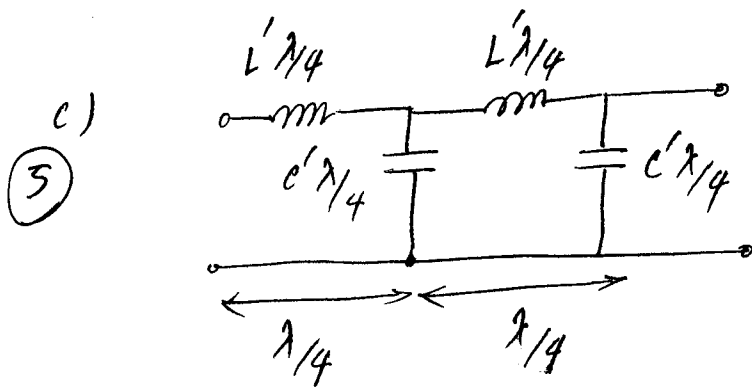


5) $Z_0 = \sqrt{\frac{L'}{C'}} = 50 \rightarrow L' = 1.67 \times 10^{-7} \text{ (H)}$
 $L' C' = \mu \epsilon \rightarrow C' = 6.67 \times 10^{-11} \text{ (F)}$

b) $u_p = c \text{ (air)}$

5) $\lambda = \frac{c}{f} = \frac{3 \times 10^8}{3 \times 10^8} = 1 \text{ (m)}$ $\beta = \frac{2\pi}{\lambda} = 2\pi \text{ (1/m)} = 6.287$



d) $\epsilon_r = 4$ $L' = 2 (L' \text{ in part a}) = 3.34 \times 10^{-7} \text{ (H)}$
 10) $C' = 2 (C' \text{ in part a}) = 1.33 \times 10^{-10} \text{ (F)}$

b) $u_p = \frac{c}{\sqrt{\epsilon_r}} = \frac{c}{2} = 1.5 \times 10^8 \text{ (m/s)}$

$\lambda = \frac{u_p}{f} = \frac{1.5 \times 10^8}{300 \times 10^6} = 0.5 \text{ (m)}$ $\beta = \frac{2\pi}{\lambda} = 4\pi = 12.57 \text{ (m)}$

2)

$$a) \Gamma = \frac{Z_L - Z_0}{Z_L + Z_0} = \frac{60 - 120j - 100}{60 - 120j + 100} = 0.2 - 0.6j = 0.6325 e^{-j1.25}$$

(4)

$$b) S = \frac{1 + 0.6325}{1 - 0.6325} = 4.44$$

(4)

$$c) V_{\max} @ \frac{\theta_r \lambda}{4\pi} + \frac{n\lambda}{2} \rightarrow l_{\max} = \frac{-1.25\lambda}{4\pi} + \frac{1 \times \lambda}{2} = 2 \text{ cm}$$

(4) $l_{\min} \text{ for voltage} = l_{\max} \text{ for max current}$

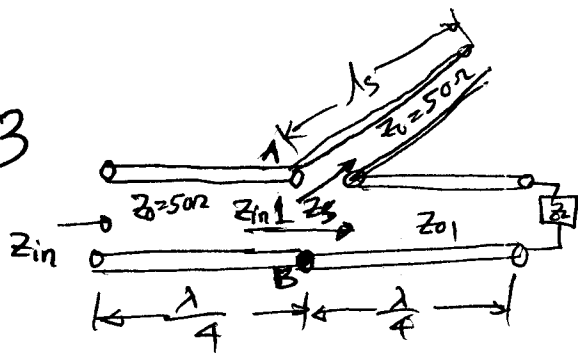
$$l_{\min} = 2 - \lambda/4 = 0.75 \text{ cm}$$

d) No, $S = \frac{1 + |\Gamma|}{1 - |\Gamma|}$ and since $|\Gamma|$ remains the same (only its phase changes) $\rightarrow S$ remains constant.

$$e) |V_0^+| = 1 \rightarrow |V_{\max}| = |V_0^+| (1 + |\Gamma|) = 1.6325 \text{ (Volts)}$$

(4) $|V_{\min}| = |V_0^+| (1 - |\Gamma|) = 0.3675 \text{ (V)}$

3



(a) For Quarter wavelength transformer

5 points

$$Z_{in1} = \frac{Z_0^2}{Z_L} = \frac{(50)^2}{100} = 25 \text{ } (\Omega)$$

For open circuit of length $l_s = \frac{\lambda}{4}$

$$Z_s = Z_{in}^{oc} = -jZ_0 \cot \beta l = -jZ_0 \cot\left(\frac{\pi}{2}\right) = 0 \text{ } (\Omega)$$

\therefore Total Input Impedance at AB Z_{in}' is

$$Z_{in}' = Z_{in1} + Z_s = 25 + 0 = 25 \text{ } (\Omega)$$

$\therefore Z_{in} = \frac{Z_0^2}{Z_{in}'}$ (Quarter wavelength transformer)

$$= \frac{(50)^2}{25} = 100 \text{ } (\Omega)$$

(b) To impedance match

5 points

$$Z_{in}' = Z_0 = 50$$

$$\therefore Z_{in}' = Z_{in1} + Z_s = 50 \quad \therefore Z_{in1} = 50$$

$$Z_{in1} = \frac{Z_0^2}{Z_L}$$

$$\therefore Z_0 = \sqrt{Z_{in1} \times Z_L} = \sqrt{50 \times 100} = 70.7 \text{ } (\Omega)$$

(c) Now $Z_L = (100 - j100) \Omega$.

10 points

To impedance match

$$Z_{in}' = Z_{in1} + Z_s = 50$$

$$Z_{in1} = \frac{(Z_0)^2}{Z_L} = \frac{(Z_0)^2}{100 - j100} = \frac{(Z_0)^2(100 + j100)}{(100)^2 + (100)^2} = \frac{(Z_0)^2(1+j)}{200}$$

and

$$Z_s = -jZ_0 \cot \beta l_s$$

$$\therefore Z_{in}' = Z_{in1} + Z_s = 50$$

$$\Rightarrow \frac{(Z_0)^2(1+j)}{200} + (-jZ_0 \cot \beta l_s) = 50$$

$$\frac{(Z_0)^2}{200} + \frac{j(Z_0)^2}{200} - jZ_0 \cot \beta l_s = 50$$

to satisfy this equation

$$\frac{(Z_0)^2}{200} = 50$$

$$\text{and } \frac{j(Z_0)^2}{200} - jZ_0 \cot \beta l_s = j0$$

$$\therefore Z_0 = \sqrt{50 \times 200} = 100 \Omega$$

$$\text{and } \frac{(Z_0)^2}{200} = Z_0 \cot \beta l_s$$

$$\Rightarrow \frac{(100)^2}{200} = 50 \cot \beta l_s$$

$$\therefore \cot \beta l_s = 1 \quad \therefore \cot \beta l_s = \cot \frac{\pi}{4}$$

$$\therefore \frac{2\pi}{\lambda} l_s = \frac{\pi}{4}, \quad l_s = \frac{\lambda}{8}$$

$$\therefore l_s = n \frac{\lambda}{2} + \frac{\lambda}{8} \quad (n=0, 1, 2, \dots)$$

$$\boxed{Z_0 = 100 \Omega, \quad l_s = n \frac{\lambda}{2} + \frac{\lambda}{8}}$$

4

$$Z_L = (50 - j25) \Omega \quad (\bar{z} \text{ represents normalized value on } z)$$

4 points

$$\therefore \text{Normalized Impedance } \bar{z}_L = 1 - j0.5$$

Plotting this value in Smith chart, we get

$$r \approx 0.24 \angle -76^\circ$$

(b) SWR circle intersects two points in real axis r_r .

4 points

Point in the right hand side gives $r_r > 1$. ($S = r_r \geq 1$)

$$\therefore S \approx 1.63$$

(c) Center and \bar{z}_L point connected line intersects the WTL scale at 0.356λ .

4 points

\bar{z}_{in} at ~~0.3~~ 0.3λ from the load ~~is~~ will be at

$$0.356\lambda + 0.3\lambda - 0.5\lambda = 0.156\lambda \text{ point.}$$

$$\therefore \bar{z}_{in} \approx 1.08 + j0.5$$

$$\therefore Z_{in} = (1.08 + j0.5) \cdot 50 = 54 + j25 (\Omega).$$

(d) Rotating the impedance by $\frac{\lambda}{4}$, we get the admittance.

4 points

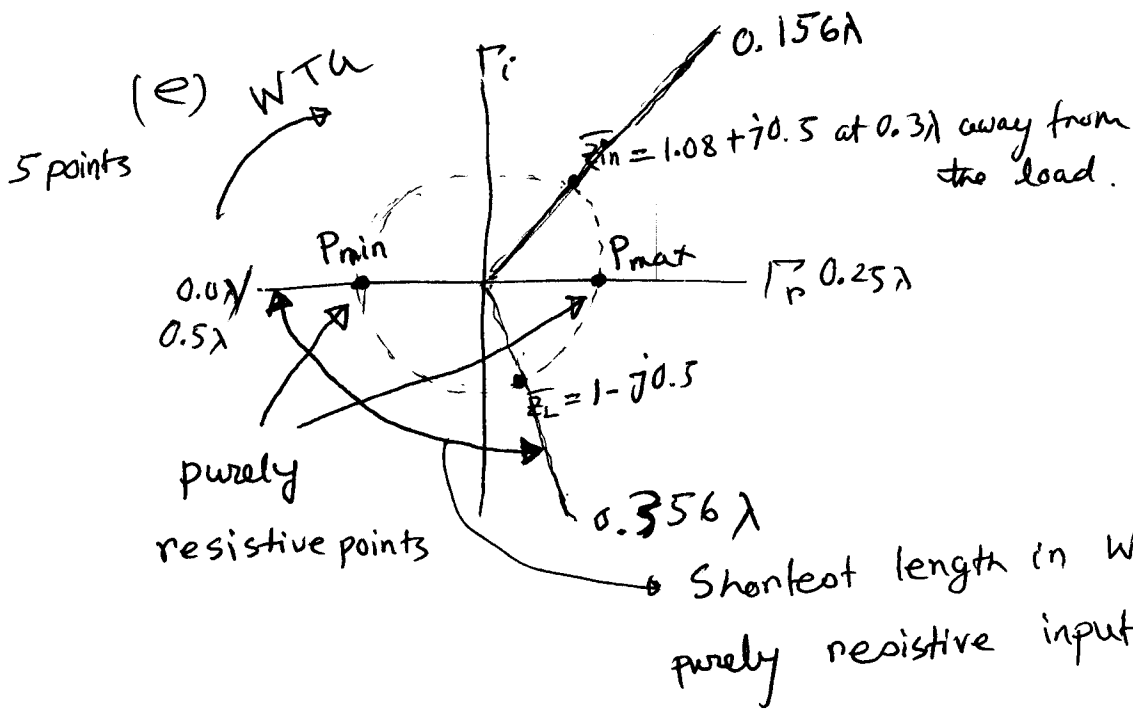
$$\therefore \bar{y}_{in} \text{ will be at } 0.156\lambda + 0.25\lambda = 0.406\lambda$$

$$\text{From Smith Chart } \bar{y}_{in} \approx 0.77 - j0.35$$

$$\therefore Y_{in} = \bar{y}_{in} Y_0 = \bar{y}_{in} / Z_0 = \frac{0.77 - j0.35}{50}$$

$$= 0.0154 - j0.007 \text{ S}$$

$$\left[Y = \frac{Y}{Y_0} = \frac{G}{Y_0} + j \frac{B}{Y_0} = g + jb, \therefore Y = Y_0 g + j Y_0 b = \frac{g}{Z_0} + j \frac{b}{Z_0} \right]$$



\therefore Shortest length $0.5\lambda - 0.356\lambda = 0.144\lambda$.

(f) Voltage maximum at P_{max} point

4 points

\therefore Position is $0.144\lambda + 0.25\lambda = 0.394\lambda$

(you can calculate as $0.25\lambda - 0.356\lambda + 0.5\lambda = 0.394\lambda$)

5 (a) According to Stokes's theorem for vector quantity A .

5 points

$$\int_S (\nabla \times A) \cdot dS = \oint_C A \cdot dL \quad [A \text{ represents vector of } A]$$

$$\therefore \oint_C E \cdot dl = \int_S (\nabla \times E) \cdot dS$$

Faraday's law gives $\oint_C E \cdot dl = - \int_S \frac{\partial B}{\partial t} \cdot dS$

$$\therefore \int_S (\nabla \times E) \cdot dS = - \int_S \frac{\partial B}{\partial t} \cdot dS$$

Taking derivative on both side (S arbitrary)

$$\boxed{\nabla \times E = - \frac{\partial B}{\partial t}}$$

(b)
5 points

From divergence theorem, for vector quantity \mathbf{A}

$$\int_V \nabla \cdot \mathbf{A} \, dv = \oint_S \mathbf{A} \cdot d\mathbf{s}$$

$$\therefore \int_S \mathbf{D} \cdot d\mathbf{s} = \int_V \nabla \cdot \mathbf{D} \, dv$$

From Gauss's law

$$\int_S \mathbf{D} \cdot d\mathbf{s} = \int_V \rho_v \, dv$$

$$\therefore \int_V \nabla \cdot \mathbf{D} \, dv = \int_V \rho_v \, dv$$

(Volume arbitrary)

Take derivative in both side

$$\boxed{\nabla \cdot \mathbf{D} = \rho_v}$$