t)}$ ^{$[t > 0]$}

for $t < 0$: $x_e(t) = \frac{1}{2} [0 + x(-t)]$

$$x_e(t) = \frac{1}{2} x(-t)$$

$$x(-t) = 2x_e(t)$$

$$\boxed{x(t) = -2x_e(-t)}$$

$t < 0$