EE101 - Engineering Electromagnetics Midterm

1. Coaxial capacitor (50 points)

conductors. Consider a piece of coaxial cable of length  $\ell$  with two dielectric layers with permittivities  $\varepsilon_1$  and  $\varepsilon_2$ . You may consider the inner conductor (radius  $a$ ) and the outer conductor shell (radius  $4a$ ) to be perfect



(a) What is the capacitance C between the inner and outer conductors? (Your answer should be in terms of only geometric and material parameters.)  $\tau : \frac{\mathcal{G}}{\epsilon} = \frac{\mathcal{G}}{\nu \epsilon} = \frac{\mathcal{G}}{\sqrt{2} \epsilon} = \frac{\mathcal{G}}{\sqrt{2} \epsilon}$ 

$$
v = -\int E \cdot dl = -\int_{0}^{2\pi} t_1 \cdot \hat{v} dv + \int_{0}^{2a} t_2 \cdot \hat{v} dv
$$
  

$$
= -\int_{0}^{2a} \frac{\partial \hat{v}}{\partial z \partial \zeta} \cdot \hat{v} dv = -\int_{0}^{2a} \frac{\partial \hat{v}}{\partial z \partial \zeta} \cdot \hat{v} dv
$$
  

$$
= \frac{\partial}{\partial z \partial \zeta} \ln \left( \frac{2a}{\zeta} \right) + \frac{\partial}{\partial z \partial \zeta} \hat{v} \left( \frac{u_0}{\zeta a} \right)
$$
  

$$
= \frac{\partial}{\partial z \partial \zeta} \ln \left( \frac{2a}{\zeta} \right) + \frac{\partial}{\partial z \partial \zeta} \hat{v} \left( \frac{u_0}{\zeta a} \right)
$$

$$
C = \frac{2}{\sqrt{1}} = 3 - \frac{2\pi\ell}{\ell_1! \sqrt{2}} - \frac{1}{\epsilon_1 \cdot \epsilon_2}
$$
  
= 
$$
\frac{2\pi\ell}{\ell_1! \sqrt{2}} - \frac{\epsilon_1 \epsilon_2}{\epsilon_1 \cdot \epsilon_2} \sqrt{11}
$$

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 $\overline{a}$ 

## Midterm

 $\neg \varepsilon$ 

(b) Consider that the dielectric layers are lossy with conductivities  $\sigma_1$  and  $\sigma_2$  respectively. What is the resistance R between the inner and outer conductors?

Example 2.1 Find the first line, the second line is 
$$
k_1 + k_2
$$
 and the third line is  $k_1 + k_2$ .

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$$
= \frac{\ln 3}{2\pi \ell} \left( \frac{1}{\sigma_1} - \frac{1}{\sigma_2} \right) \sqrt{\frac{2\pi \ell}{2\pi \ell}}
$$

$$
= \frac{\ln 3}{2\pi \ell} \left( \frac{\sigma_1 + \sigma_2}{\sigma_1 \sigma_2} \right) \sqrt{|\pi|}
$$

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(a) (20 points) For part (a), ignore the metal ring in the figure.

Consider a solenoid with N turns and length *t,* and radius *a,* is driven by a current source  $i(t)=i<sub>0</sub>cos(\omega t)$ . As shown in the figure, we use the convention that a positive current is associated with current in the  $\phi$  direction. You may assume that the solenoid is long ( $\ell \gg a$ ). What is the voltage difference  $v(t)=V_+ - V$ , that appears across the terminals?

Your answer should be in terms of geometric, material, and fundamental parameters only (i.e. N,  $\ell$ ,  $a, b, \epsilon, \mu, \sigma,$  etc.).

$$
\oint_{\alpha} u_{\alpha}H = 1
$$
\n
$$
\frac{1}{n} \int_{\alpha}^{1} u_{\alpha}H = H_{\alpha} \hat{L} = \frac{Tr_{0} \hat{L}}{2} \hat{L}
$$
\n
$$
= \frac{1}{n} \oint_{\alpha} \frac{\partial u_{\alpha}H}{\partial x} = \oint_{\alpha} \beta \cdot dS
$$
\n
$$
= \iint_{\alpha}^{1} \frac{\partial u_{\alpha}H}{\partial x} = \frac{1}{n} \oint_{\alpha} \
$$

 $13/20$ 

EE101 - Engineering Electromagnetics Midterm

(b) (20 points) A metal ring with height h, radius b, and thickness d, and conductivity  $\sigma$  is placed around the center of the solenoid as shown in the figure. You may assume  $d \ll b$ . What is the current density  $J(t)$  that flows in the metal ring (give vector direction)? When answering this, you may neglect any contribution due the currents J(t) in the metal ring.

sicouding to Teneding's law, the changing magnetic field will  
induce a caated and a simple circle a negative field to  
optics the first magnetic field  
at the magnetic field  
at the magnetic field  
at the object of the direction of -2,  
at the magnitude of 
$$
\frac{0.65 (ud)}{2\pi r}
$$
  
 $\Rightarrow H = \frac{0.605 (at)}{2\pi r}$  (2)

$$
\frac{1}{2} \times \sqrt{1 - \frac{24}{94} - \frac{24}{94}} + \frac{1}{4} \left( \frac{24}{94} - \frac{14}{94} \right) + \frac{1}{4} \left( \frac{24}{94} - \frac{14}{94} \right) + \frac{1}{4} \left( \frac{1}{94} \left( \frac{1}{6} \left( \frac{1}{6} \right) - \frac{1}{94} \right) \right)
$$

$$
= -\hat{\phi} \left( \frac{\partial^{\mu} \hat{z}}{\partial \hat{z}} \right)
$$
  

$$
= -\hat{\phi} \frac{\partial \omega \hat{z}}{\partial \hat{z}} \left( \frac{\partial \hat{z}}{\partial \hat{z}} \right)
$$

$$
\int \frac{1}{\sqrt{1-x^2}} \cos(\omega t) \cos(\omega t)
$$
\n
$$
= \frac{\log(\omega t)}{\log(\omega t)} \quad \text{and} \quad \frac{1}{\sqrt{1-x^2}} \quad \text{
$$

 $10/20$ 

**EE101** - **Engineering Electromagnetics** Midterm

(c) ( $\mathcal{R}$  points) Does the approximation used in part (b) (neglecting any contribution due the currents **J**(t) in the metal ring) correspond to the low frequency limit ( $\omega$  is small compared to  $\tau_m^{-1}$ ), or the high frequency limit ( $\omega$  is large compared to  $\tau_m^{-1}$ )? Give a qualitative explanation why.

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Midterm

## UCLA Department of Electrical Engineering EE101 – Engineering Electromagnetics Winter 2012 Midterm, February 14 2012, (1:45 minutes)

OLUTIONS

Name

Student number

This is a closed book exam  $-$  you are allowed 1 page of notes (front+back).

Check to make sure your test booklet has all of its pages – both when you receive it and when you turn it in.

Remember – there are several questions, with varying levels of difficulty, be careful not to spend too much time on any one question to the exclusion of all others.

Exam grading: When grading, we focusing on evaluating your level of understanding, based on what you have written out for each problem. For that reason, you should make your work clear, and provide any necessary explanation. In many cases, a correct numerical answer with no explanation will not receive full credit, and a clearly explained solution with an incorrect numerical answer will receive close to full credit. CIRCLE YOUR FINAL ANSWER.

If an answer to a question depends on a result from a previous section that you are unsure of, be sure to write out as much of the solution as you can using symbols before plugging in any numbers, that way at you will still receive the majority of credit for the problem, even if your previous answer was numerically incorrect.

Please be neat – we cannot grade what we cannot decipher.



EE101 - Engineering Electromagnetics This page intentionally left blank.

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Midterm

1. Coaxial capacitor  $(50 \text{ points})$ 

Consider a piece of coaxial cable of length  $\ell$  with two dielectric layers with permittivities  $\varepsilon_1$  and  $\varepsilon_2$ . You may consider the inner conductor (radius  $a$ ) and the outer conductor shell (radius  $4a$ ) to be perfect conductors.



(a) What is the capacitance  $C$  between the inner and outer conductors? (Your answer should be in terms of only geometric and material parameters.)

E-field from line charge 
$$
\rho_R
$$
 = region 1  $\alpha < r < 2a$   $\overline{E}_r = \hat{r} \frac{\rho_e}{2\pi r \epsilon_1}$   
regularized  $\overline{E}_2 = \hat{r} \frac{\rho_e}{2\pi r \epsilon_2}$ 

$$
\bigvee_{i} = \bigvee(\alpha) - \bigvee(2\alpha) = -\int_{2\alpha}^{\infty} \frac{Pe}{2\pi r \epsilon_{i}} = \frac{Fe}{2\pi \epsilon_{i}} \ln(2)
$$
  

$$
\bigvee_{2} = \bigvee(2\alpha) - \bigvee(\frac{1}{2}\alpha)\bigg|_{2} = -\int_{4\alpha}^{2\alpha} \frac{Pe}{2\pi r \epsilon_{2}} = \frac{Pe}{2\pi \epsilon_{2}} \ln(2)
$$

V2 = v...<br>
Total potential differente between conductors:  $V = V_1 + V_2 = \frac{\rho_1 Q_1}{2\pi} \left( \frac{1}{\epsilon_1} + \frac{1}{\epsilon_2} \right)$ <br>
Total change on time conductor  $Q = R$ <br>  $C = \frac{Q}{V} = \frac{2\pi R}{\lambda_1 (1)} = \frac{\epsilon_1 \epsilon_2}{\epsilon_1 \epsilon_2}$ Alternat :  $G = \frac{2\pi le_{1}}{ln(2)}$   $G = \frac{2\pi le_{2}}{ln(2)}$ 

 $C = \frac{1}{\frac{1}{c_1} + \frac{1}{c_2}} = \frac{2\pi e}{\frac{ln(n)}{c_1} + \frac{c_1c_2}{c_1} + \frac{1}{c_2}}$ Page 3 of 13

### Midterm

(b) Consider that the dielectric layers are lossy with conductivities  $\sigma_1$  and  $\sigma_2$  respectively. What is the resistance  $R$  between the inner and outer conductors?





(a)  $(20 \text{ points})$ For part (a), ignore the metal ring in the figure.

Consider a solenoid with N turns and length  $\ell$ , and radius  $a$ , is driven by a current source  $i(t)=i_0\cos(\omega t)$ . As shown in the figure, we use the convention that a positive current is associated with current in the  $\phi$  direction. You may assume that the solenoid is long  $(l \gg a)$ . What is the voltage difference  $v(t) = V_{+} - V$ , that appears across the terminals?

Your answer should be in terms of geometric, material, and fundamental parameters only (i.e. N,  $\ell$ ,  $a, b, \epsilon, \mu, \sigma$ , etc.).

Inductione	0-1	long	Solenoid	$L = \frac{\mu_0 N^2 A}{R} = \frac{\mu_0 N^3 \pi}{R}$
B-Reld $m \sin \theta$ solenoid = $\vec{B}_{m \sin \theta} = \frac{i(\theta) a_0 N}{R} \hat{Z}$				
1(14)	$\frac{d\Phi}{d\theta} = \pi a^2 \mu_0 N^2 \hat{H}$			
1(24)	$\frac{d\Phi}{d\theta} = -\frac{1}{2} \frac{d\mu_0 N \pi_0}{R} \sin(\omega t)$			
1(34)	$\frac{d\Lambda}{d\theta} = N \frac{d\Phi}{d\theta}$			
1(4)	$\frac{d\Phi}{d\theta} = -\frac{1}{2} \frac{d\mu_0 N \pi_0}{R} \sin(\omega t)$			
1(4)	$\frac{d\Lambda}{d\theta} = N \frac{d\Phi}{d\theta}$			
1(4)	$\frac{d\Lambda}{d\theta} = N \frac{d\Phi}{d\theta}$			
1(4)	$\frac{d\Lambda}{d\theta} = \frac{N \frac{d\Lambda}{d\theta}}{R}$			
1(4)	$\frac{d\Lambda}{d\theta} = \frac{N \frac{d\Lambda}{d\theta}}{R}$			
1(4)	$\frac{d\Lambda}{d\theta} = \frac{N \frac{d\Lambda}{d\theta}}{R}$			
1(4)	$\frac{d\Lambda}{d\theta} = \frac{N \frac{d\Lambda}{d\theta}}{R}$			
1(4)	$\frac{d\Lambda}{d\theta} = \frac{N \frac{d\Lambda}{d\theta}}{R}$			
1(4)	$\$			

Midterm

A metal ring with height h, radius b, and thickness d, and conductivity  $\sigma$  is placed (b)  $(20 \text{ points})$ around the center of the solenoid as shown in the figure. You may assume  $d \ll b$ . What is the current density  $J(t)$  that flows in the metal ring (give vector direction)? When answering this, you may neglect any contribution due the currents  $J(t)$  in the metal ring.

Consider Flow make Meth loop as a function of time.  
\nFrom part (a): B-field inside solenoid Bhs is in which of time.  
\n
$$
\oint \vec{E} d\vec{l} = \int_{0}^{4\pi} \vec{E}_{\phi} b d\phi = -\frac{d\Phi}{dt}
$$
\n
$$
\oint \vec{E} d\vec{l} = \int_{0}^{4\pi} \vec{E}_{\phi} b d\phi = -\frac{d\Phi}{dt}
$$
\n
$$
2\pi b \vec{E}_{\phi} = +i \frac{\omega u_{\phi} N \pi \delta^{2}}{R} \sin(\omega t)
$$
\n
$$
\vec{E}_{\phi}(t) = +i \frac{\omega u_{\phi} N \pi \delta^{2}}{R} \sin(\omega t)
$$
\n
$$
\vec{E}_{\phi}(t) = +i \frac{\omega u_{\phi} N \pi \delta^{2}}{R} \sin(\omega t)
$$
\n
$$
\vec{E}_{\phi}(t) = \frac{i \sigma u u_{\phi} N \pi}{2R} \sin(\omega t)
$$



increasing most rapidly,  $\tau_{\phi}$ <br>is negative + produces clockwae

Hive assume  $d26$ , then  $\vec{J}$  is approximately independent of  $\vec{r}$ .  $\vec{J}$  is independent of  $\phi$  by symmetry

(c) ( $\oint_0$  points) Does the approximation used in part (b) (neglecting any contribution due the currents  $J(t)$  in the metal ring) correspond to the low frequency limit ( $\omega$  is small compared to  $\tau_m^{-1}$ ), or the high fr

In the low frequency limit, we can neglect the  
field produced by the currents in the ring. This  
Contribution is proportional to the time, derivative of  
the flux, which increases linearly with 
$$
W
$$
 (Faraday's Law).  
For  $w \ll \tau_m^{-1}$   $\frac{d\Phi}{dt}$  is shkizently small  
that thefield produced by TH can be ignored.

The currents in the metal are driven by the solenoidal E-field,  
\nwhich one produced according to Faraday's Law, which are proportional to w.  
\n
$$
\frac{\partial}{\partial t} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
$$
\n
$$
\frac{\partial}{\partial t} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
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\frac{\partial}{\partial t} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
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\frac{\partial}{\partial t} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
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\frac{\partial}{\partial t} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
$$
\n
$$
\frac{\partial}{\partial \phi} = \sigma \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi} \frac{\partial}{\partial \phi}
$$

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### Table 3-1: Summary of vector relations.

Table 3-2: Coordinate transformation relations.



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# KERANTEN PINTEREEN GEGORIEGE GALD ANGEN GEFARERS CARTESIAN (RECTANGULAR) COORDINATES (x, y, z)

$$
\nabla V = \hat{\mathbf{x}} \frac{\partial V}{\partial x} + \hat{\mathbf{y}} \frac{\partial V}{\partial y} + \hat{\mathbf{z}} \frac{\partial V}{\partial z}
$$
  
\n
$$
\nabla \cdot \mathbf{A} = \frac{\partial A_x}{\partial x} + \frac{\partial A_y}{\partial y} + \frac{\partial A_z}{\partial z}
$$
  
\n
$$
\nabla \times \mathbf{A} = \begin{vmatrix} \hat{\mathbf{x}} & \hat{\mathbf{y}} & \hat{\mathbf{z}} \\ \frac{\partial}{\partial x} & \frac{\partial}{\partial y} & \frac{\partial}{\partial z} \\ A_x & A_y & A_z \end{vmatrix} = \hat{\mathbf{x}} \left( \frac{\partial A_z}{\partial y} - \frac{\partial A_y}{\partial z} \right) + \hat{\mathbf{y}} \left( \frac{\partial A_x}{\partial z} - \frac{\partial A_z}{\partial x} \right) + \hat{\mathbf{z}} \left( \frac{\partial A_y}{\partial x} - \frac{\partial A_x}{\partial y} \right)
$$
  
\n
$$
\nabla^2 V = \frac{\partial^2 V}{\partial x^2} + \frac{\partial^2 V}{\partial y^2} + \frac{\partial^2 V}{\partial z^2}
$$

# CYLINDRICAL COORDINATES  $(r, \phi, z)$

$$
\nabla V = \hat{\mathbf{r}} \frac{\partial V}{\partial r} + \hat{\Phi} \frac{1}{r} \frac{\partial V}{\partial \phi} + \hat{\mathbf{z}} \frac{\partial V}{\partial z}
$$
  
\n
$$
\nabla \cdot \mathbf{A} = \frac{1}{r} \frac{\partial}{\partial r} (r A_r) + \frac{1}{r} \frac{\partial A_{\phi}}{\partial \phi} + \frac{\partial A_{z}}{\partial z}
$$
  
\n
$$
\nabla \times \mathbf{A} = \frac{1}{r} \begin{vmatrix} \hat{\mathbf{r}} & \hat{\Phi}r & \hat{\mathbf{z}} \\ \frac{\partial}{\partial r} & \frac{\partial}{\partial \phi} & \frac{\partial}{\partial z} \\ A_r & r A_{\phi} & A_z \end{vmatrix} = \hat{\mathbf{r}} \left( \frac{1}{r} \frac{\partial A_{z}}{\partial \phi} - \frac{\partial A_{\phi}}{\partial z} \right) + \hat{\Phi} \left( \frac{\partial A_{r}}{\partial z} - \frac{\partial A_{z}}{\partial r} \right) + \hat{\mathbf{z}} \frac{1}{r} \left[ \frac{\partial}{\partial r} (r A_{\phi}) - \frac{\partial A_{r}}{\partial \phi} \right]
$$
  
\n
$$
\nabla^{2} V = \frac{1}{r} \frac{\partial}{\partial r} \left( r \frac{\partial V}{\partial r} \right) + \frac{1}{r^{2}} \frac{\partial^{2} V}{\partial \phi^{2}} + \frac{\partial^{2} V}{\partial z^{2}}
$$

# SPHERICAL COORDINATES  $(R, \theta, \phi)$

$$
\nabla V = \hat{\mathbf{R}} \frac{\partial V}{\partial R} + \hat{\theta} \frac{1}{R} \frac{\partial V}{\partial \theta} + \hat{\phi} \frac{1}{R \sin \theta} \frac{\partial V}{\partial \phi}
$$
  
\n
$$
\nabla \cdot \mathbf{A} = \frac{1}{R^2} \frac{\partial}{\partial R} (R^2 A_R) + \frac{1}{R \sin \theta} \frac{\partial}{\partial \theta} (A_\theta \sin \theta) + \frac{1}{R \sin \theta} \frac{\partial A_\phi}{\partial \phi}
$$
  
\n
$$
\nabla \times \mathbf{A} = \frac{1}{R^2 \sin \theta} \begin{vmatrix} \hat{\mathbf{R}} & \hat{\theta}R & \hat{\phi}R \sin \theta \\ \frac{\partial}{\partial R} & \frac{\partial}{\partial \theta} & \frac{\partial}{\partial \phi} \\ A_R & R A_\theta & (R \sin \theta) A_\phi \end{vmatrix}
$$
  
\n
$$
= \hat{\mathbf{R}} \frac{1}{R \sin \theta} \left[ \frac{\partial}{\partial \theta} (A_\phi \sin \theta) - \frac{\partial A_\theta}{\partial \phi} \right] + \hat{\theta} \frac{1}{R} \left[ \frac{1}{\sin \theta} \frac{\partial A_R}{\partial \phi} - \frac{\partial}{\partial R} (R A_\phi) \right] + \hat{\phi} \frac{1}{R} \left[ \frac{\partial}{\partial R} (R A_\theta) - \frac{\partial A_R}{\partial \theta} \right]
$$
  
\n
$$
\nabla^2 V = \frac{1}{R^2} \frac{\partial}{\partial R} \left( R^2 \frac{\partial V}{\partial R} \right) + \frac{1}{R^2 \sin \theta} \frac{\partial}{\partial \theta} \left( \sin \theta \frac{\partial V}{\partial \theta} \right) + \frac{1}{R^2 \sin^2 \theta} \frac{\partial^2 V}{\partial \phi^2}
$$

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### <u>SHI MATA GARA DI SA</u> Robits ser

Scalar (or dot) product  $\mathbf{A} \cdot \mathbf{B} = AB \cos \theta_{AB}$ Vector (or cross) product,  $\hat{\mathbf{n}}$  normal to plane containing A and  $\mathbf{B}$  $A \times B = \hat{n}AB \sin \theta_{AB}$  $A \cdot (B \times C) = B \cdot (C \times A) = C \cdot (A \times B)$  $A \times (B \times C) = B(A \cdot C) - C(A \otimes B)$  $\nabla(U+V) = \nabla U + \nabla V$  $\nabla(UV) = U \nabla V + V \nabla U$  $\nabla \cdot (\mathbf{A} + \mathbf{B}) = \nabla \cdot \mathbf{A} + \nabla \cdot \mathbf{B}$  $\nabla \cdot (U\mathbf{A}) = U \nabla \cdot \mathbf{A} + \mathbf{A} \cdot \nabla U$  $\nabla \times (U\mathbf{A}) = U\nabla \times \mathbf{A} + \nabla U \times \mathbf{A}$  $\nabla \times (\mathbf{A} + \mathbf{B}) = \nabla \times \mathbf{A} + \nabla \times \mathbf{B}$  $\nabla \cdot (\mathbf{A} \times \mathbf{B}) = \mathbf{B} \cdot (\nabla \times \mathbf{A}) - \mathbf{A} \cdot (\nabla \times \mathbf{B})$  $\nabla \cdot (\nabla \times \mathbf{A}) = 0$  $\nabla \times \nabla V = 0$  $\nabla \cdot \nabla V = \nabla^2 V$  $\nabla \times \nabla \times \mathbf{A} = \nabla (\nabla \cdot \mathbf{A}) - \nabla^2 \mathbf{A}$  $\int_{\mathcal{V}} (\nabla \cdot \mathbf{A}) d\mathcal{V} = \oint_{S} \mathbf{A} \cdot d\mathbf{s}$ Divergence theorem ( $S$  encloses  $\nu$ )  $\int_{S} (\nabla \times \mathbf{A}) \cdot d\mathbf{s} = \oint_{C} \mathbf{A} \cdot d\mathbf{l}$ Stokes's theorem  $(S$  bounded by  $C$ )

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